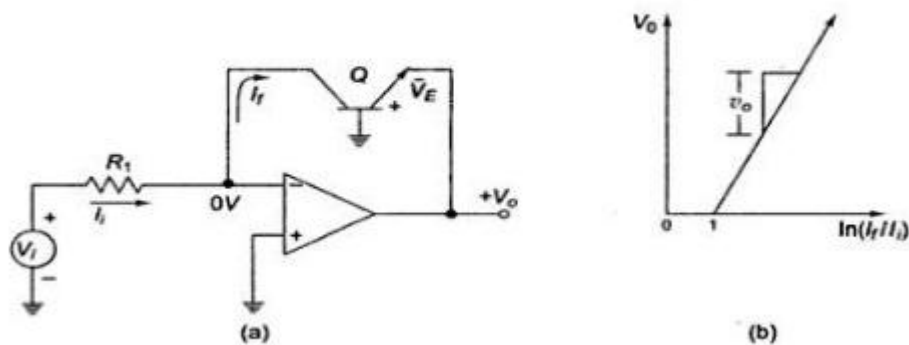


## 2.5 LOG AMPLIFIER



**Figure 2.5.1 Fundamental log-amp circuit and its characteristics**

[source: [https://www.brainkart.com/subject/Linear-Integrated-Circuits\\_220/](https://www.brainkart.com/subject/Linear-Integrated-Circuits_220/)]

There are several applications of log and antilog amplifiers. Antilog computation may require functions such as  $\ln x$ ,  $\log x$  or  $\sin hx$ .

### USES

- Direct dB display on a digital Voltmeter and Spectrum analyzer.
- Log-amp can also be used to compress the dynamic range of a signal.

A grounded base transistor is placed in the feedback path. Figure 2.5.1 is the fundamental log-amplifier circuit and its characteristics. Since the collector is placed in the feedback path. Since the collector is held at virtual ground and the base is also grounded, the transistor's voltage-current relationship becomes that of a diode and is given by,

$$I_E = I_s \left[ e^{\frac{qV_{BE}}{kT}} - 1 \right]$$

and since  $I_c = I_E$  for a grounded base transistor  $I_c = I_s e^{kT}$

$I_s$  = emitter saturation current  $\approx 10^{-13}$  A

$k$  = Boltzmann's constant

$T$  = absolute temperature (in °K)

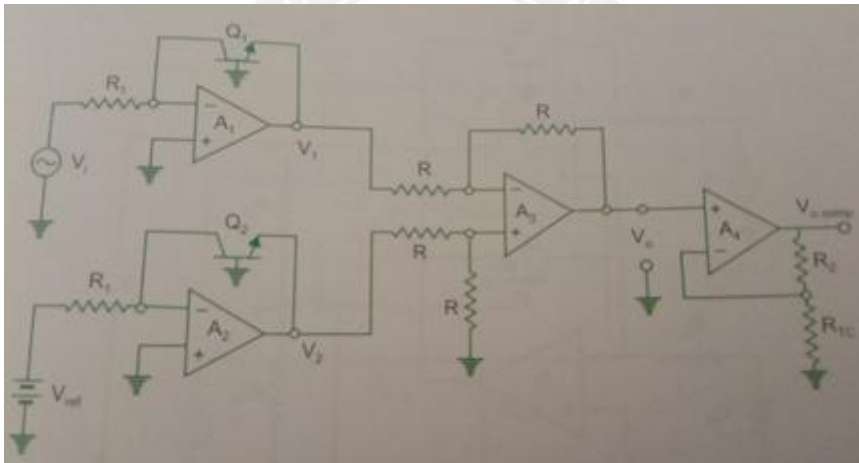
$$V_o = -\frac{kT}{q} \ln\left(\frac{V_i}{R_1 I_S}\right) = -\frac{kT}{q} \ln\left(\frac{V_i}{V_R}\right)$$

where  $V_{ref} = R_1 I_S$

The output voltage is thus proportional to the logarithm of input voltage. Although the circuit gives natural log (ln), one can find log<sub>10</sub>, by proper scaling

$$\text{Log}_{10} X = 0.4343 \ln X$$

The circuit has one problem. The emitter saturation current  $I_S$  varies from transistor to transistor and with temperature. Thus a stable reference voltage  $V_{ref}$  cannot be obtained. This is eliminated by the circuit given below



**Figure 2.5.2 Log-amp with saturation current and temperature compensation**

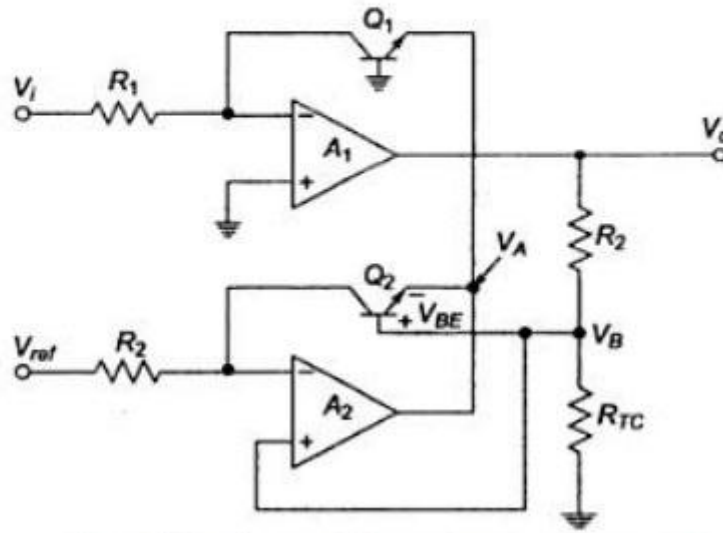
[source: "Linear Integrated Circuits" by D.Roy Choudhry, Shail Bala Jain, Page-177]

