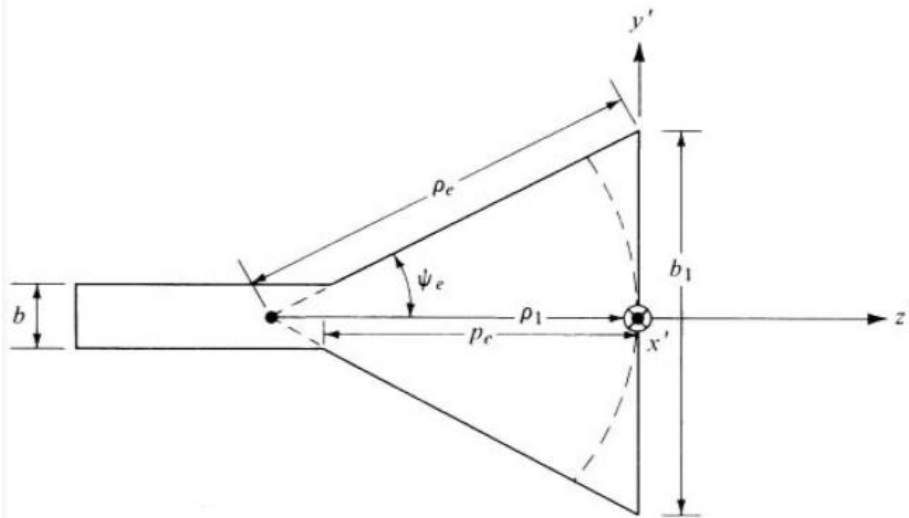
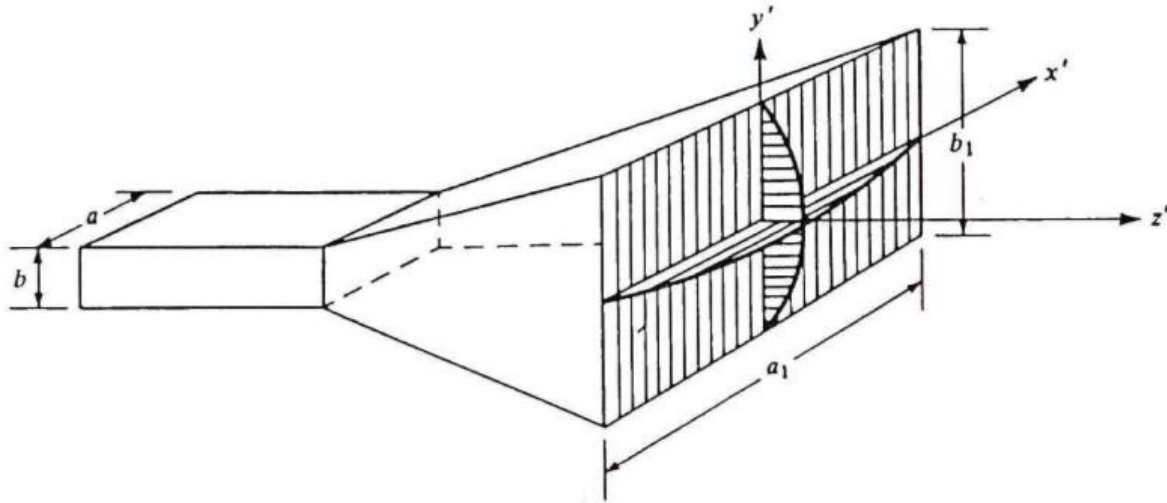
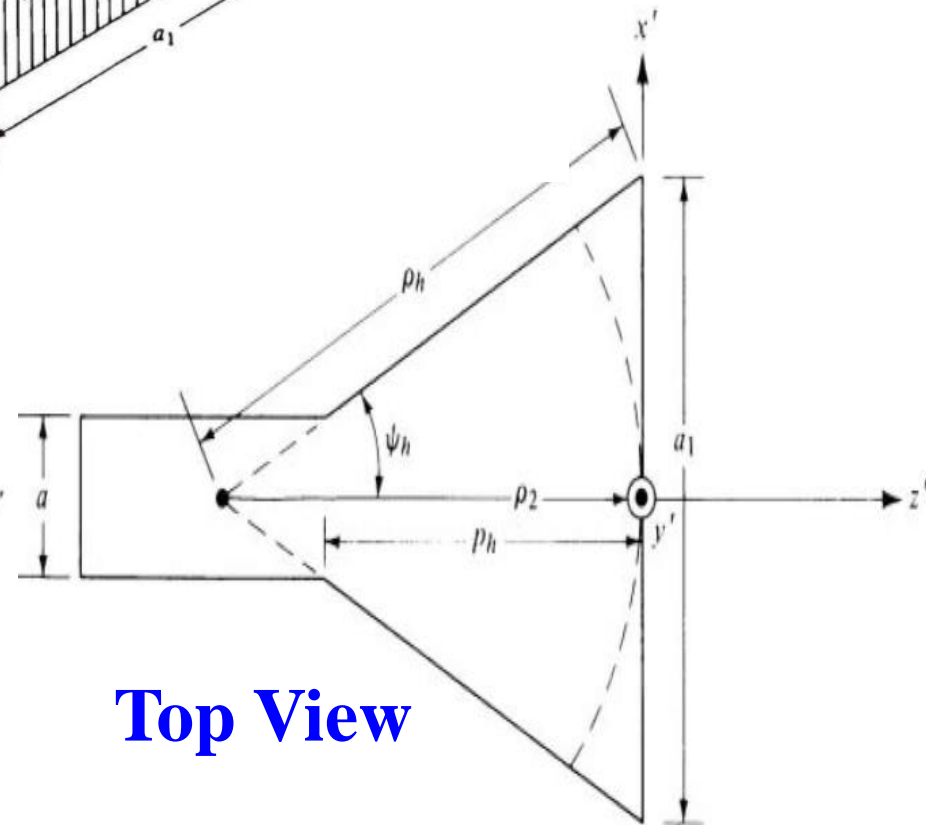


Pyramidal Horn Antenna



Side View



Top View

Pyramidal Horn Antenna

$$E'_y(x', y') = E_0 \cos\left(\frac{\pi}{a_1} x'\right) e^{-j\left[k\left(\frac{x'^2}{2\rho_2} + \frac{y'^2}{2\rho_1}\right)\right]}$$

Condition for Physical Realization:

$$p_e = (b_1 - b) \left[\left(\frac{\rho_e}{b_1} \right)^2 - \frac{1}{4} \right]^{1/2}$$

$$p_h = (a_1 - a) \left[\left(\frac{\rho_h}{a_1} \right)^2 - \frac{1}{4} \right]^{1/2}$$

$$p_e = p_h$$

Pyramidal Horn: Design Procedure

Directivity of
Pyramidal Horn
Antenna can be
obtained using
Directivity
curves for E-and
H-Planes
Sectoral Horn
antenna

$$D_p = \frac{\pi \lambda^2}{32ab} D_E D_H$$

Alternatively

$$G_0 \approx \frac{1}{2} \left(\frac{4\pi}{\lambda^2} a_1 b_1 \right)$$

$$a_1 \approx \sqrt{3\lambda\rho_2} \approx \sqrt{3\lambda\rho_h} \quad \rho_2 \approx \rho_h$$

$$b_1 \approx \sqrt{2\lambda\rho_1} \approx \sqrt{2\lambda\rho_e} \quad \rho_1 \approx \rho_e$$

$$\rho_e = (b_1 - b) \sqrt{\left(\frac{\rho_e}{b_1}\right)^2 - \frac{1}{4}}$$

$$\rho_h = (a_1 - a) \sqrt{\left(\frac{\rho_h}{a_1}\right)^2 - \frac{1}{4}}$$

Pyramidal Horn Design Steps

$$\left(\sqrt{2\chi} - \frac{b}{\lambda}\right)^2 (2\chi - 1) = \left[\frac{G_0}{2\pi} \sqrt{\frac{3}{2\pi}} \frac{1}{\sqrt{\chi}} - \frac{a}{\lambda}\right]^2 \left[\frac{G_0^2}{6\pi^3} \frac{1}{\chi} - 1\right]$$

$$\rho_e = \chi \lambda \Rightarrow \chi = \frac{\rho_e}{\lambda}$$

$$\rho_h = \frac{G_0^2}{8\pi^3} \left(\frac{1}{\chi}\right) \lambda$$

$$1. \chi \approx \chi_1 = x(\text{trial}) = \frac{G_0}{2\pi \sqrt{2\pi}}$$

$$2. \rho_e = \chi \lambda, \rho_h = \frac{G_0^2}{8\pi^3} \frac{1}{\chi} \lambda$$

$$3. a_1 = \sqrt{3\lambda\rho_2} \approx \sqrt{3\lambda\rho_h} = \frac{G_0}{2\pi} \sqrt{\frac{3}{2\pi\chi}} \lambda$$

$$b_1 = \sqrt{2\lambda\rho_1} \approx \sqrt{2\lambda\rho_e} = \sqrt{2\chi\lambda}$$

$$4. p_e, p_h$$

Pyramidal Horn Design: Example

Given: X-Band (8.2-12.4 GHz), $f = 11$ GHz Horn; Gain = 22.6 dB

$a = 0.9$ in (2.286 cm), $b = 0.4$ in (1.016 cm)

Find: Dimensions Of Pyramidal Horn

Solution

$$G_0(\text{dB}) = 22.6 = 10 \log_{10} G_0 \Rightarrow G_0 = 10^{2.26} = 181.97$$

$$\text{At } f = 11 \text{ GHz} \Rightarrow \lambda = \frac{30 \times 10^9}{11 \times 10^9} = 2.7273 \text{ cm}$$

$$b = \frac{1.016}{2.7273} \lambda = 0.3725 \lambda; \quad a = \frac{2.286}{2.7273} \lambda = 0.8382 \lambda$$

Pyramidal Horn Design: Example (Contd.)

1. Initial value of χ

$$\chi_1 = \frac{G_0}{2\pi\sqrt{2\pi}} = \frac{181.97}{2\pi\sqrt{2\pi}} = 11.5539$$

which does not satisfy(12-56), or

$$\left(\sqrt{2\chi} - \frac{b}{\chi}\right)^2 (2\chi - 1) = \left(\frac{G_0}{2\pi} \sqrt{\frac{3}{2\pi}} \frac{1}{\sqrt{\chi}} - \frac{a}{\lambda}\right)^2 \left(\frac{G_0^2}{6\pi^3} \frac{1}{\chi} - 1\right)$$

After few tries, a more accurate value is

$$\chi = 11.1157$$

2. $\rho_e = \chi\lambda = 11.1157\lambda = 30.316 \text{ cm} = 11.935 \text{ in.}$

$$\rho_h = \frac{G_0^2}{8\pi^3} \left(\frac{1}{\chi}\right) \lambda = 12.0094\lambda = 32.753 \text{ cm} = 12.895 \text{ in.}$$

Pyramidal Horn Design: Example (Contd.)

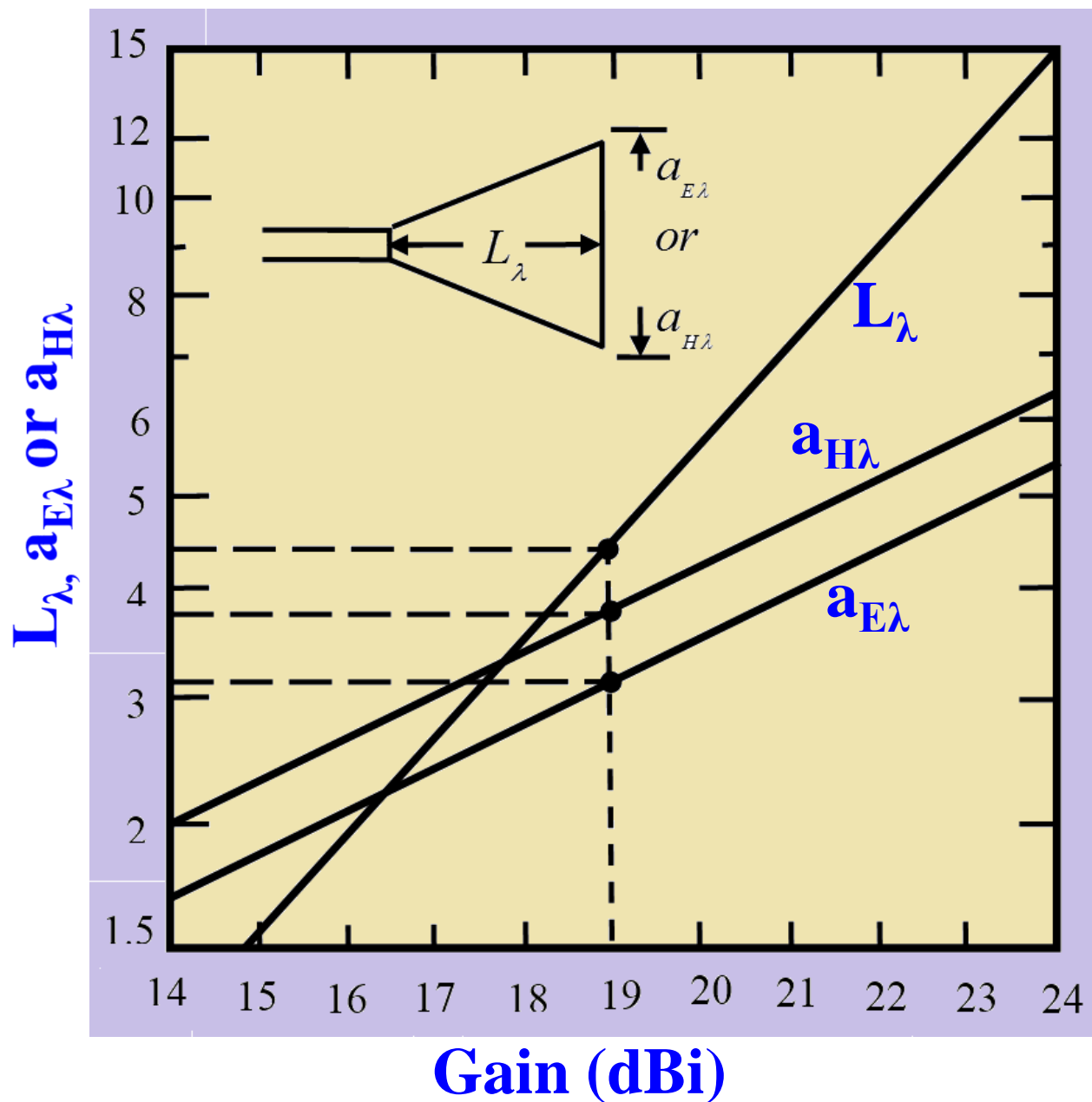
$$3. \quad a_1 = \sqrt{3\lambda\rho_2} \approx \sqrt{3\lambda\rho_h} = \frac{G_0}{2\pi} \sqrt{\frac{3}{2\pi\chi}} \lambda = 6.002\lambda$$
$$= 16.370 \text{ cm} = 6.445 \text{ in.}$$

$$b_1 = \sqrt{2\lambda\rho_1} \approx \sqrt{2\lambda\rho_e} = \sqrt{2\chi} \lambda = 4.715\lambda$$
$$= 12.859 \text{ cm} = 5.063 \text{ in.}$$

