

ELECTRONIC ORGAN DESIGN

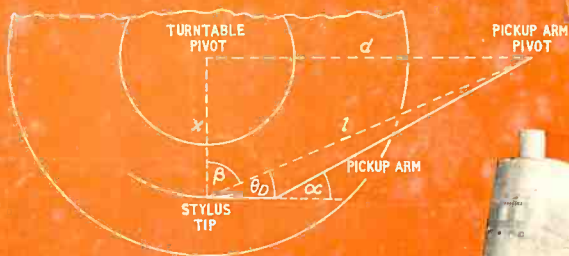
MAY 1966

Three shillings

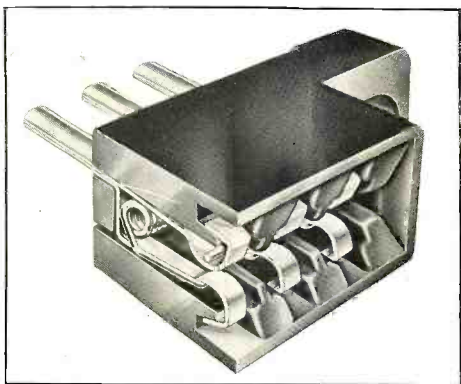
Wireless World

ELECTRONICS • TELEVISION • RADIO • AUDIO

PICKUP TRACKING
ELECTRONICS



FERRANTI



EDGE CONNECTORS

for printed
circuit boards

- Low rate floating spring.
- Controlled contact pressure.
- Unsurpassed reliability.
- Polarisng achieved without removal of contacts
- Hard gold plated to 0.0002 inch minimum.
- ✳ Gold Flashed and Silver Plated contacts supplied on request.
- Constant contact performance with different board thicknesses.

At present available in 8, 16, 24, 32, & 40 Pole sizes

FERRANTI *First into the Future*



NOW!



Avometer

Model 8 Mk.3

Thanks to greatly increased production at our new Dover factory the Model 8 Mk. 3 Avometer—renowned for its exceptional performances and reliability—is now available for prompt delivery. Why wait? There's no better moment than *now* to get in touch with one of the addresses below.

Where to get your Avometer

(or any other Avo instrument)

FARNELL INSTRUMENTS LTD.
SANDBECK WAY, WETHERBY, YORKS.
 Telephone: Wetherby 2691 2/3/4

Anglesey, Caernarvonshire, Cheshire, County Durham, Cumberland, Denbighshire, Derbyshire, Flintshire, Lancashire, Lincolnshire, Staffordshire, Westmorland, Yorkshire.

WIRELESS ELECTRIC LTD.
ST. THOMAS STREET, BRISTOL 1.
 Telephone: 9/8434

Cornwall, Devon, Dorset, Gloucestershire, Hampshire, Hertfordshire, Somerset, South Wales, Wiltshire, Worcestershire.

ELESCO ELECTRONICS LTD.
1103 ARGYLE STREET, GLASGOW, C.3.
 Telephone: CENtral 5956

All parts of Scotland.

For Avo stockists in all other counties please contact the Home Sales Department at the address below.

For details of Avo representation OVERSEAS contact the Export Sales Department at the address below.



AVO LTD

AVOCET HOUSE - 92-96 VAUXHALL BRIDGE ROAD - LONDON S.W.1 Telephone: VICtoria 3404

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WW-004 FOR FURTHER DETAILS.

M
 YOUNG & BECKETT

A

the experts'
choice

Stentorian



F. C. IUDD
*Technical Editor
Amateur Tape Recording and
Hi-Fi Magazine**



JOHN GILBERT
*Technical Editor
"Music Trades
Review"*

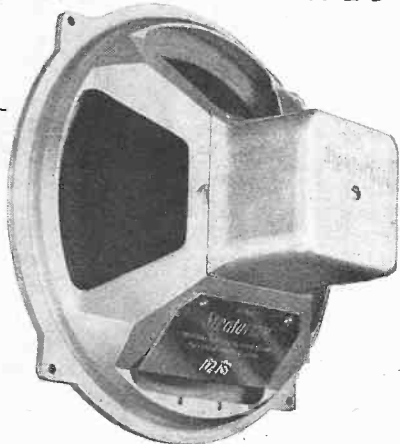
See and hear us
at the

**AUDIO
FAIR
STAND 46**

**Demonstration
Room 204**



DONALD ALDOUS
*Technical Editor
"Audio and
Record Reviews"*



MODEL H.F. 1016 'MAJOR'

This unit makes use of the high flux density available in the magnet system of the previous H.F.1016 unit. A curved diaphragm is used with a rigid centre-section coupled to the voice coil. The rigid coupling and the design of the cone termination give a balanced response over the whole audio range. The unit is specially suitable for use in the smaller type of enclosure having a volume of approximately $1\frac{1}{2}$ cubic feet.

Specification:

Chassis—die cast aluminium; Cone—graded pulp cambric surround;
Cone dia.—10in.; Pole dia.—1in.; Flux density—16,000 gauss; Total
flux—64,000 maxwells; Impedance—15 ohms.

Price: £10 . 7 . 6 (inc. tax)

Other Stentorian Speakers ▶

CROSSOVER
NETWORKS:
CX3,000 £11/13/3
CX1,500 £2/2/-
CX500 £1/8/9

Type	Flux Density	Price	Type	Flux Density	Price
10" H.F.1012*	12,000 gauss	£5 1 3	T359 tweeter	9,000 gauss	£1 15 6
8" H.F.816*	16,000 gauss	£6 18 6	T816	16,000 gauss	£6 11 3
8" H.F.812*	12,000 gauss	£4 4 3	T12 tweeter	16,000 gauss	£15 5 6
8" H.F.810	10,000 gauss	£3 2 9	T10 tweeter	14,000 gauss	£5 2 0

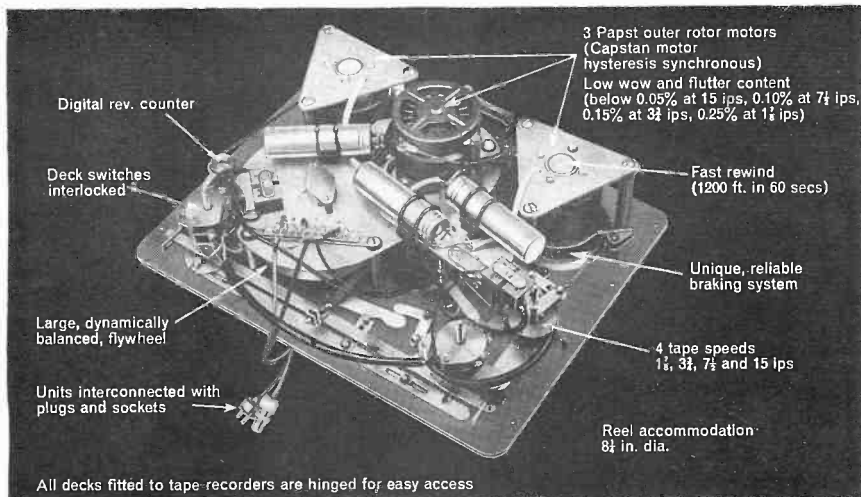
*These three Speakers incorporate a universal impedance speech coil.