Figure 2.5.2 shows the log-amp with saturation current and temperature compensation. The input is applied to one log-amp, while a reference voltage is applied to another log-amp. The two transistors are integrated close together in the same silicon wafer. This provides a close match of saturation currents and ensures good thermal tracking. Assume  $I_{S1} = I_{S2} = I_S$

Thus the reference level is now set with a single external voltage source. Its dependence on device and temperature has been removed. The voltage  $V_o$  is

still dependent upon temperature and is directly proportional to T. This is compensated by the last op-amp stage A4 which provides a non-inverting gain of  $(1+R_2/R_{TC})$ . Temperature compensated output voltage  $V_L$ . Figure 2.5.3 shows the logarithmic amplifier using two op-amps.

$$V_L = \left(1 + \frac{R_2}{R_{TC}}\right) \frac{kT}{q} \ln\left(\frac{V_i}{V_R}\right)$$



**Figure 2.5.3 Logarithmic amplifier using two op-amps.**

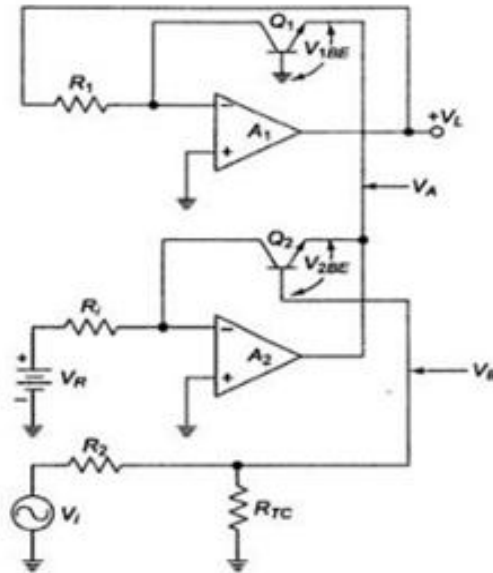
[source: "Linear Integrated Circuits" by D.Roy Choudhry, Shail Bala Jain, Page-178]

Where  $R_{TC}$  is a temperature-sensitive resistance with a positive coefficient of temperature (sensor) so that the slope of the equation becomes constant as the temperature changes.

### ANTILOG AMPLIFIER

A circuit to convert logarithmically encoded signal to real signals. Transistor in inverting input converts input voltage into logarithmically varying currents.

The circuit is shown in figure 2.5.4 below is the antilog amplifier. The input  $V_i$  for the antilog-amp is fed into the temperature compensating voltage divider  $R_2$  and  $R_{TC}$  and then to the base of  $Q_2$ . The output of A2 is fed back to  $R_1$  at the inverting input of op amp A1. The non-inverting inputs are grounded



**Figure 2.5.4 Antilog Amplifier**

[source: "Linear Integrated Circuits" by D.Roy Choudhry, Shail Bala Jain, Page-179]

$$V_{1BE} = \frac{kT}{q} \ln\left[\frac{V_L}{R_1 I_S}\right] \quad \text{and} \quad V_{2BE} = \frac{kT}{q} \ln\left[\frac{V_B}{R_1 I_S}\right] \quad \text{and} \quad V_A = -V_{1BE} \quad \text{and} \quad V_B = \frac{R_{TC}}{R_2 + R_{TC}} V_i$$

$$V_{Q2E} = V_B + V_{2BE} = \frac{R_{TC}}{R_2 + R_{TC}} V_i - \frac{kT}{q} \ln\left[\frac{V_B}{R_1 I_S}\right]$$

$$V_{Q2E} = V_A$$

Therefore,

$$-\frac{kT}{q} \ln\left(\frac{V_L}{R_1 I_S}\right) = \frac{R_{TC}}{R_2 + R_{TC}} V_i + \frac{kT}{q} \ln\left(\frac{V_B}{R_1 I_S}\right)$$

Rearranging, we get

$$\begin{aligned} \frac{R_{TC}}{R_2 + R_{TC}} V_i &= -\frac{kT}{q} \ln\left(\frac{V_L}{R_1 I_S}\right) - \frac{kT}{q} \ln\left(\frac{V_B}{R_1 I_S}\right) \\ &= -\frac{kT}{q} \ln\left(\frac{V_L}{V_B}\right) \end{aligned}$$

We know that  $\log_{10} x = 0.4343 \ln x$ .

$$\text{Therefore,} \quad -0.4343 \left(\frac{q}{kT}\right) \left(\frac{R_{TC}}{R_2 + R_{TC}}\right) V_i = 0.4343 \ln\left(\frac{V_L}{V_B}\right)$$

$$-0.4343 \left(\frac{q}{kT}\right) \left(\frac{R_{TC}}{R_2 + R_{TC}}\right) V_i = \log_{10} \left(\frac{V_L}{V_B}\right)$$

$$-KV_i = \log\left(\frac{V_L}{V_B}\right)$$

$$K = 0.4343 \left(\frac{q}{kT}\right) \left(\frac{R_{TC}}{R_2 + R_{TC}}\right)$$

$$V_L = V_B 10^{-KV_i}$$

The output  $V_o$  of the antilog- amp is fed back to the inverting input of  $A_1$  through the resistor  $R_1$ . Hence an increase of input by one volt causes the output to decrease by a decade.