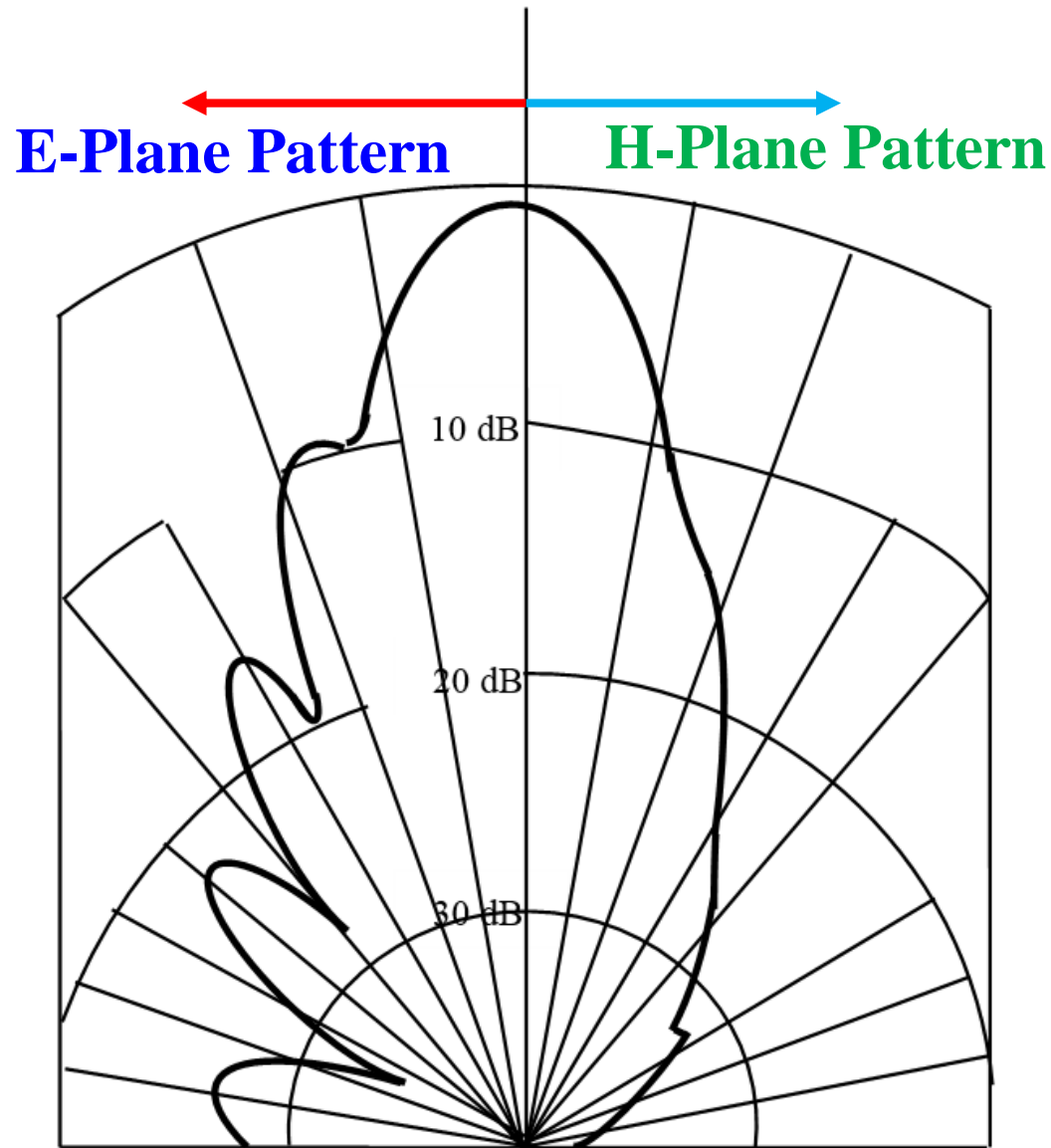
$$4. \quad p_e = (b_1 - b) \left[\left(\frac{p_e}{b_1} \right)^2 - \frac{1}{4} \right]^{1/2} = 10.005\lambda$$
$$= 27.286 \text{ cm} = 10.743 \text{ in.}$$

$$p_h = (a_1 - a) \left[\left(\frac{p_h}{a_1} \right)^2 - \frac{1}{4} \right]^{1/2} = 10.005\lambda$$
$$= 27.286 \text{ cm} = 10.743 \text{ in.}$$

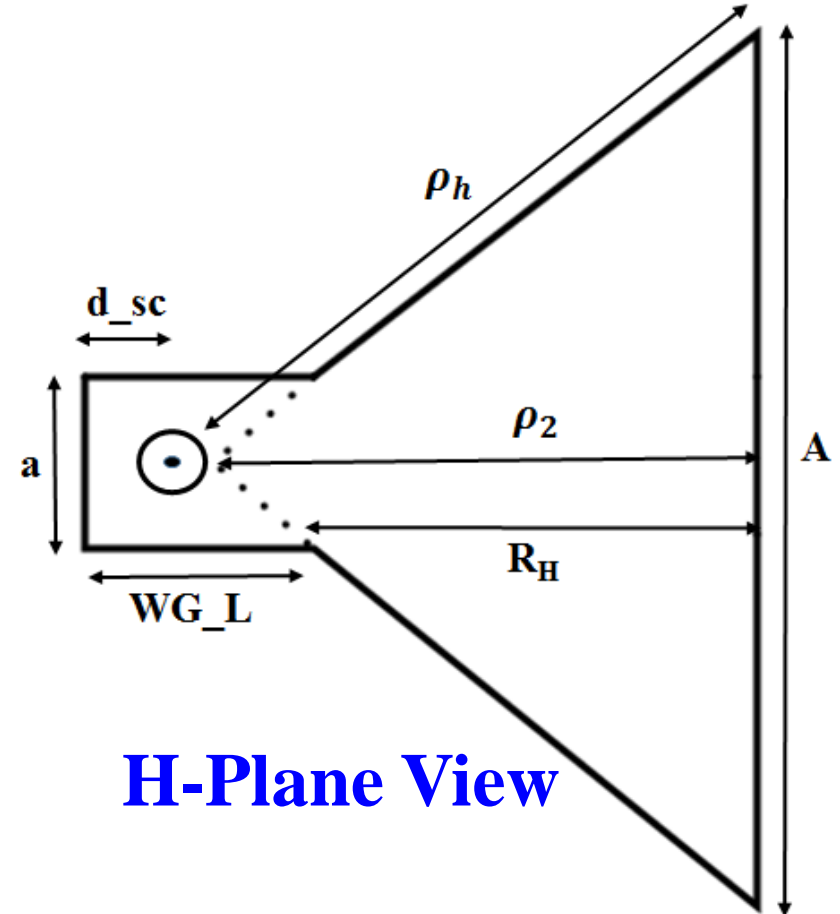
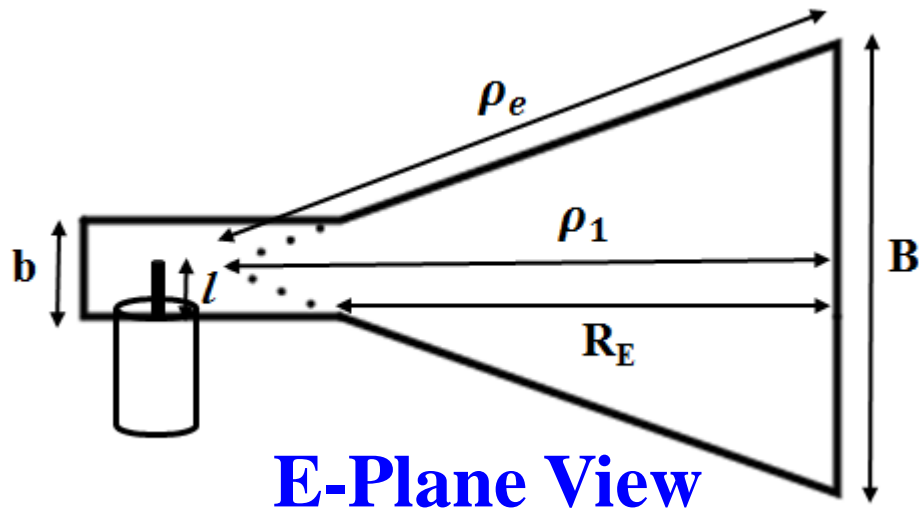
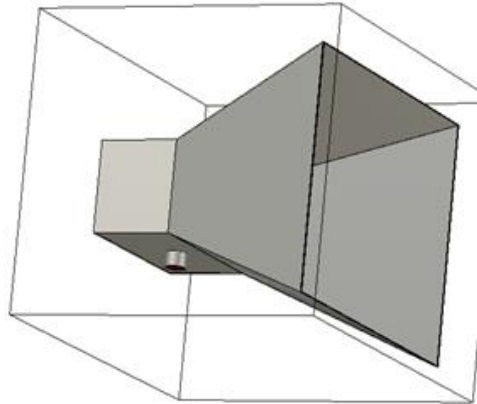
Optimum Dimensions vs. Directivity



Radiation Pattern of Pyramidal Horn Antenna



Coaxial Feed Pyramidal Horn Antenna

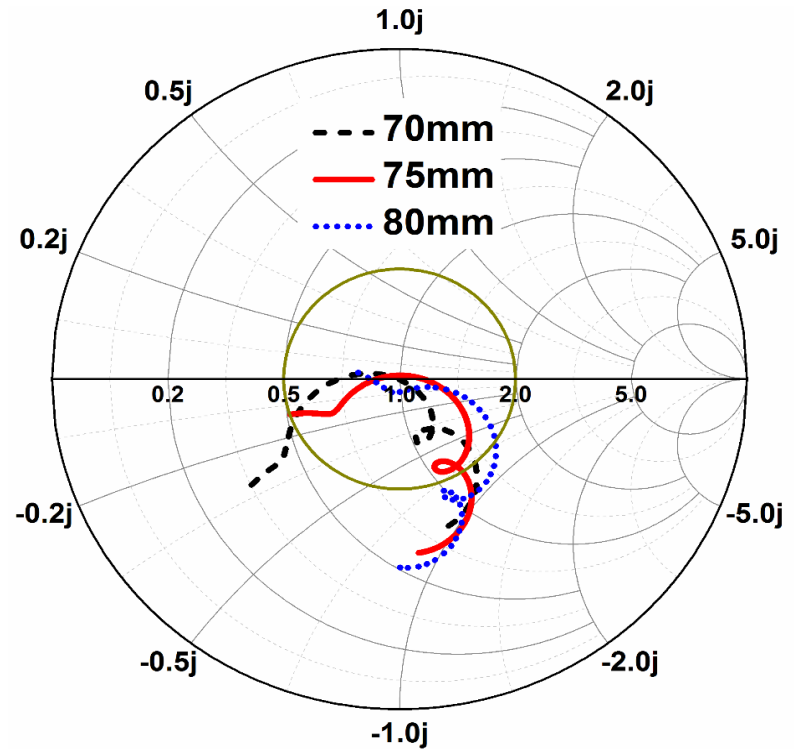
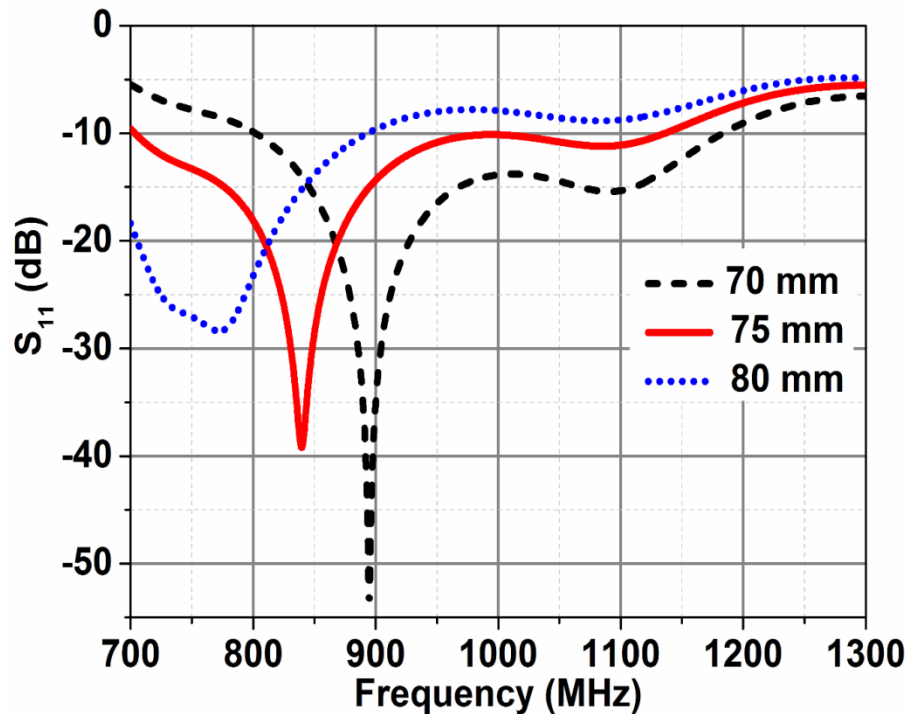


Reference: Hemant Kumar and Girish Kumar, “Design and Parametric Analysis of Pyramidal Horn Antenna with High Efficiency”, Proceedings of International Symposium on Microwave and Optical Technology (ISMOT) 2015, pp. 134-137.