WHITELEY ELECTRICAL RADIO CO. LTD.

MANSFIELD · NOTTS · ENGLAND · Telephone: MANSFIELD 1762-8

London Office: 109 Kingsway, London W.C.2.



We have been accused . . .

of hiding our light under a bushel, the light being our superb 4 speed tape deck, so we have taken the hint and given you a large illustration pointing out some of its principle features. This deck is used on all Brenell models and there are versions available to take 10 1/2" NAB reels. Also we supply tape decks and matching amplifiers separately for building into your own equipment cabinet. Write for details of the Brenell range.



**MARK 5
SERIES 3**
**MONO—HALF TRACK—TWO HEADS—
MAGIC EYE**
(Available with recording level meter at extra cost.)

High quality amplifier with power output of 2½ watts r.m.s. and a frequency response of 40–20,000 c/s—can be used independently of tape recorder—narrow gapped record/playback head for extended frequency response—double-gapped ferrite erase head to minimise erase noise—headphone monitoring.



**MARK 5
TYPE M SERIES 3**
**MONO—HALF TRACK—THREE HEADS—
RECORDING LEVEL METER**
Separate record and playback heads—separate record and playback amplifiers—amplifier frequency response 25–28,000 c/s ±3dB—power output 2 watts r.m.s.—separate bass and treble controls—mixing of input signals—speaker monitoring whilst recording.

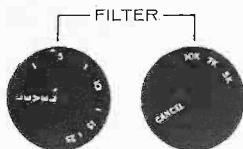


STB2
**MONO/STEREO—HALF TRACK (Record/
playback)—QUARTER TRACK (playback)
FOUR HEADS—TWO EDGEWISE METERS**
Designed for use with high fidelity stereo installations—adjustable attenuators on all input channels to ensure perfect matching with all auxiliary equipment—dual concentric recording level and playback level controls—cathodic follower output—four channel mixing on mono programme sources—twin recording and twin playback pre-amplifiers—comparison of original and recorded signal—adjustable bias level—recording facilities for 1/2 and 2/2 track—playback facilities for 1/2, 2/2, 1/4 and 2/4 tracks—sound-on-sound facilities—two edgewise meters for recording level, tape output level and bias level.
Optional extra: stereo power amplifiers and monitoring speakers.

Brenell

BRENELL ENGINEERING CO. LTD.
231/5 Liverpool Road, London, N.1.
Telephone: NORTH 8271 (5 lines)

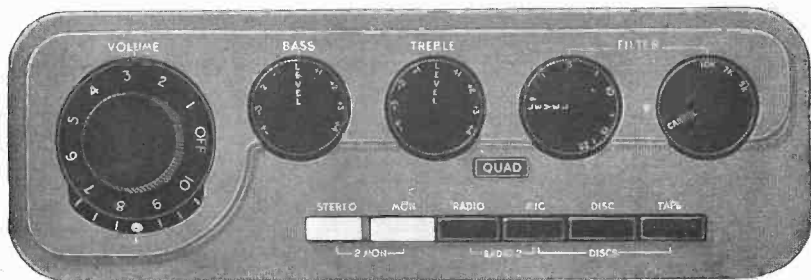
WW—006 FOR FURTHER DETAILS.



More of the music and less of the distortion!

The processing and production of discs produces a waveform in which distortion rises extremely rapidly as the upper frequency limit is approached. To obtain the best quality under such conditions the very high harmonics must be attenuated at a rate which is a function of the rate of rise of distortion.

The QUAD filters provide both switched adjustment for frequency and continuously variable adjustment for rate of attenuation. The degree of attenuation need therefore never be greater than necessary to clean up the programme but is always adequate even for very bad cases.



For the closest approach to the original sound

Our slogan for fifteen years and our design objective for twice that long. Ask your dealer for details of the QUAD range of high fidelity units or write direct to

**Ref. W.W. Acoustical Manufacturing Co. Ltd.,
Huntingdon. Huntingdon 361.**



WW-007 FOR FURTHER DETAILS.

The hard-to-find man is the man with advanced technical knowledge

In the field of Electronics today the greatest demand is for men to fill the positions of skilled technicians and support engineers. These are the 'hard-to-find' men for whom the way ahead to rewarding and interesting jobs lies wide open.

CREI HOME STUDY COURSES offer advanced technical education to the man who realises that technical knowledge must be current and up-to-date if progress is to be made in a world in which new ideas, new techniques and new applications develop almost overnight.

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City & Guilds of London Institute: Subject 49 and
Supplementary Studies Subject 300.

For further information write to:

**C.R.E.I. (London) (Dept. WW67)
WALPOLE HOUSE, 173/176 SLOANE STREET, LONDON, S.W.1
Telephone: BELgravia 8662**

(INTERNATIONAL DIVISION OF CAPITOL RADIO ENGINEERING INSTITUTE, WASHINGTON, D. C.)

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JUST LIKE THE OTHER 800 MILLION- UNIQUE

Erie have made approximately 800 million of the Type 9 insulated carbon composition resistor since they were first put into production as far back as the early thirties. Every one is identical, yet every one is unique.

The care and attention to detail in their manufacture, and the rigorous inspection checks at every stage, ensure that not only does every component operate efficiently, but is in itself an example of Erie's unique manufacturing skill.

The same care and attention that goes

into quality control applies throughout all the company's activities; your orders and enquiries for instance. Every order, whether it be for a million, or for a strictly limited number for a highly specialist purpose, receives the same quiet efficient attention.