Coaxial Feed Pyramidal Horn Antenna Designed at 900 MHz

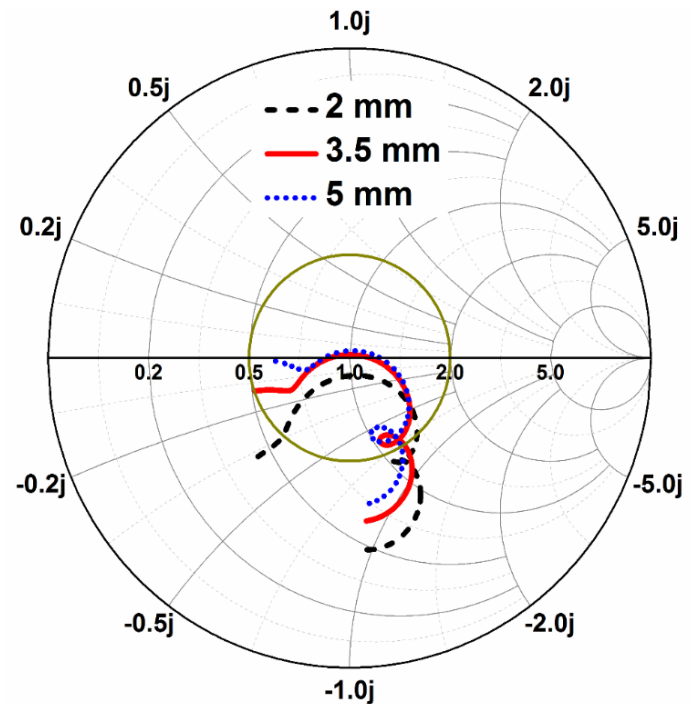
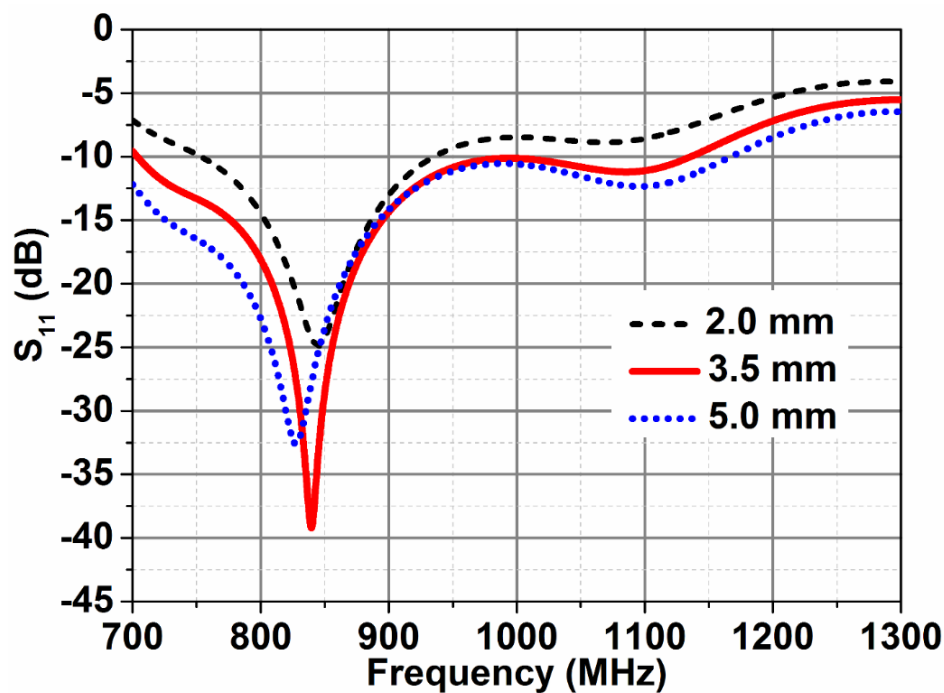
<i>Parameter</i>	Value (mm)	Description
<i>A</i>	450	Aperture Width
<i>B</i>	320	Aperture Height
<i>a</i>	240	Waveguide Width
<i>b</i>	120	Waveguide Height
<i>WG_L</i>	110	Waveguide Length
$R_E = R_H$	250	Horn Length
<i>l</i>	75	Probe Length
<i>r</i>	3.5	Probe Radius
<i>d_sc</i>	67.5	Distance of feed from short

Effect of Probe Feed Length



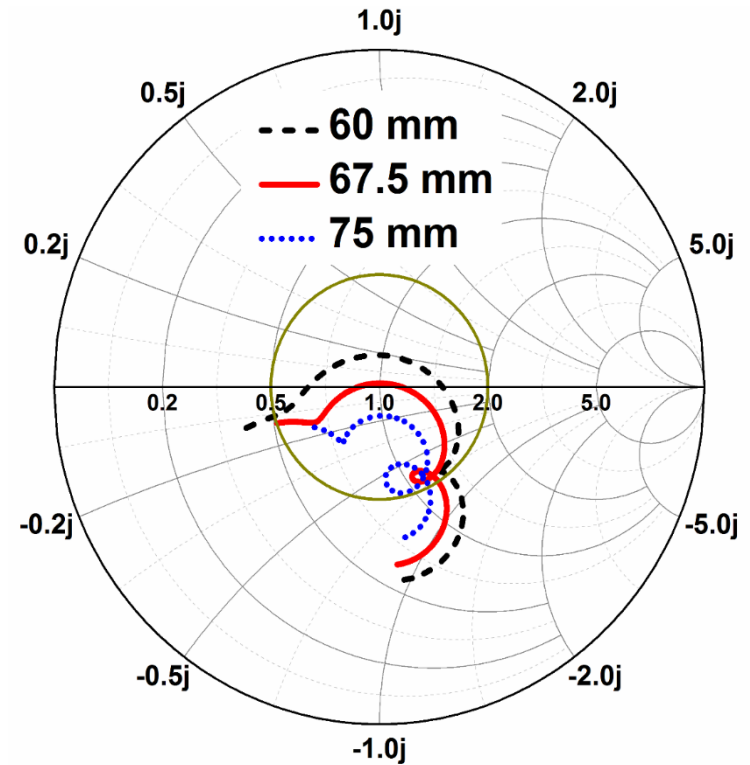
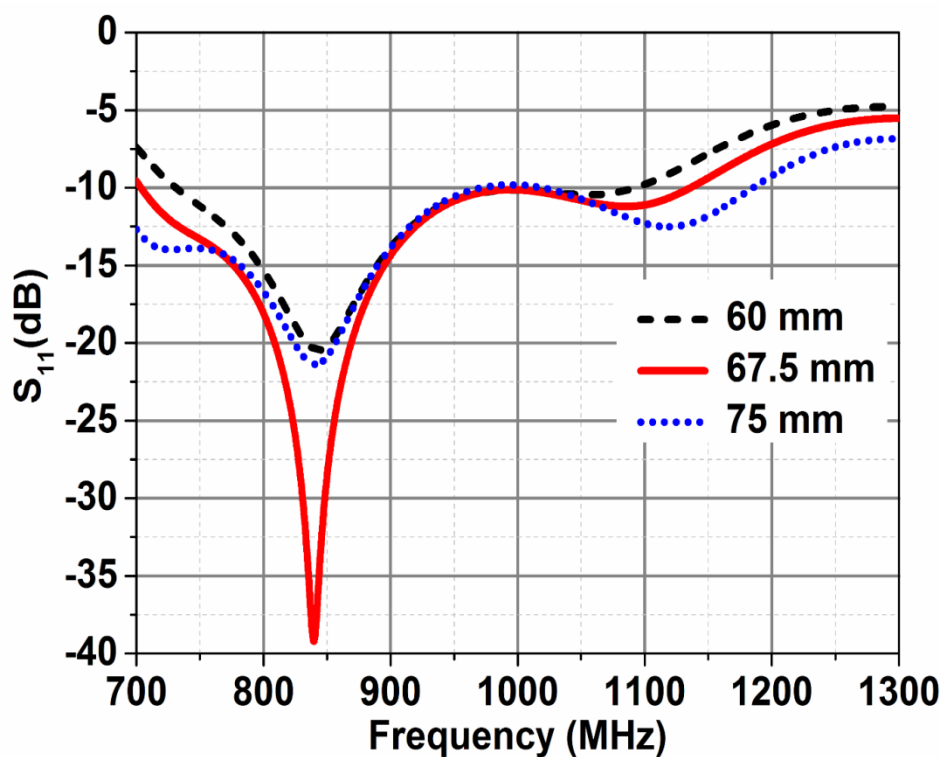
As the probe length increases from 70 to 80 mm, the resonance frequency decreases from 895 to 790 MHz and the input impedance curve rotates clockwise.

Effect of Probe Feed Radius



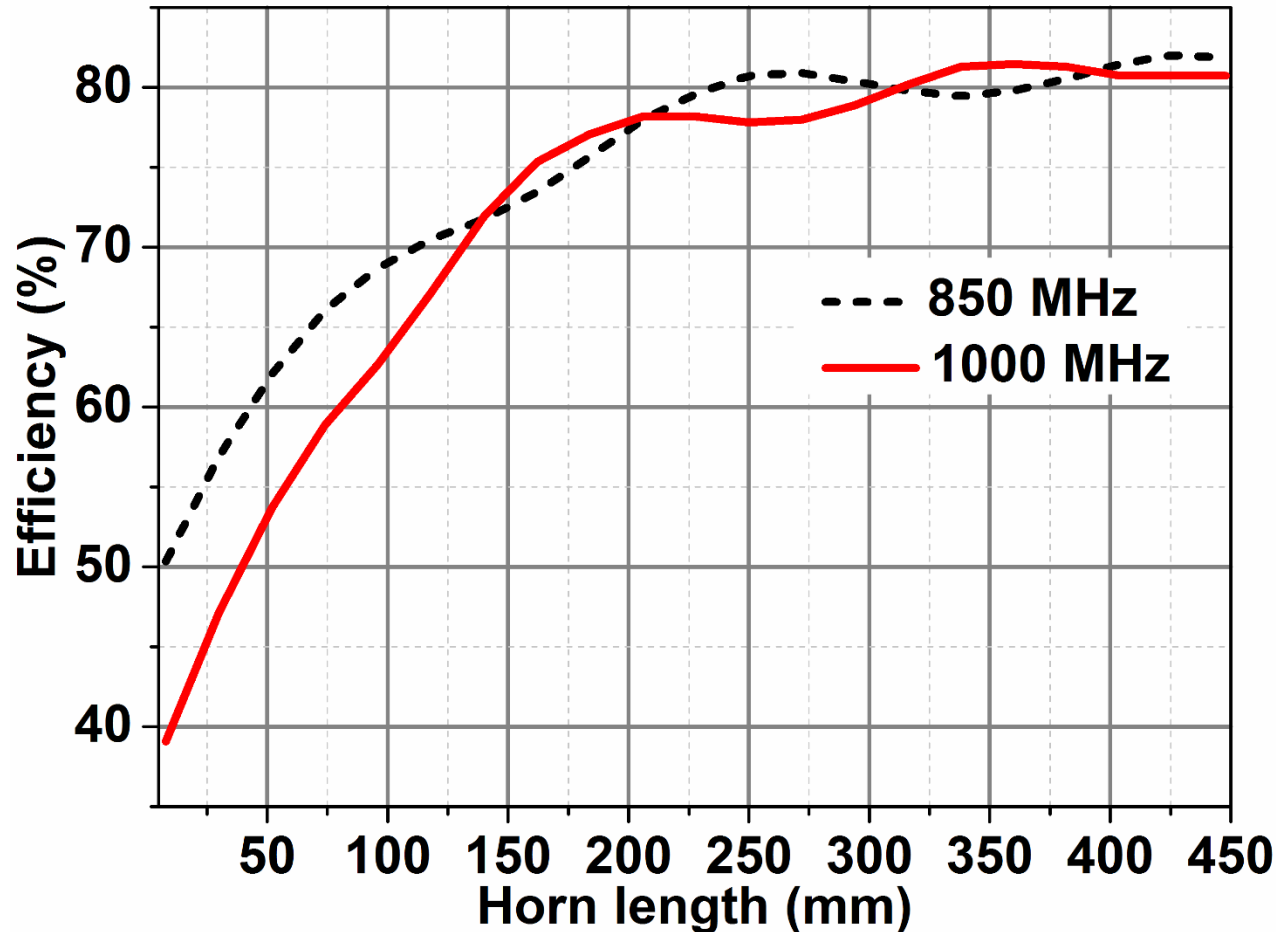
As the probe radius increases from 2 to 5mm, the resonance frequency decreases slightly due to increase in the fringing fields and bandwidth increases.

Effect of Probe Feed Location



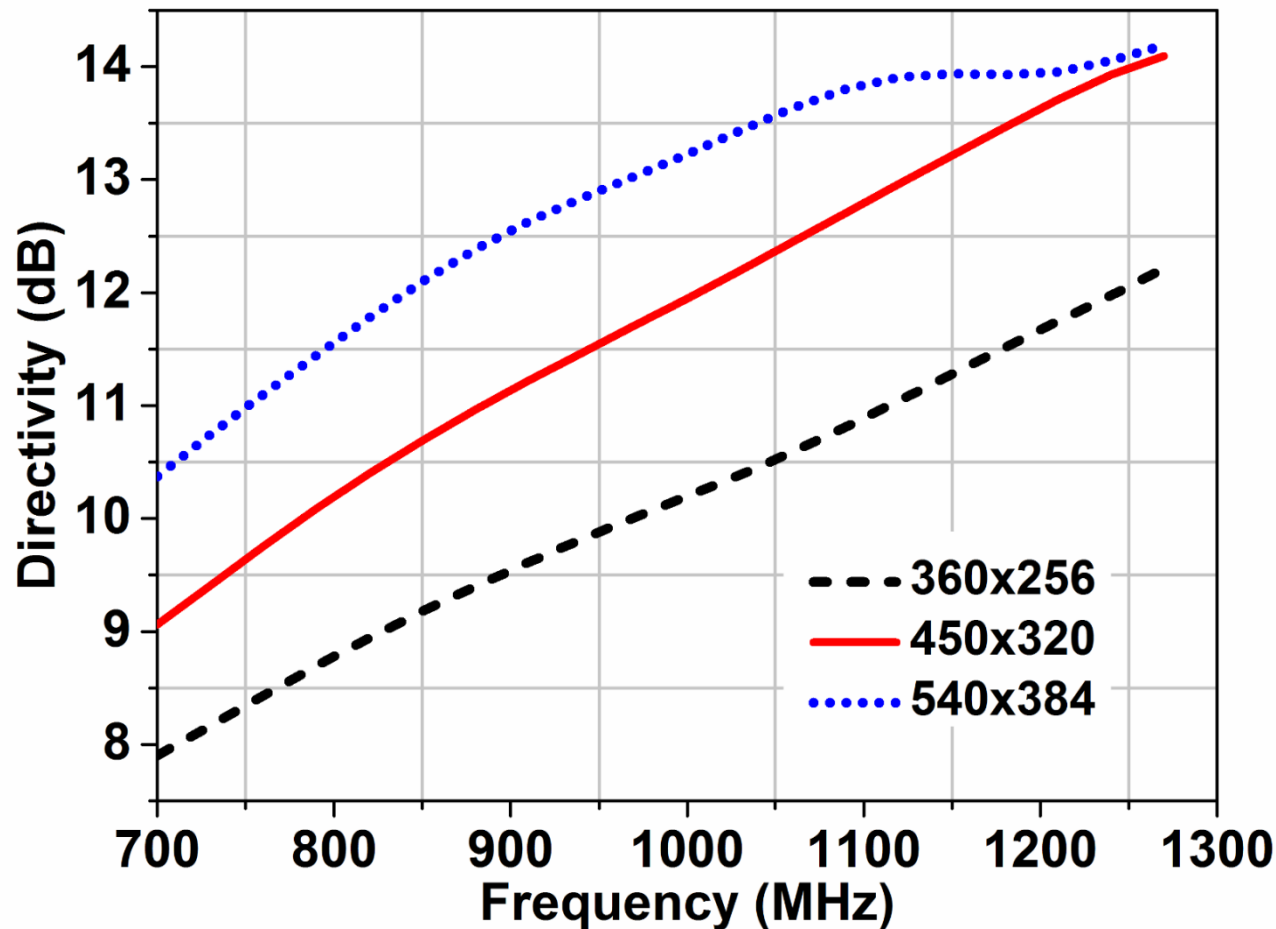
As the probe feed location is moved towards shorting wall (i.e., decreased from 75 to 60 mm), the input impedance becomes inductive so the curve shifts upward.

Effect of Horn Length on Efficiency



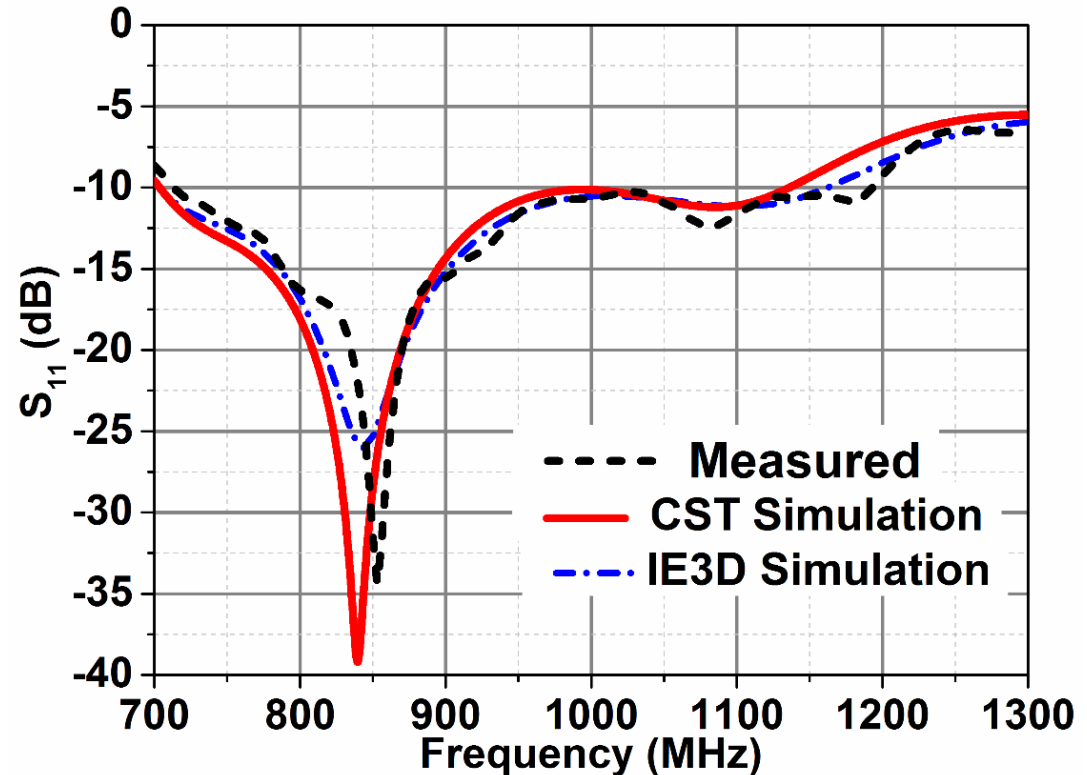
For Horn Length $R_E = R_H > 150$ mm, efficiency $\geq 72\%$
and for $R_E = R_H > 250$ mm, efficiency $\approx 80\%$

Effect of Horn Aperture on Directivity



As aperture area increases, directivity increases. But for larger aperture as frequency increases, phase error increases, which decreases the gain of the horn antenna.

Simulated and Measured S_{11} of Coaxial Feed Pyramidal Horn Antenna



Bandwidth for $S_{11} < -10$ dB :