Every enquiry, too, no matter how apparently simple or difficult receives the same enthusiastic attention from our technical and sales staff.

Erie take a pride in performance



Write for catalogue to:

ERIE RESISTOR LIMITED,

Great Yarmouth, Norfolk, England.

Telephone: Great Yarmouth 4911.

Cables: Resistor Great Yarmouth Telex: 97421

TAKE SIX

to measure a megacycle



STOP PRESS

- * Variable sensitivity control on A.C. input.
- * Variable display timer (0.5-5 secs).
- * Period measurement over 1, 10 & 100 cycles
- * Provision for external ref.
- * Binary coded decimal output (optional extra).

Six big digits, clear and unambiguous, to measure frequency and time up to one megacycle and down to one microsecond—clearly, without flicker.

Six into one only just goes

Not much spare space on the panel of this instrument—the digits are big and the unit is small, compact, portable. Makes for convenience and easy reading, with an adjustable support that holds firm at any convenient angle.

Smaller and smaller!

Design of the Venner TSA 3436 Mark 2 is based on solid experience of applying transistorised digital techniques to measuring instruments. Small as it is, the TSA 3436 Mark 2 will give precise readings up to a megacycle and down to a microsecond.

ONLY £215 (Price applicable to U.K. only)

Laboratories and technical colleges welcome the compactness, strength and simplicity-in-use of the TSA 3436 Mark 2. In industry, its dependability, and the clarity of its display, make it uniquely suitable for inspection and quality control applications.

Full technical details will be sent on request.

VENNER TSA 3436 MARK 2 TIME/FREQUENCY METER

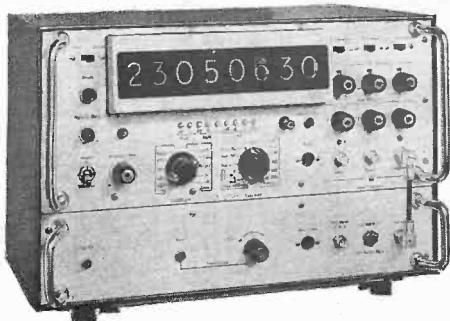
VENNER ELECTRONICS LIMITED

KINGSTON BY-PASS · NEW MALDEN · SURREY, MALDEN 2442

Overseas subsidiaries

VENNER A. G. BURGERNZELWEG 16-3000 BERNE-SWITZERLAND · VENNER N. V. HELMSTRAATS DEIN HAAG · HOLLAND
WW-010 FOR FURTHER DETAILS.

RACAL Digital counters combine versatility with economy



SA.550 N.A.T.O. Codification 6625-99-971-8519

Racal SA.550 100 Mc/s Digital Frequency Meter

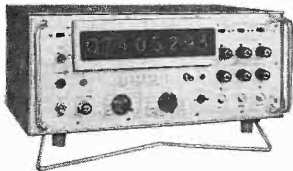
Direct Frequency Measurements from signal levels as low as 100 mV are possible without tuning or interpolation. Measurement capability can be extended by the use of the active probe unit type SA.544.

- All Solid-State Design
- 8 Digit Inline Display
- 0-55°C. operating Ambient
- D.C. Logic Switching
- Digital Printout Facilities
- Internal or External Standard

The SA.540 Universal Counter Timer illustrated below is for time, period and frequency measurement up to 11 Mc/s.

SA.550 U.K. Price £1,100

SA.540 U.K. Price £795



SA.540

RACAL

Write for fuller details to:
Racal Instruments Ltd.,
Dukes Ride, Crowthorne, Berks
Tel: Crowthorne 2272/3 or
Bracknell 941 Telex

LADDER

FILTERS



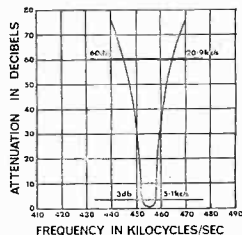
ADVANTAGES

Fantastic space saving
Excellent stability with time and temperature
Single or multiple resonant units

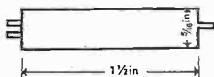
TYPICAL SPECIFICATION

Centre Frequency	300 to 600 kc/s \pm 2 kc/s
Bandwidth	2 to 45 kc/s at 6 db
Insertion Loss	1 to 15 db (dependent on B/W)
Impedance	Either 1200 or 1500 ohms in and out, dependent on B/W
Case Size	8 mm diameter 38 mm long
Operating Temperature Range	-40°C to +85°C
Shape Factor 60/5 db	1.3 : 1 to 2.6 : 1 Dependent on B/W

Curve below shows response of our ladder filter type TL 10 D 16A. Note the outstanding shape factor

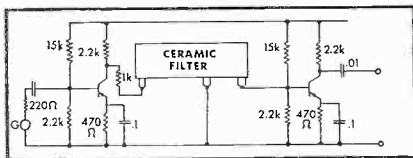


Keep your eye on Brush



SIZE

This illustrates the typical dimensions for a ladder filter. The case is a plated brass cylinder with glass end seals and pigtail leads. A 1 1/2 inch length will provide sufficient volume to give 60 db stop band rejection.



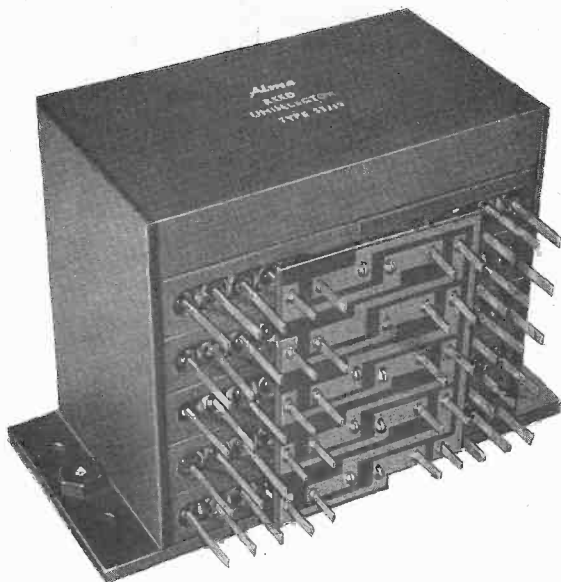
BRUSH CLEVITE COMPANY LIMITED

HYTHE SOUTHAMPTON

PHONE HYTHE 3031 GRAMS BRUDEV HYTHE SOUTHAMPTON TELEX 47687
WW-012 FOR FURTHER DETAILS.