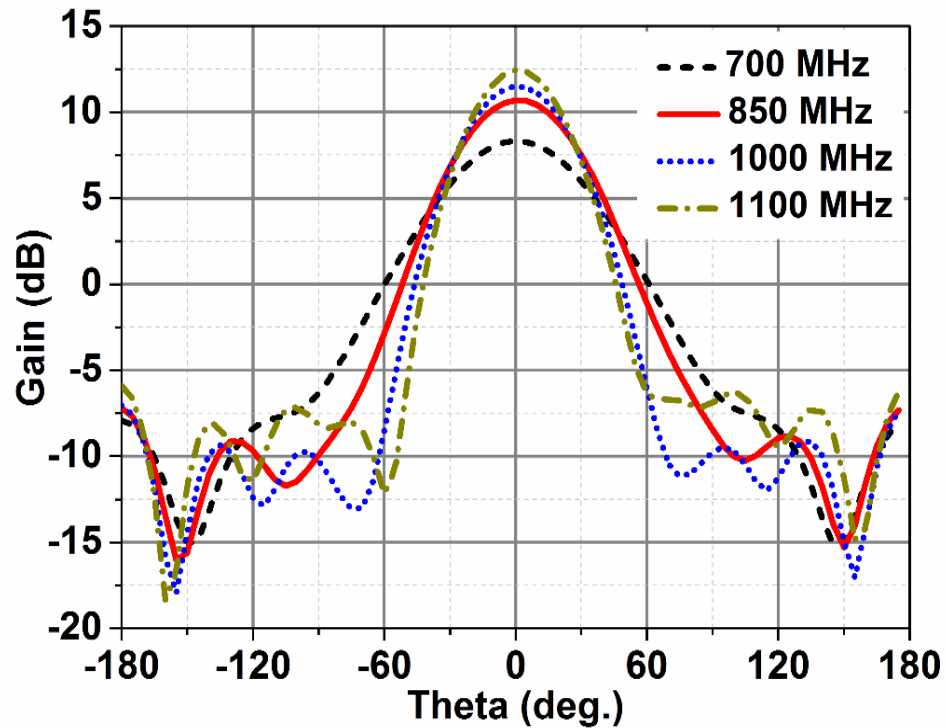
CST Simulation : 47%

IE3D Simulation : 49.5%

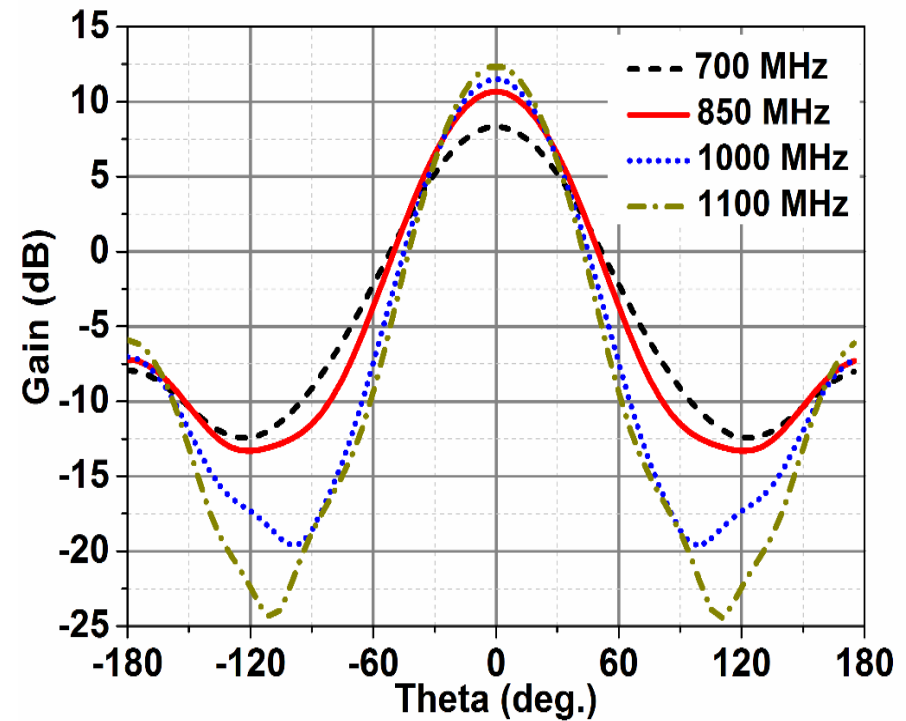
Measured Results : 52%

Simulated Radiation Pattern of Coaxial Feed Pyramidal Horn Antenna

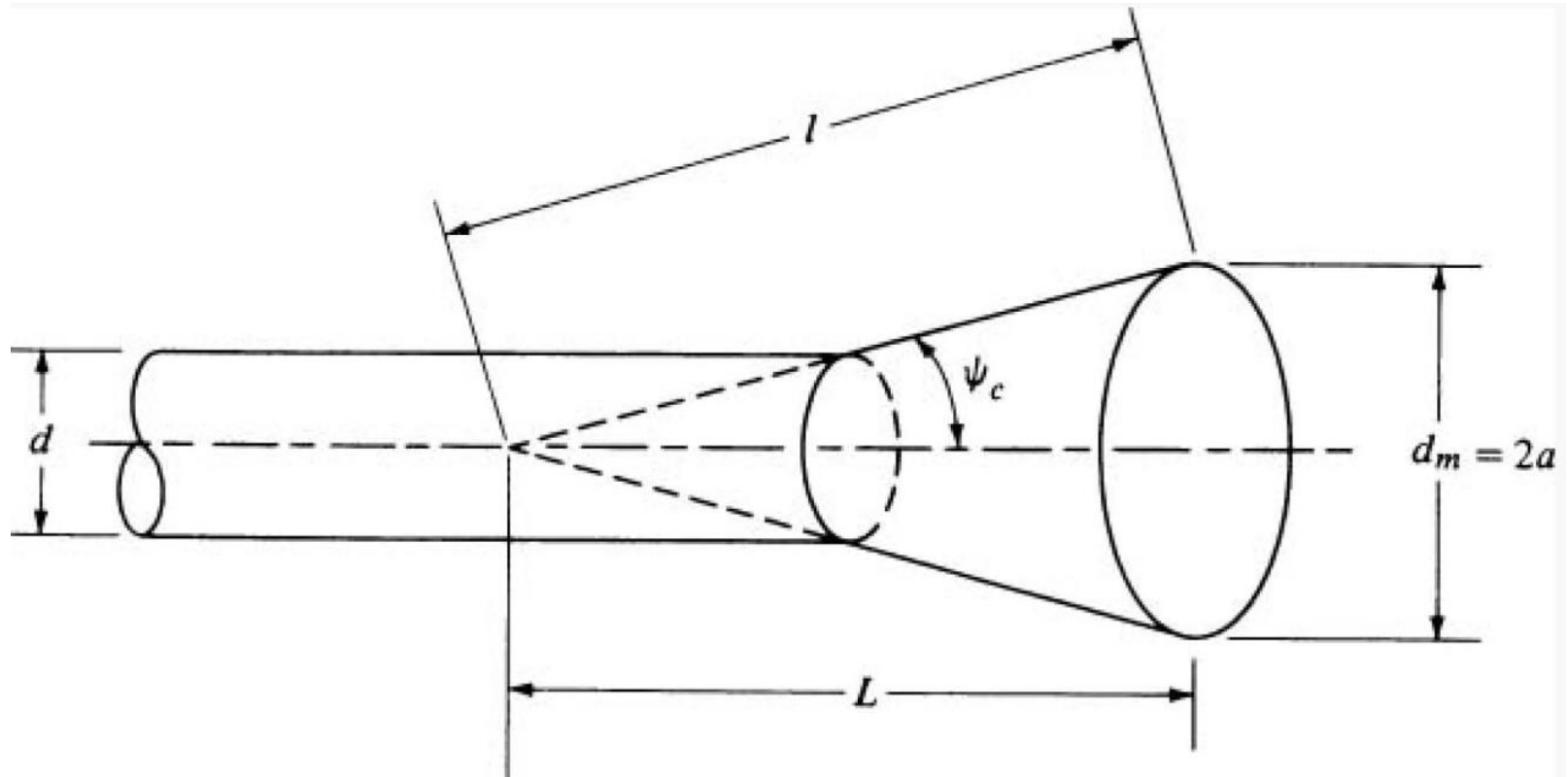
Simulated E-Plane Radiation Pattern



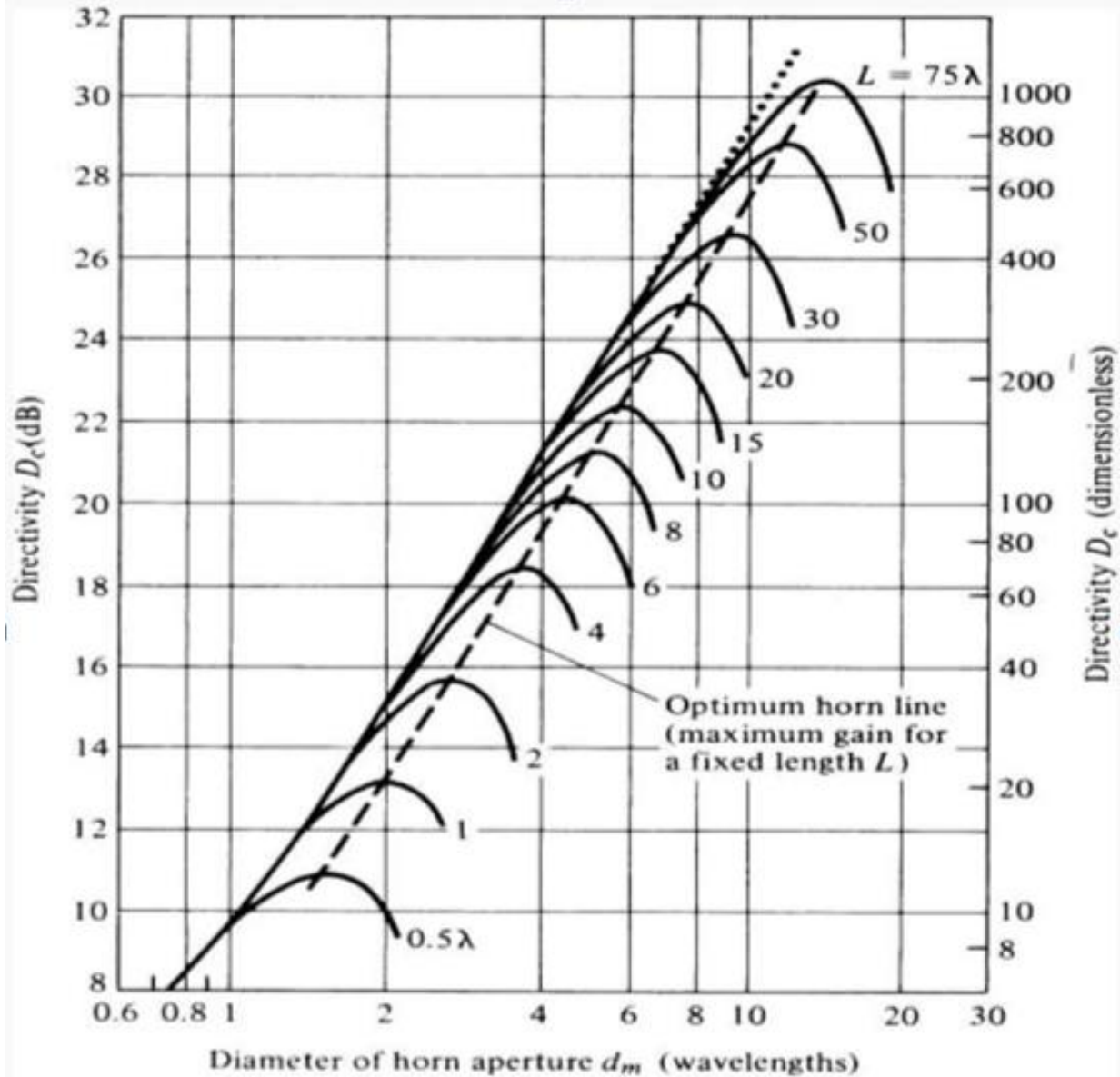
Simulated H-Plane Radiation Pattern



Conical Horn Antenna



Conical Horn: Directivity Curve



Conical Horn Antenna: Directivity

$$s = \frac{d_m^2}{8\lambda l} = \text{maximum phase deviation (in } \lambda \text{)}$$

The gain of conical horn is optimum when:

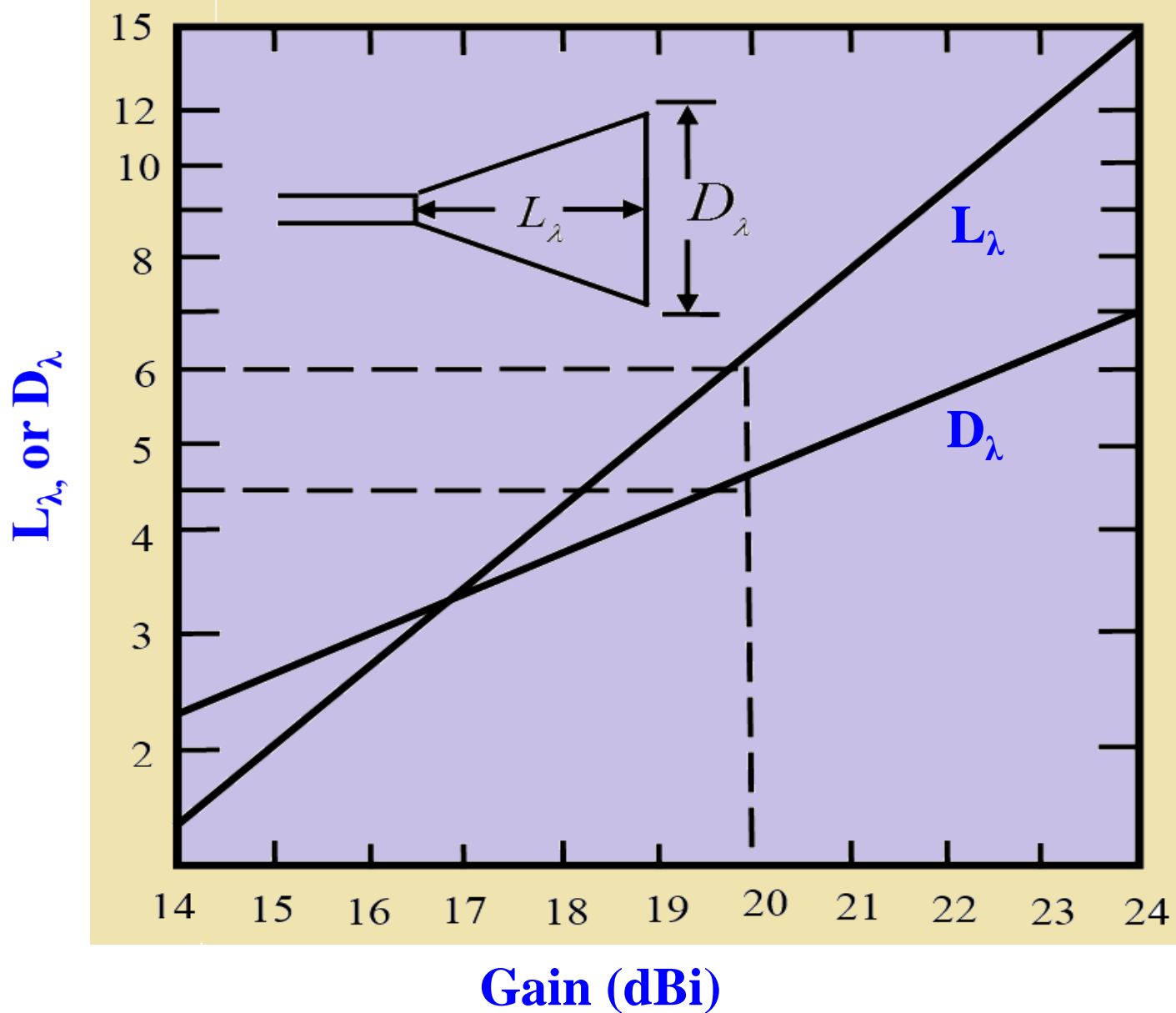
$$d_m \approx \sqrt{3\lambda l}$$

Thus

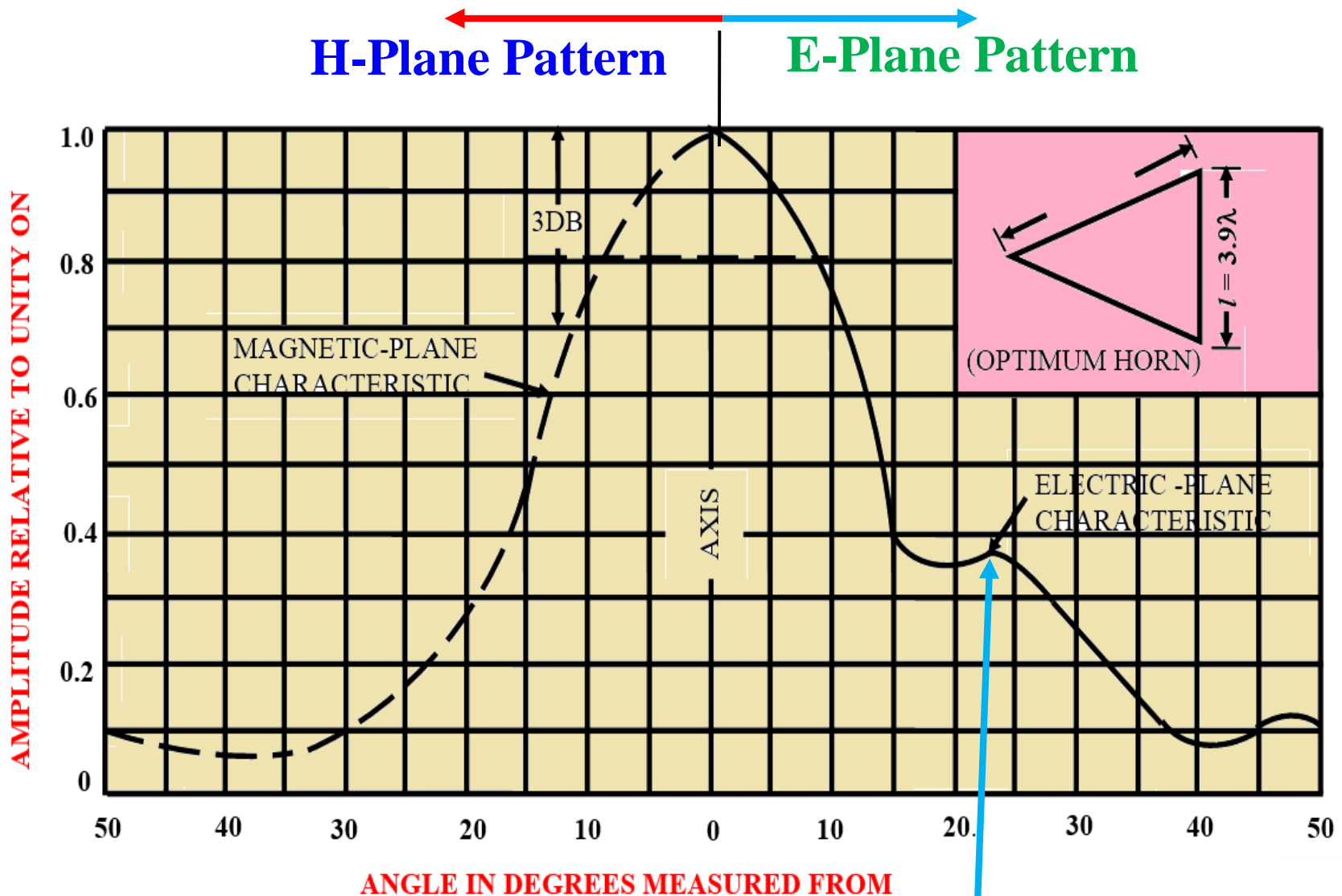
$$s \Big|_{\text{optimum gain}} = \frac{d_m^2}{8l\lambda} \Big|_{d_m=\sqrt{3\lambda l}} = \frac{3\lambda l}{8\lambda l} = \frac{3}{8} \Rightarrow \delta_{max} = 135^\circ$$

**Phase Error too high:
Not Recommended**

Conical Horn Optimum Dimensions vs. Directivity

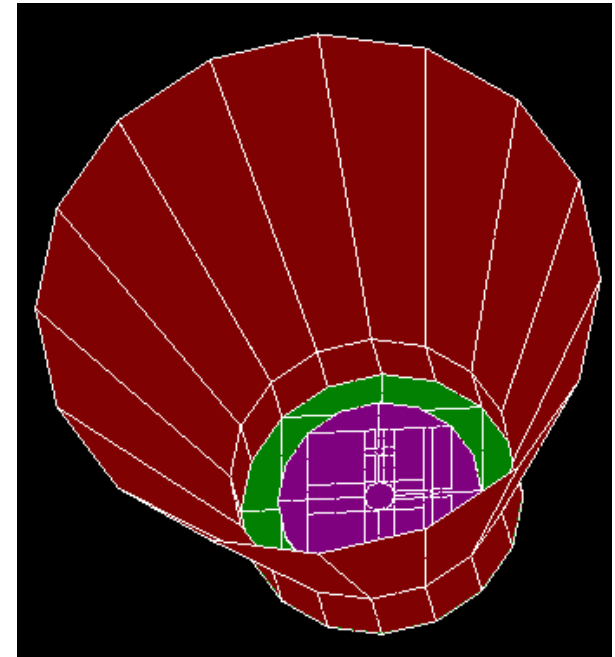
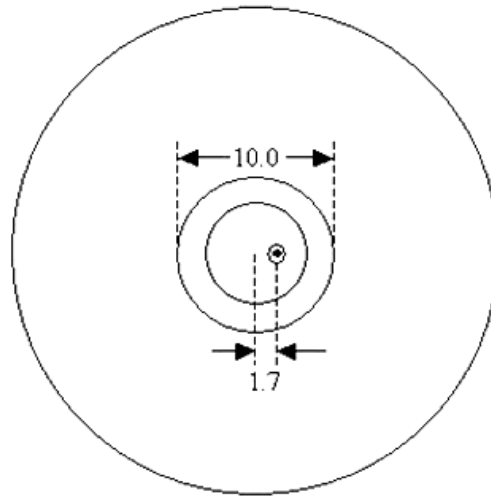
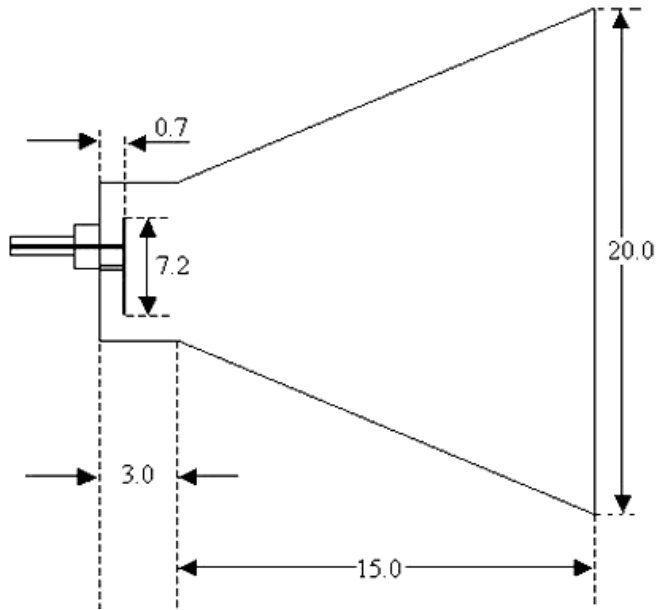