REED

RELAYS REED UNISELECTOR



Designed for those special switching applications which are beyond the capabilities of conventional uniselectors and stepping switches. Any number of ways up to 100 can be supplied with up to 6 banks. Reverse operation can be provided and automatic homing to any pre-selected number.

Send for the provisional data sheet today.

- ★ Higher speed of operation
- ★ Fully sealed contacts requiring no adjustments or maintenance
- ★ Instantaneous re-set without going through all contacts
- ★ Changeover or normally closed contacts can be supplied if required
- ★ Independent "make" contacts are employed instead of a common wiper circuit
- ★ Low contact resistance

ALMA COMPONENTS LIMITED

Park Road • Diss • Norfolk • Telephone: Diss 2287 • Telex 18162

WW-013 FOR FURTHER DETAILS.



All Advance Transistorised Timer Counters offer more facilities than other makes at comparable prices. The current comprehensive range will be further extended during 1966 with the introduction of several new counters.

Advance Timer Counters provide frequency measurement up to 100 Mc/s, timing in micro-seconds and counting of random pulses up to 10^7 pulses per second. Accurate timing pulses variable from micro-seconds to seconds are obtainable with accuracies up to 2 parts in 10^6 , which by using optional external frequency standards can be increased to parts in 10^8 .

Frequency Measurement: 0-10 Mc/s, extended by Frequency Dividers to 100 Mc/s.

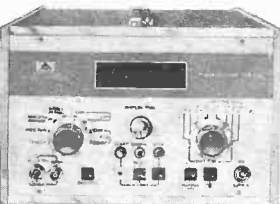
Gate Time: A wide range from 10 seconds to 1 millisecond.

Period Measurements: Multi period measurements from 10 to 10^6 periods.

Time Measurements: From micro-seconds to hundreds of hours (99999990 seconds).

Counting: 1-9999999 regular or random pulses.

ADVANCE TIMER COUNTERS



5 MC/s TIMER COUNTER TC4A

Frequency Measurement. Range: 0 to at least 5 Mc/s. **Gate Times:** 0.001, 0.01, 0.1, 1, 10S. Decimal point automatically positioned for display in Kc/s. **Sensitivity:** NORMAL input 10 c/s to 5 Mc/s, sinusoidal signal from 100 mV RMS. Maximum input 250V RMS up to 10 Kc/s and 10V RMS up to 5 Mc/s.

Period Measurement. Single and Multiple: Single period with variable time units or multiple periods up to 10^6 with μ S units.

Time Measurement. Unit time pulses: 1 μ S to 10S in decade steps. Maximum period displayed 10 6 S (approx. 28 hours).

Output Timing Pulses. Decode divisions: 10 8 pulses per second to one pulse per 10 seconds.

Counting. Range: 1 to 9999, regular or random pulses.

Frequency Standard. Internal: 1 Mc/s crystal oscillator, oven controlled at $\pm 65^\circ$ C. Accurate to 1 part 10^8 .

PRICE £205 ex Works

If you would like to know more about the Advance range of Timer Counters, please write for details or telephone for further information.



ADVANCE ELECTRONICS LIMITED

Instrument Division, Roebuck Road, Hainault, Ilford, Essex. Tel: Hainault 4444

G

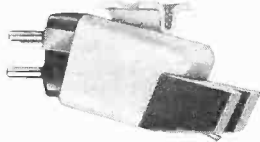
NEW SHAPES OF SOUND



The Goldring-Lenco GL 68

The many proved features of Goldring-Lenco transcription units such as infinitely variable speed adjustment, pick-up lowering device, automatic idler wheel disengagement are retained on the GL 68, which is the first unit to incorporate the new G.65 arm. This is of low mass tubular construction with stylus pressure adjustment by sliding counterweight, and provision for height adjustment to suit any chosen cartridge. The interchangeable head slide (taking all cartridges with standard $\frac{1}{2}$ " fixing centres) makes use of self-cleaning wiper contacts. Swiss precision motor. Continuously variable speed adjustment. Less than 1% speed variation for 13% mains voltage variation. Adjustable click-in positions for the four standard record speeds. Pick-up raising/lowering device coupled to on/off switch. Automatic disengagement of idler wheel. Full 12" diameter turntable. Wired for stereo.
GL 68 Transcription Unit £16.16.0d. + £2.14.7 P.T.

Recommended cartridges for the GL68 are: Pickering V15 (AM1 and AME1) Pickering 380A, Goldring CS90 & CS91E.



Pickering V15 AM1 and AME1 Micro-Magnetic Cartridge

Weighing only 5 gm., these high output, high compliance stereo—mono cartridges are perfect for low mass arms. 15° tracking angle gives minimum distortion. Hermetically sealed. Replaceable push-in diamond stylus assembly with retracting stylus arm for added protection to records. (0.0007" tip radius for V15/AM1 and elliptical, with even higher compliance, for V15/AME1).

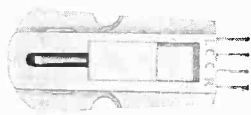
V15/AM1 £9.9.0d. + £1.10.9d. P.T.
V15/AME1 £13.15.0d. + £2.4.8d. P.T.



Pickering 380A Cartridge

Moving-magnet for exceptional mono or stereo reproduction. Features the exclusive V-guard push-in diamond stylus unit which prevents damage through accidental dropping of arm on record. The Pickering 380A ensures high channel

separation and virtually eliminates needle talk, hiss or distortion. Hermetically sealed, it tracks at 2 gm., faithfully reproduces the most exacting records.
£12.12.0d. + £2.0.11d. P.T.



Goldring CS90 and CS91E Cartridges

These are stereo ceramic cartridges with excellent frequency response and cross-talk separation. Low tip mass, replaceable diamond stylus (CS90—0.0005" or 0.0007" tip radius; CS91E—elliptical) coupled with high compliance enables these cartridges to be played at light tracking weights. CS 90 £4.4.0d. + £0.13.8d. P.T. CS91E £6.6.0d. + £1.0.6d. P.T.

C68 Cabinet and Cover for GL68. Elegant sapele mahogany cabinet with removable, clear Perspex dust cover. Size: 14" x 17" x 7".
£8.19.6d. + £1.12.0d. P.T.

G

Goldring Manufacturing Company (G.B.) Ltd.,
486-488 High Road, Leytonstone, London, E.11
Telephone: Leytonstone 6343

WW—015 FOR FURTHER DETAILS.

Continuous Frequency Coverage from 1.5 c/s to 150 kc/s in 5 Ranges at Decade Intervals

TRANSISTOR

R. C. OSCILLATORS

TYPES TG150, TG150M

TG150D & TG150DM

SPECIFICATION:

FREQUENCY

1.5 c/s to 150 kc/s
 $\pm 3\% \pm 0.15$ c/s.

STABILITY

<0.05% drift after 30 seconds.
<0.3% drift for 30%
fall of supply voltage.
<0.05% drift per °C at 1 kc/s.

DISTORTION

<0.1% at 1 kc/s;
<0.3% from 50 c/s to 15 kc/s;
<1.5% below 50 c/s
and above 15 kc/s.

SINE WAVE OUTPUT

Variable up to 2.5V
into 600 Ω .
<1% variation with frequency.
<0.5% change for
30% fall of supply voltage.

SQUARE WAVE OUTPUT

Variable up to 2.5V.
Rise time 1% of period $\pm 0.2\mu$ s

ATTENUATOR

20dB, 40dB and 60dB; 600 Ω

SUPPLY

Self-contained PP9
batteries, life 400 hours,
or, 200/250V A.C. when Power
Supply Unit is fitted.

SIZE

10in. high x 6in. wide x 4in. deep.

WEIGHT

6 pounds.



SEE US AT THE I.E.A. SHOW
STAND N.815
FIRST FLOOR NATIONAL
GALLERY MAY 23-28

TYPE	TG150	TG150M	TG150D	TG150DM
Output Waveforms:	Sine only	Sine only	Sine and Square	Sine and Square
Output Meter:	None	0-2.5V and dB	None	0-2.5V and dB
Price with batteries:	£32	£42	£35	£45

LEVELL

PORTABLE INSTRUMENTS

Mains Power Supply Unit £7.10.0

Leather carrying case £3.10.0

Fully detailed leaflets are available on our complete range of portable instruments

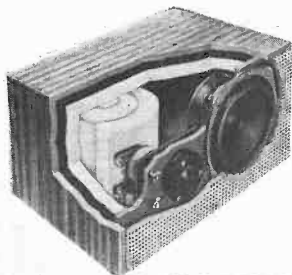
LEVELL ELECTRONICS LIMITED

PARK ROAD, HIGH BARNET, HERTS ————— Telephone: BARnet 5028

WW-016 FOR FURTHER DETAILS.

SUPERIOR PERFORMANCE

Celestion Studio Series LOUDSPEAKERS



The Ditton 10

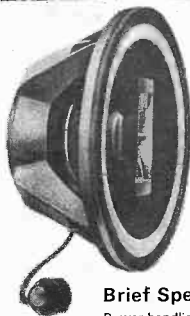
The Ditton 10 is a compact high fidelity reproducer measuring only 12 $\frac{3}{4}$ " x 6 $\frac{3}{4}$ " x 8 $\frac{1}{4}$ ". We, along with many thousands of satisfied users throughout the world, consider it to be the finest loudspeaker in its class—sales certainly substantiate this.

If you are in doubt ask your audio dealer to demonstrate the Ditton 10 against any other comparable system.

Brief Specification

Power handling capacity	10 watts R.M.S.
Overall frequency response	35-15,000 c/s
Impedance	15 ohms
Size	12 $\frac{3}{4}$ " x 6 $\frac{3}{4}$ " x 8 $\frac{1}{4}$ " (323mm x 171mm x 203mm)

Price £19.6.0. inc. P.T.



CX2012

When installed in a suitably designed enclosure the CX 2012 12" Co-axial loudspeaker provides truly professional sound quality.

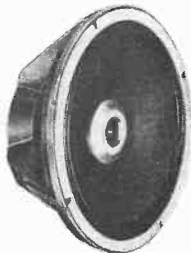
A highly compliant cone surround allows maximum linear movement and minimises harmonic distortion.

Sensitivity of the co-axially mounted, horn-loaded, compression high note unit may be adjusted by means of the "Brilliance" control provided.

Brief Specification

Power handling capacity	20 watts R.M.S.
Overall frequency response	30-18,000 c/s
Impedance	15/16 ohms

Price
£17.10.0.



CX1512

A lower powered alternative to Model CX 2012, Model CX 1512 provides the high standards of performance demanded by professional users.

Price £12.5.0.

Brief Specification

Power handling capacity	15 watts R.M.S.
Overall frequency response	30-15,000 c/s
Impedance	15/16 ohms

SMALLER ENCLOSURES

Celestion research engineers have now evolved two new enclosure designs of only 2.5 cu. ft. to accommodate the above co-axial loudspeakers. Their external dimensions (using $\frac{3}{4}$ " timber) are only 30" x 17 $\frac{1}{2}$ " x 11 $\frac{1}{2}$ ". Full details are given on the colour brochure.

Celestion

Studio
Series

CELESTION LTD

Celestion Ltd., Ferry Works, Thames Ditton, Surrey.

Please send me full details of the Studio Series range of loudspeakers.