Measured Pattern of Conical Horn



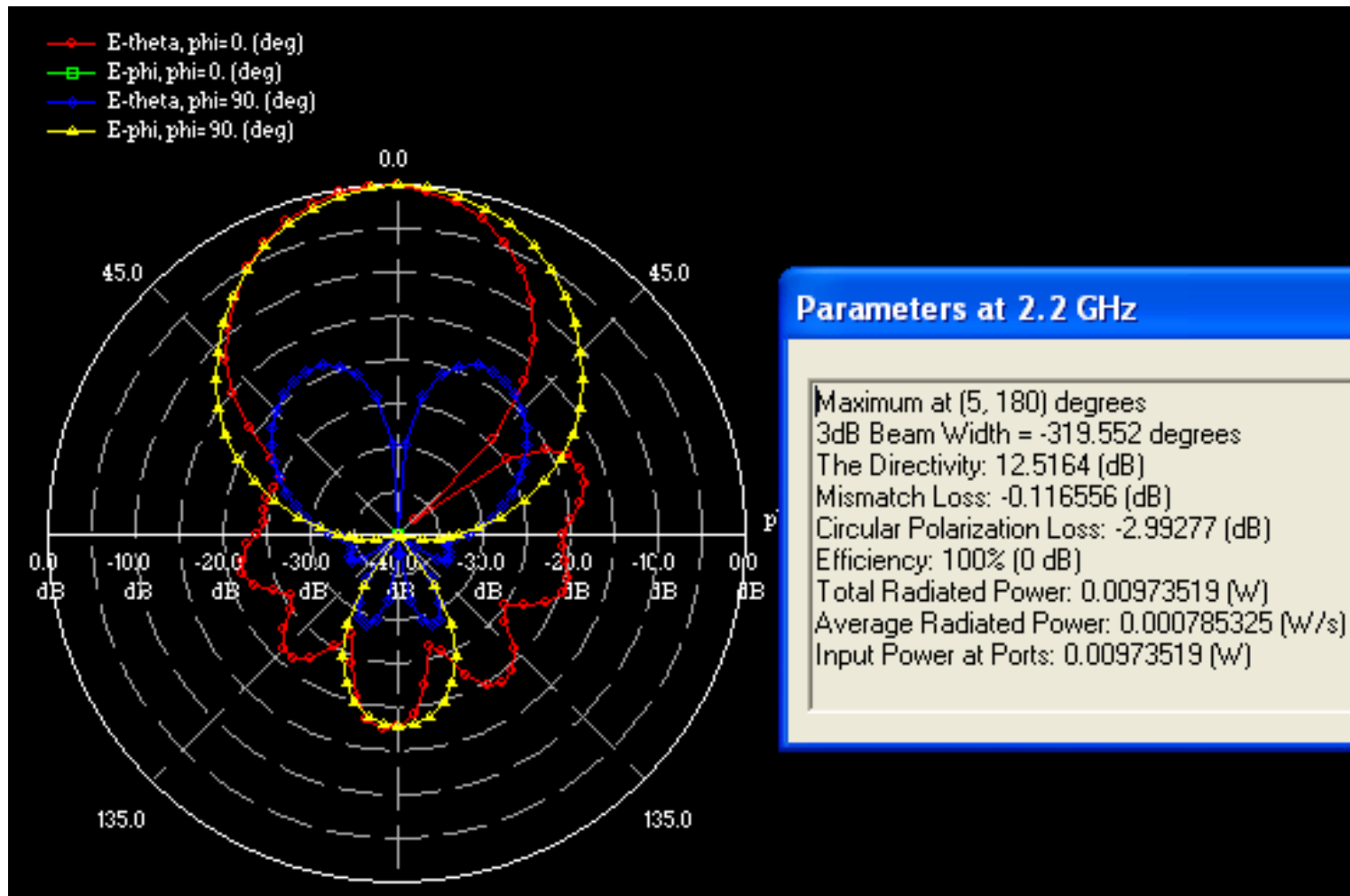
$20 \text{ Log } 0.37 = -8.6 \text{ dB}$. Higher SLL due to large phase error.

MSA Integrated with Conical Horn



Suspended CMSA integrated inside a Conical Horn Antenna. Simulation using IE3D software.

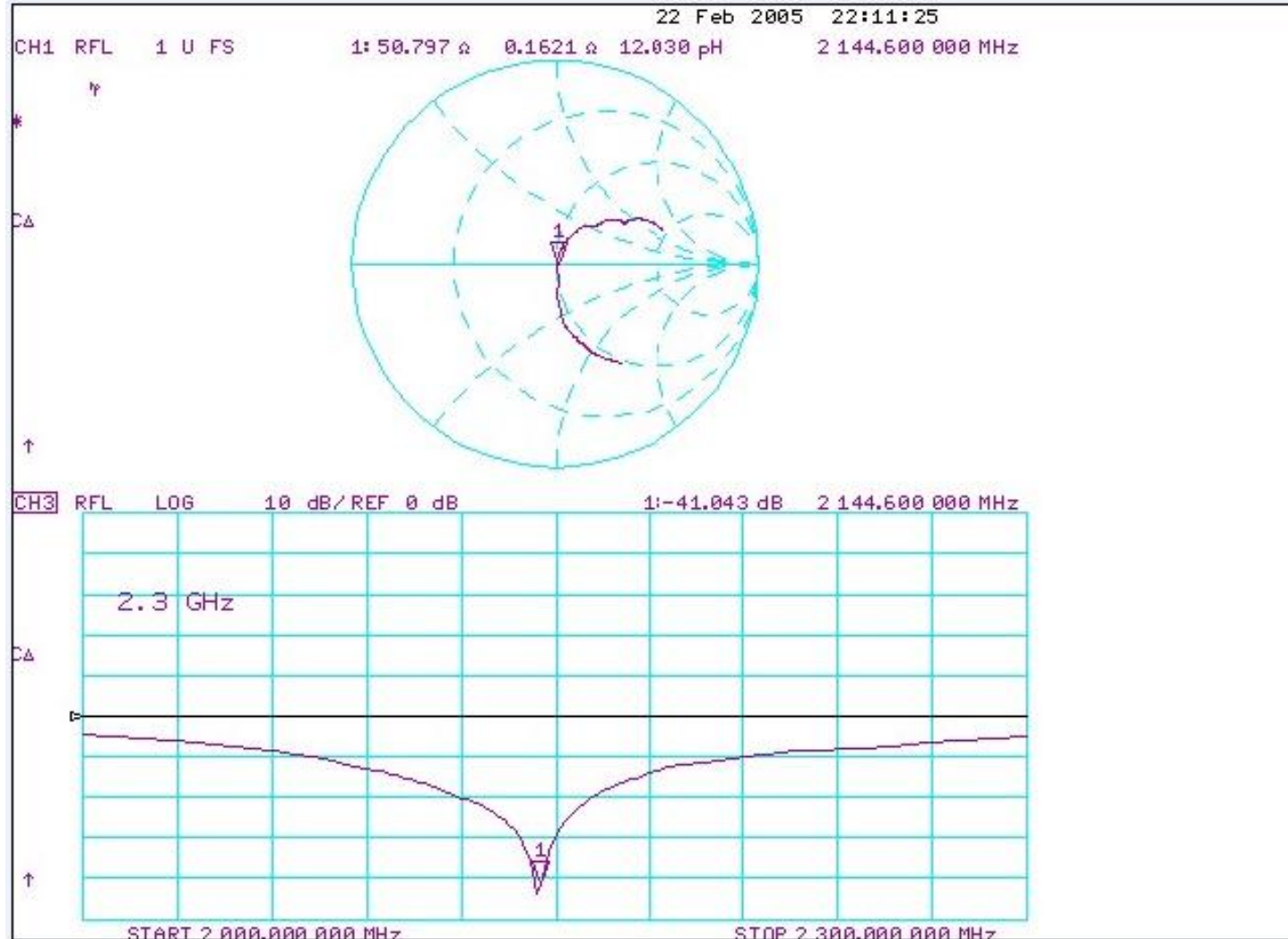
Radiation Pattern of Integrated Conical Horn



Gain of Suspended CMSA = 9 dB

Gain of Integrated Conical Horn Antenna = 12.5 dB

Measured Results of Integrated Conical Horn



Measured BW for $|S_{11}| \leq -10$ dB is from 2070 to 2210 MHz

Horn Antennas

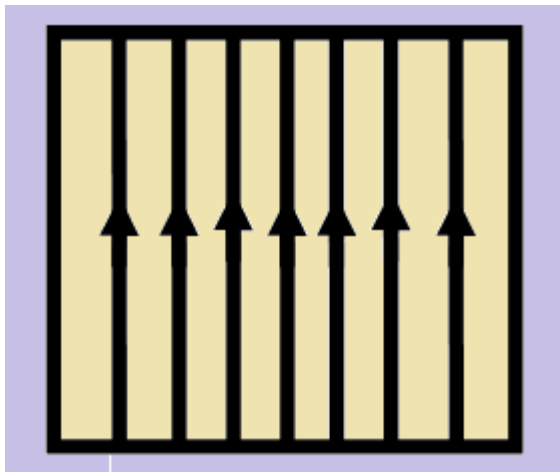
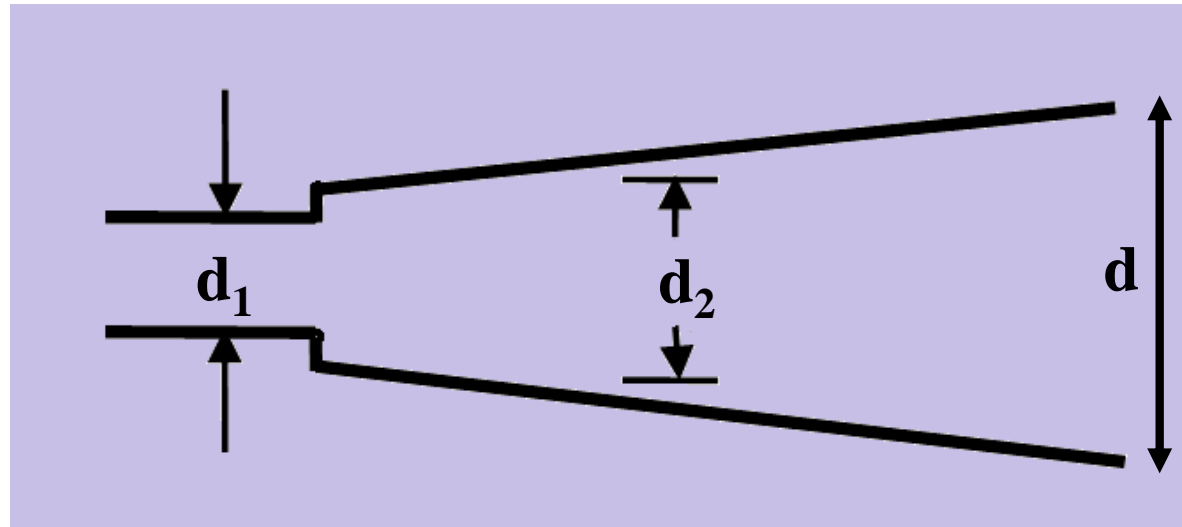
Prof. Girish Kumar