Name _____

Address _____

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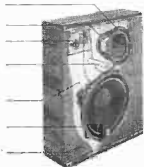
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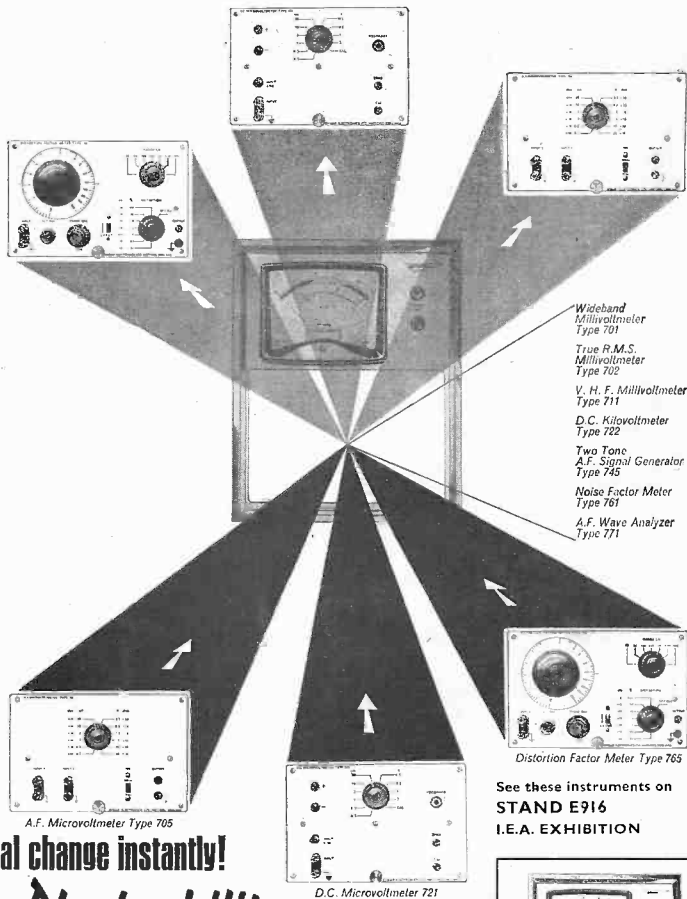
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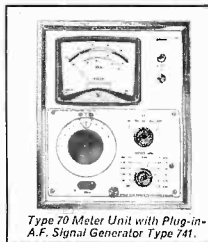
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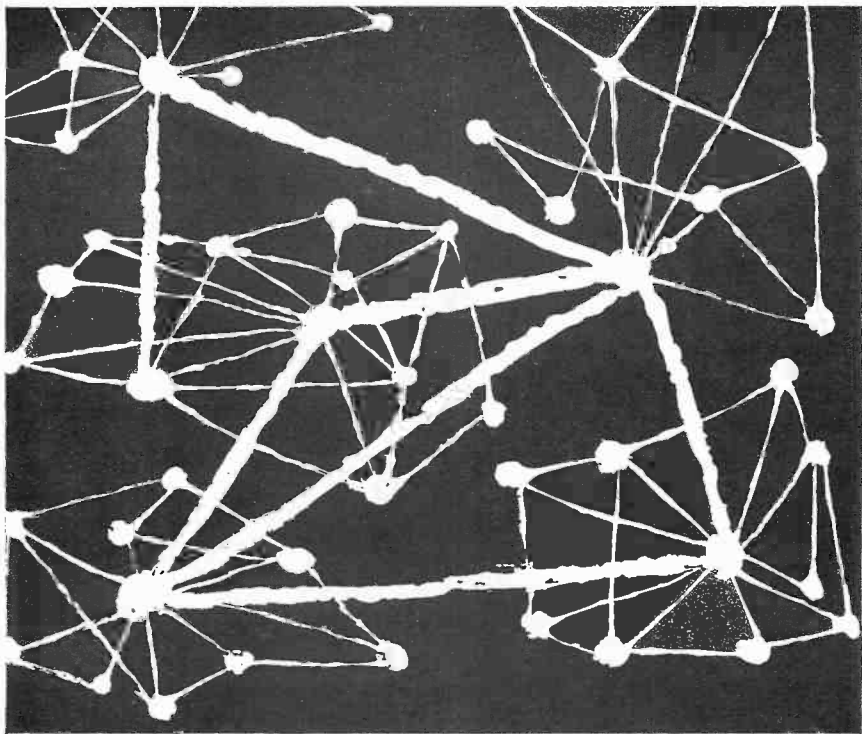
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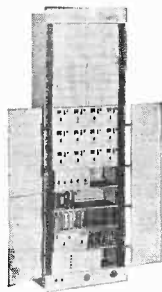
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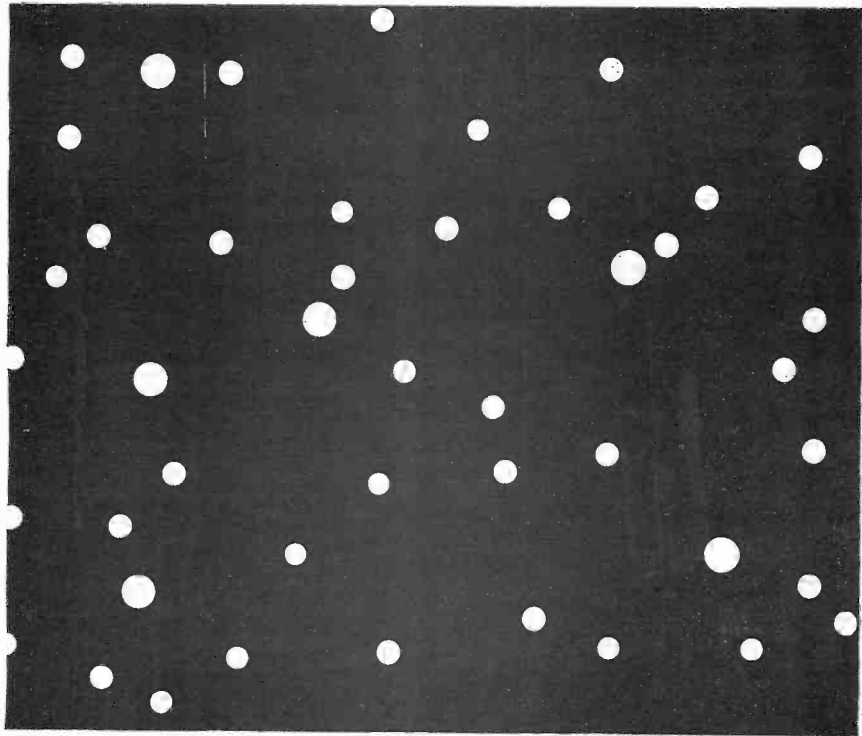
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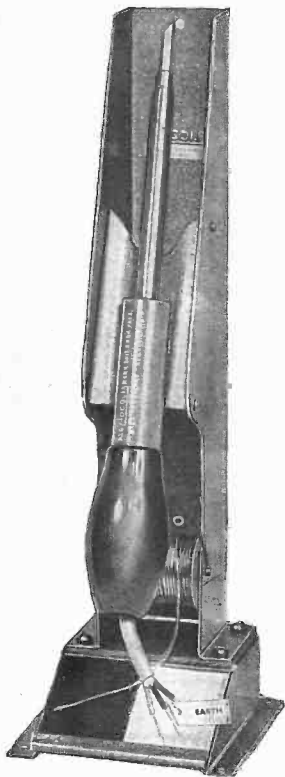


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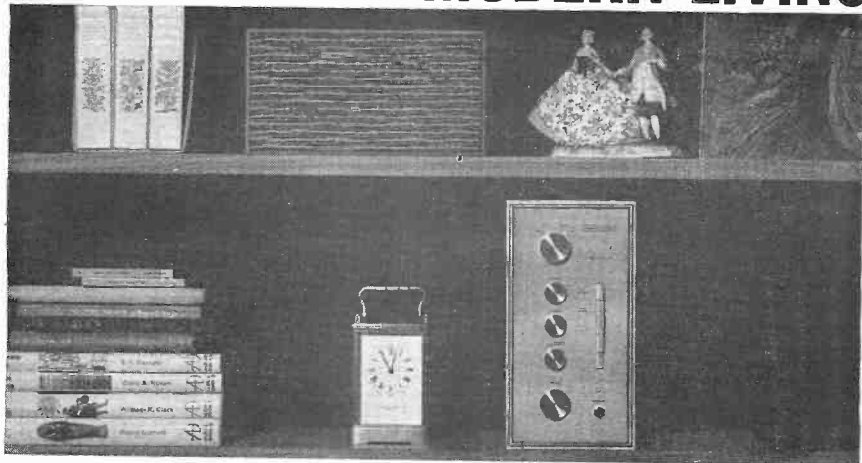
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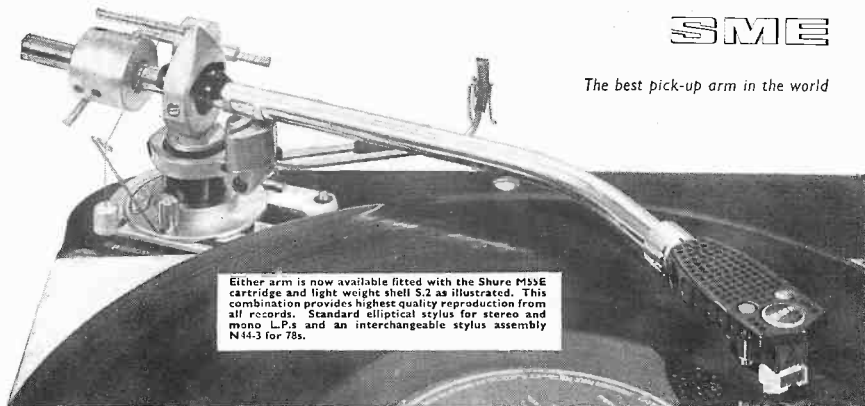
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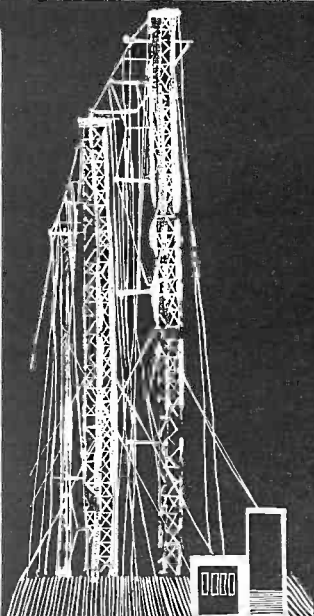
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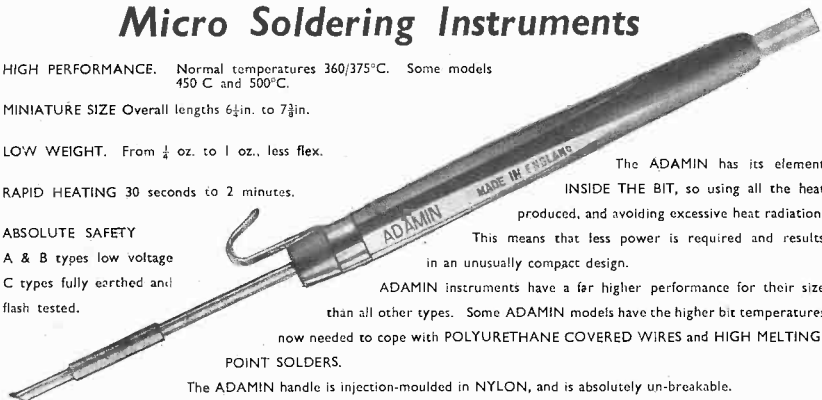
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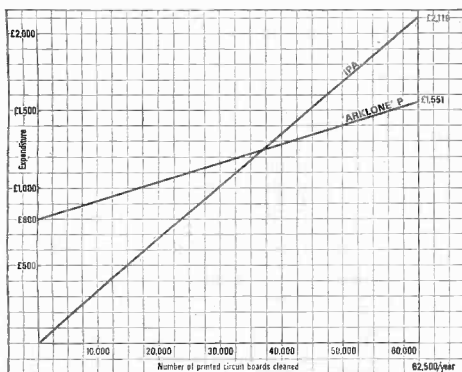
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'ARKLONE' is also first-class at cleaning computer core memories, tape recording and play-back heads, and other components consisting wholly or partly of plastics materials, or containing resins or elastomers—spectacle frames, for instance.

We'll say it again: 'ARKLONE' is a costly cleaning solvent that would be cheap at twenty times the price!

If you want to work out the cost of changing your cleaning process to 'ARKLONE'—get in touch with ICI. One of our representatives will be glad to assist you.