Electrical Engineering Department, IIT Bombay

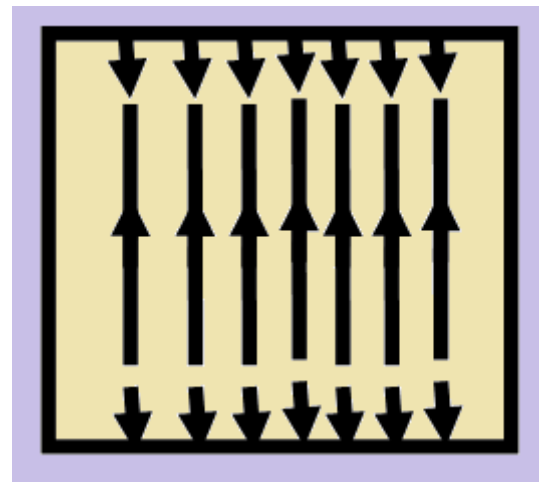
gkumar@ee.iitb.ac.in

(022) 2576 7436

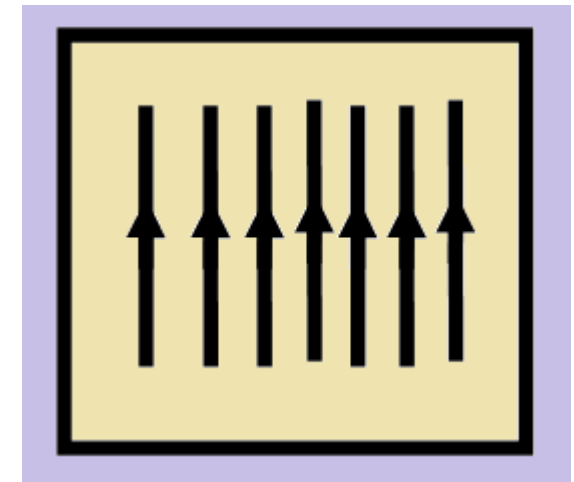
Dual Mode Pyramidal Horn Antenna



TE_{10}

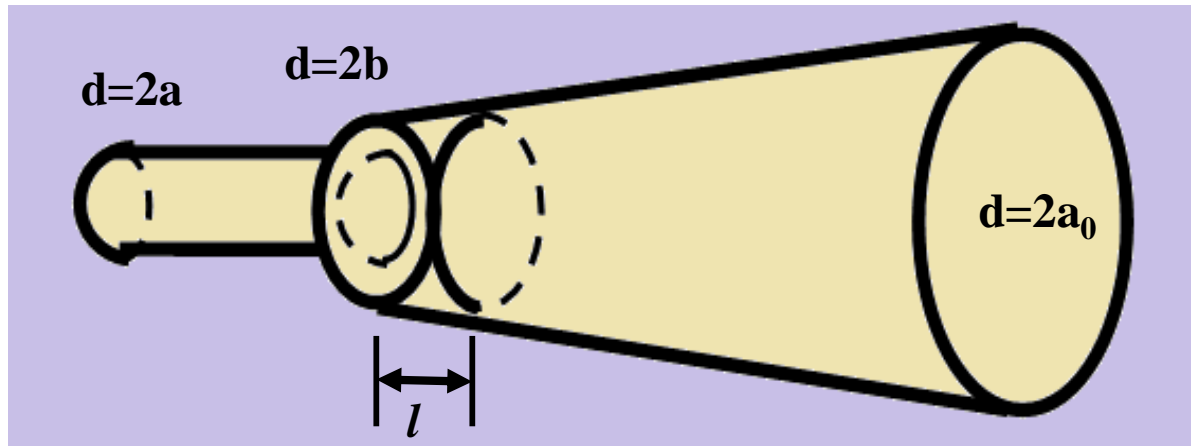


TE_{12}/TM_{12}



Multimode

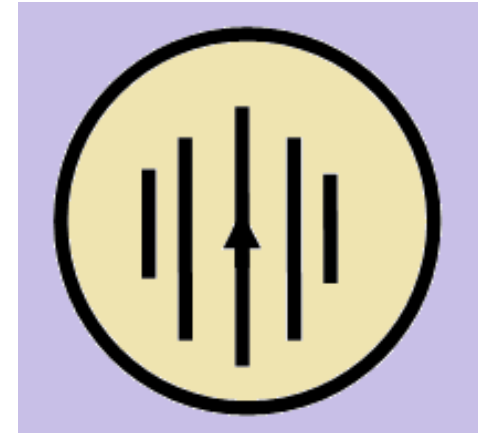
Dual Mode Conical Horn Antenna



TE_{11}

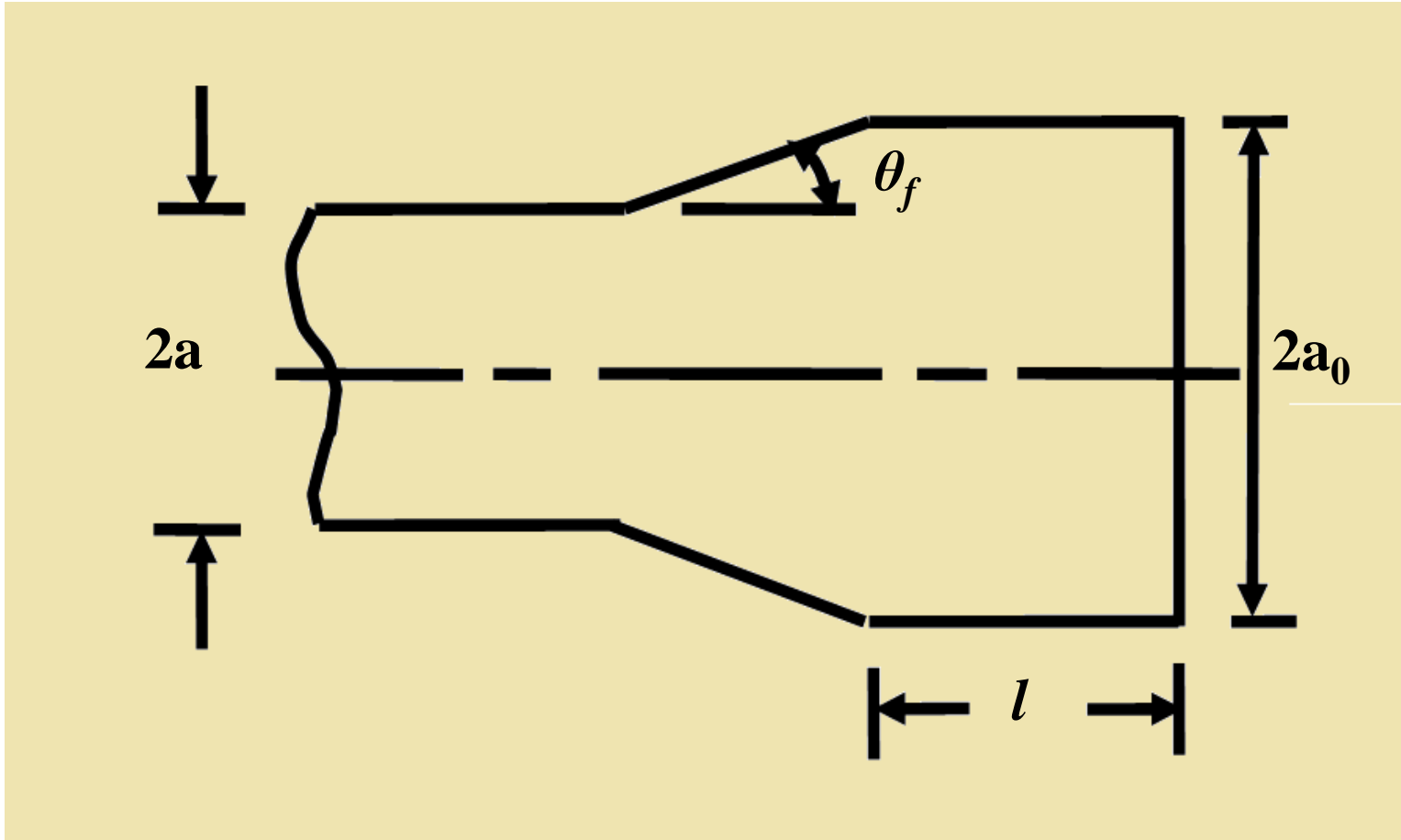


TM_{11}

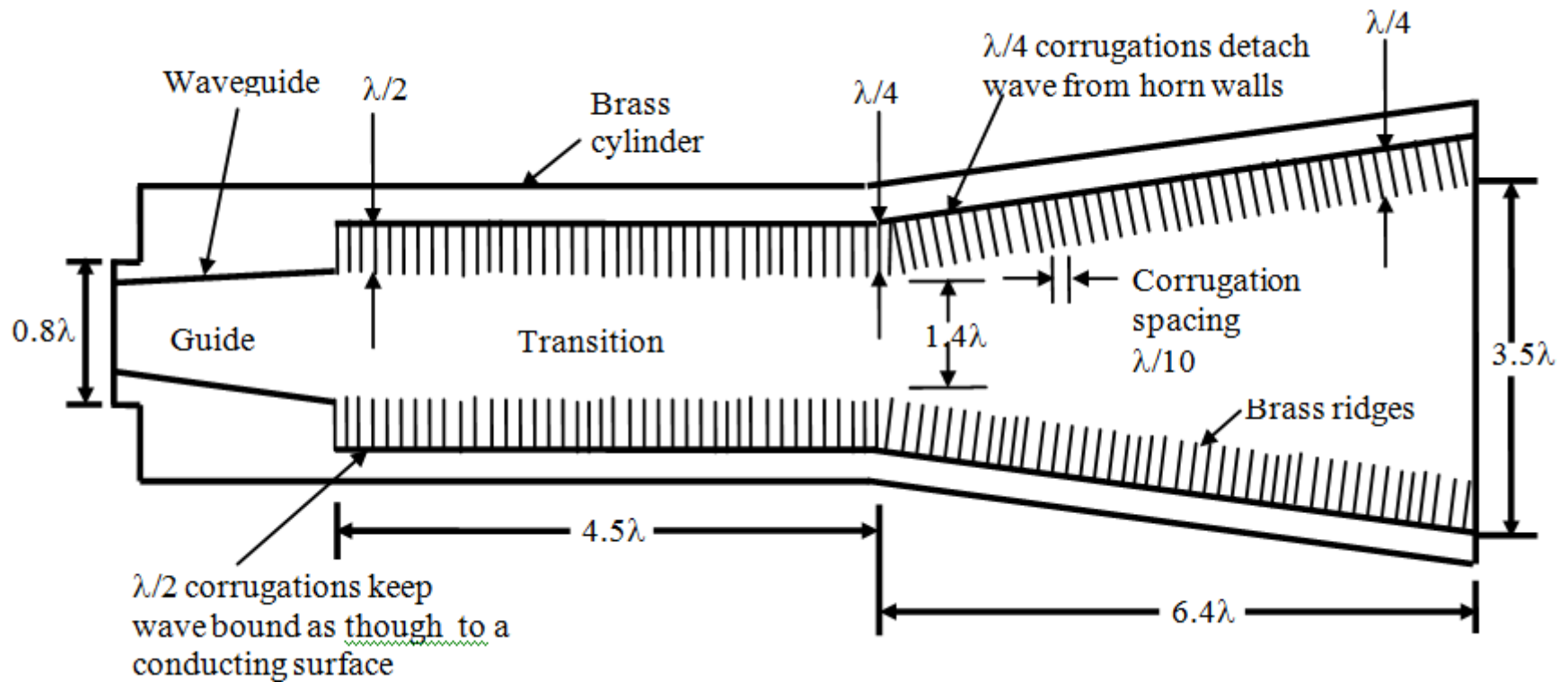


Dual Mode

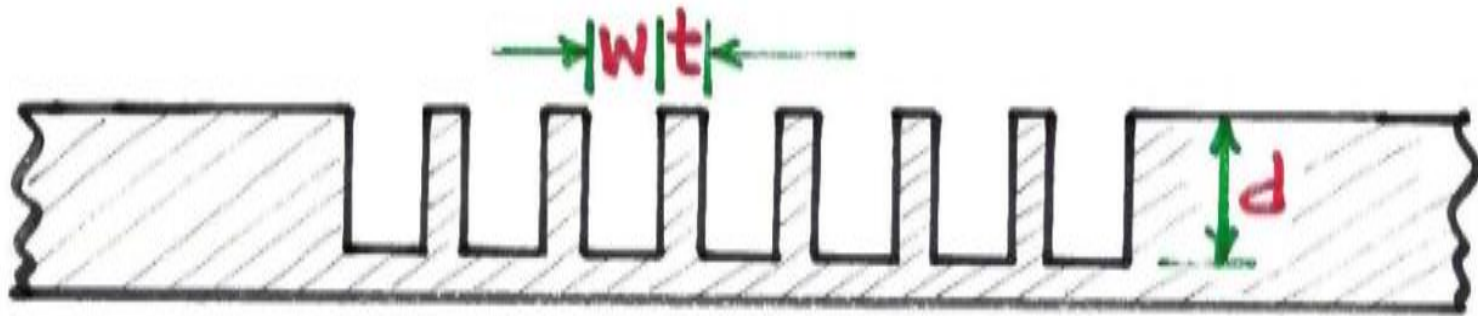
Step-Less Dual Mode Conical Horn



Circular Corrugated Horn Antenna



Corrugated Surface



Typical Values of d , No. of Teeth, w and t :

Depth of the gap (d) = 0.25λ to 0.5λ

No. of Teeth (n) = 4 to 10 per λ

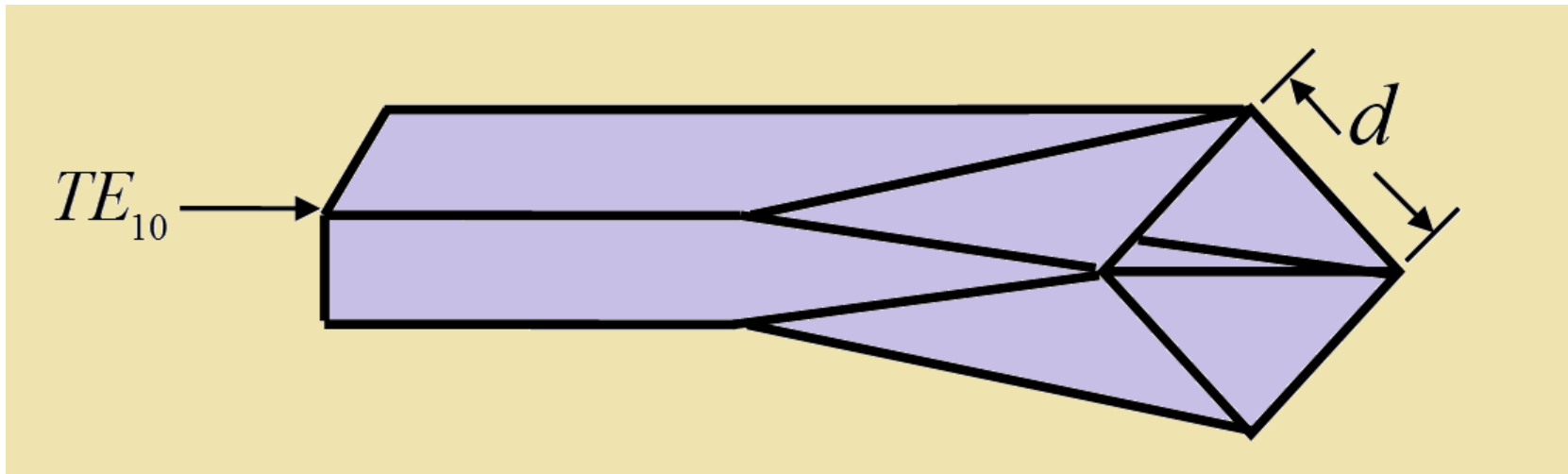
Width of the gap (w) = 0.05λ to 0.2λ

Teeth thickness (t) = 0.02λ to 0.1λ

Corrugated Conical Horn

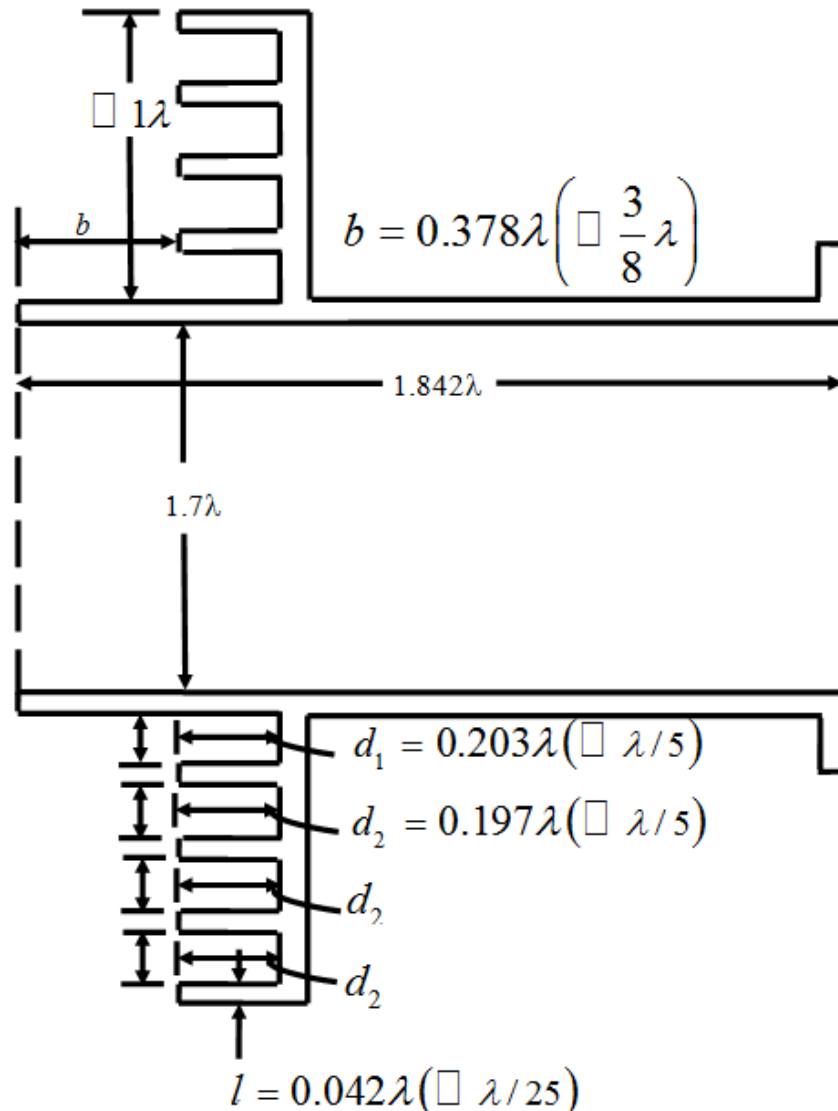


Multimode Horn Antenna



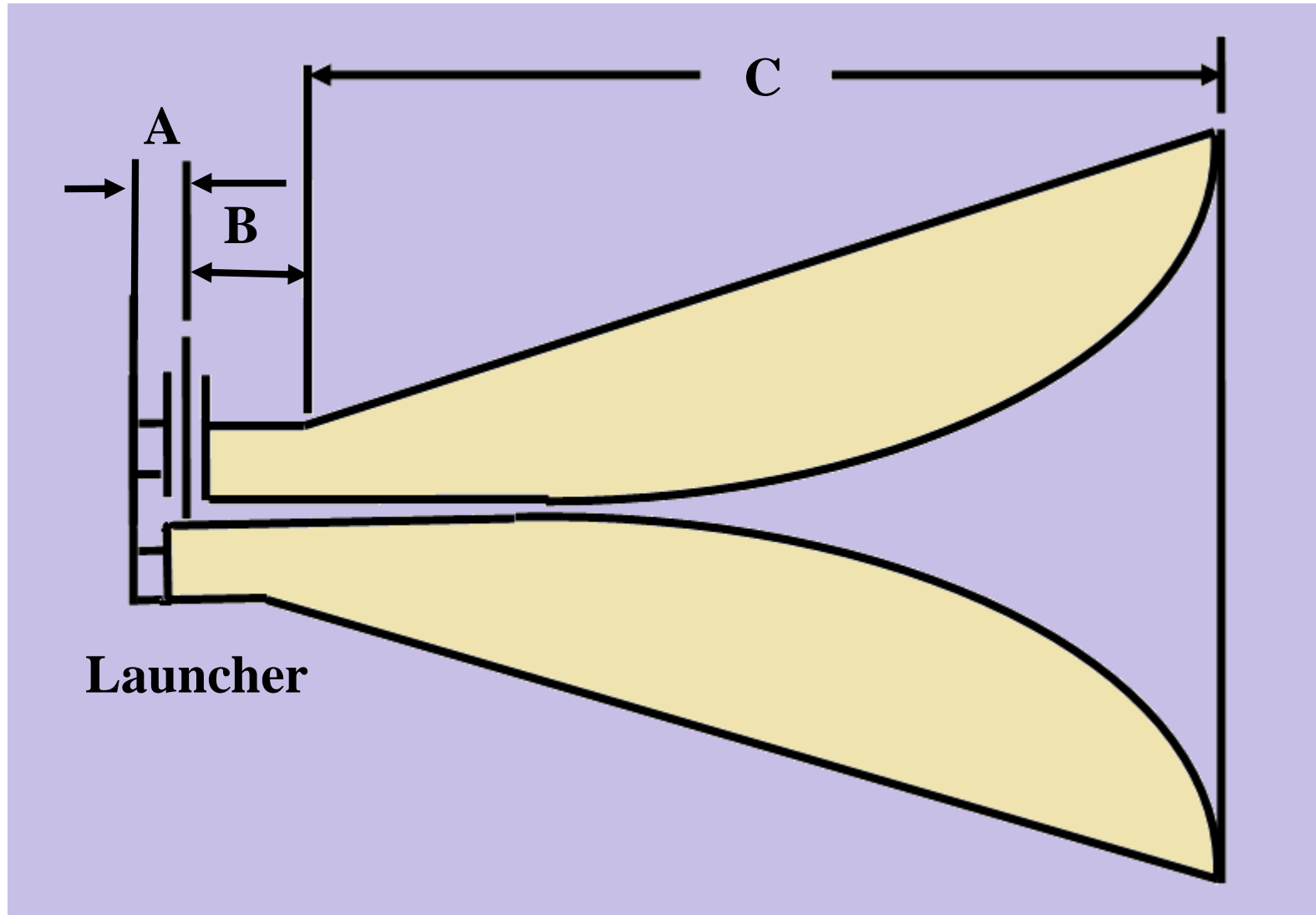
TE_{10} and TE_{01} : Excited with Equal Amplitude and Phase in a square waveguide

Circular Waveguide with Flange

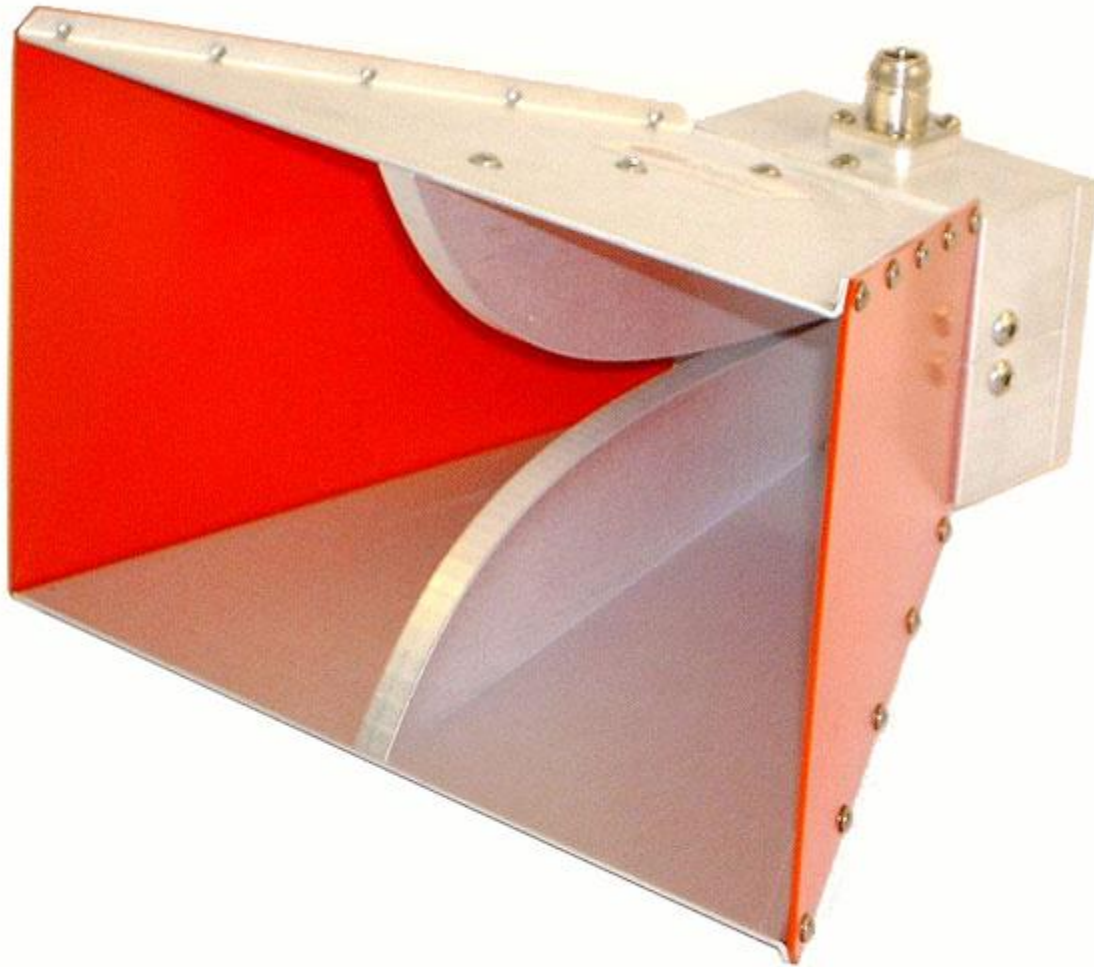


Circular waveguide with flange and 4 chokes for wide-beam-width high-efficiency feed of low F/D parabolic reflectors.

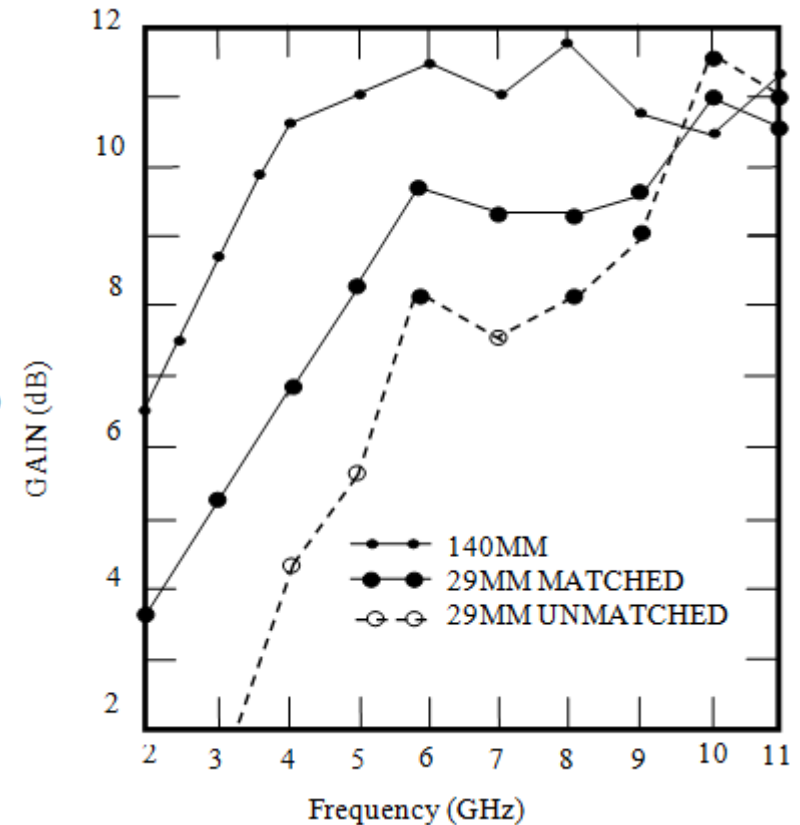
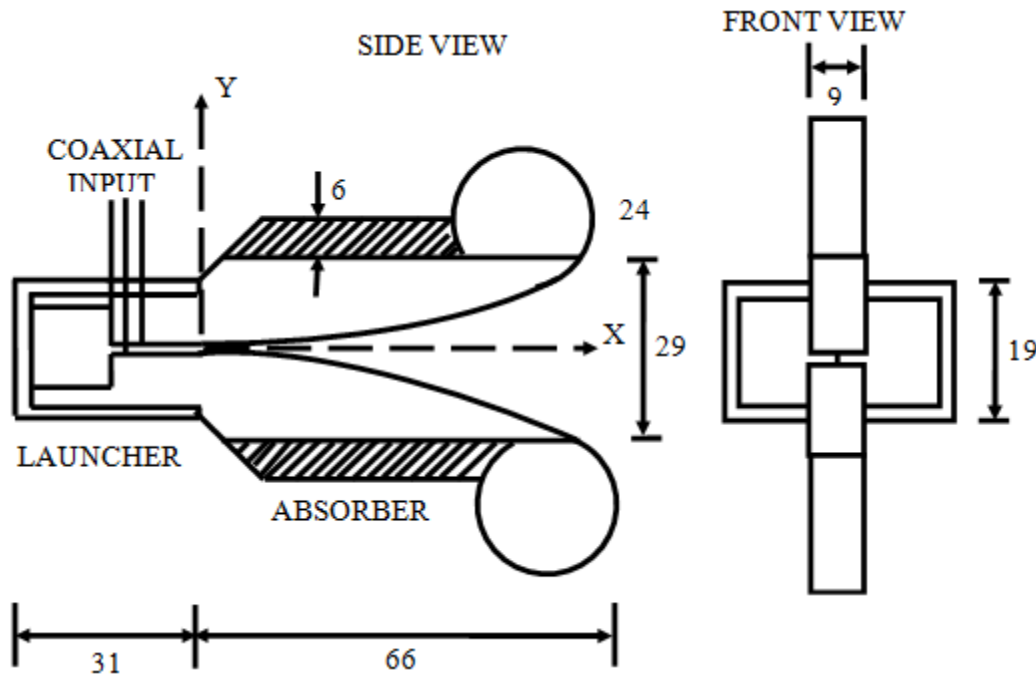
Broadband Exponentially Tapered Horn



Broadband Dual Ridged Horn



Compact Aperture Matched Horn Antenna



Exponential Ridges are used to increase bandwidth.

Aperture matching at the end is done to improve VSWR, reduce scattering and increase the gain.