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Labour charges	2083	
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Services	nil	
Total/year	2119	8

AFTER

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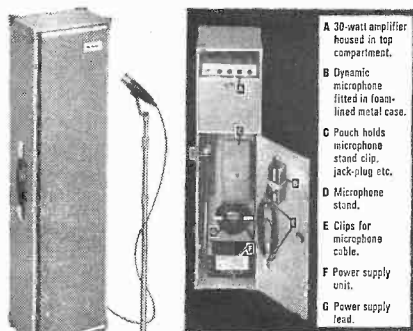
	£	s
Capital cost (plant written off in one year)	800	
Labour charges	500	
Solvent consumption (1 gallon/week at £4.13.9 per gallon)	234	
Services (electricity and water)	17	
Total/year	1551	

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'ARKLONE' is the ICI registered
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AR88



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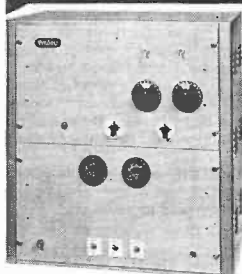
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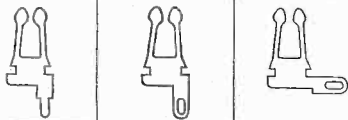
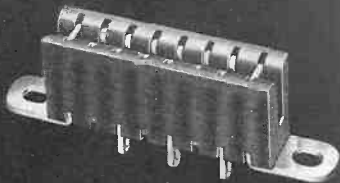
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Permacon edge connectors



35

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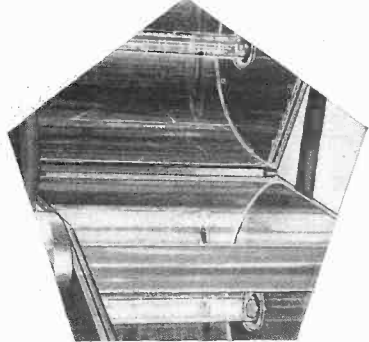
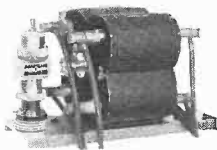
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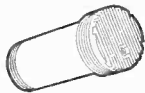
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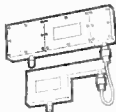
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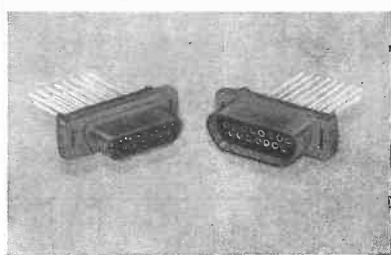
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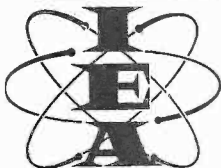


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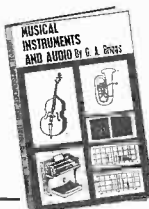
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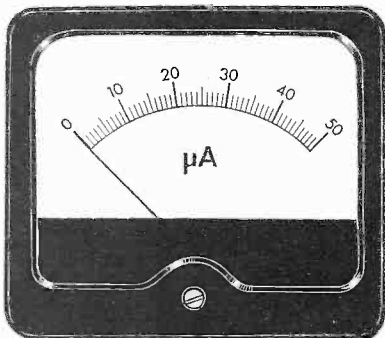
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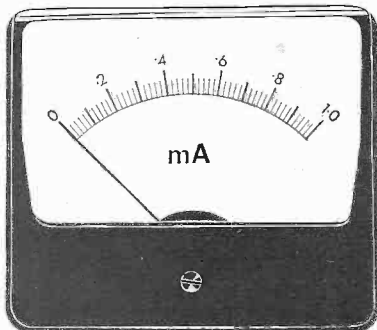
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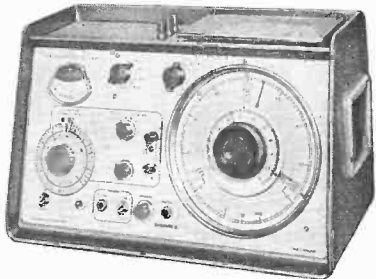
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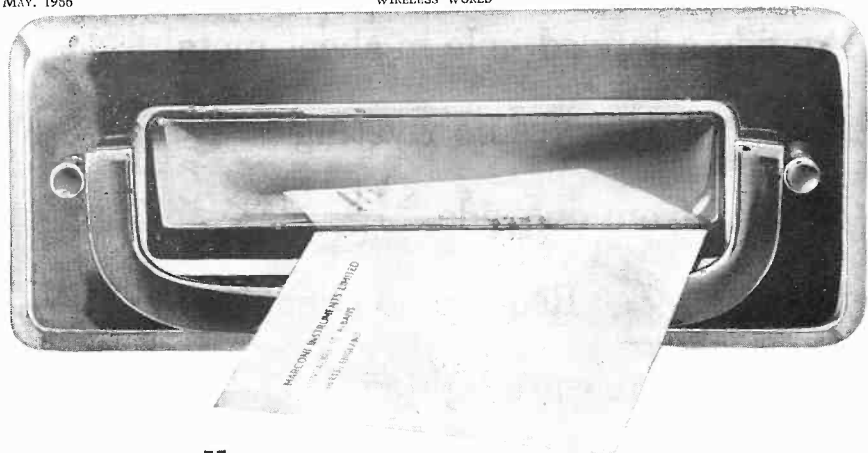
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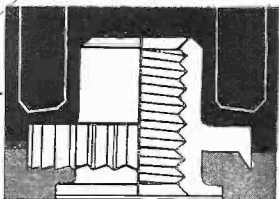
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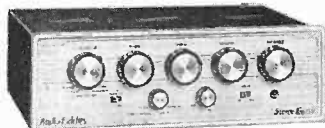
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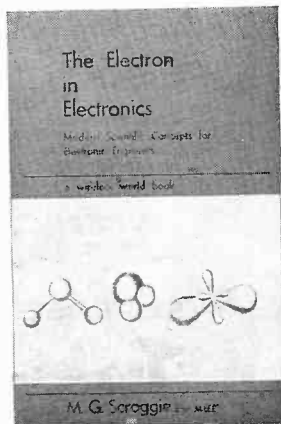
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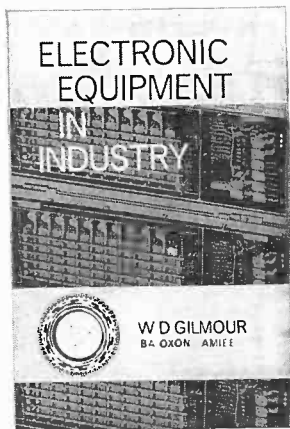
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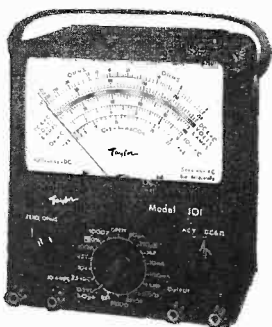
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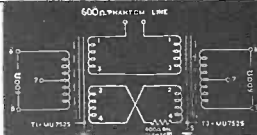
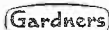
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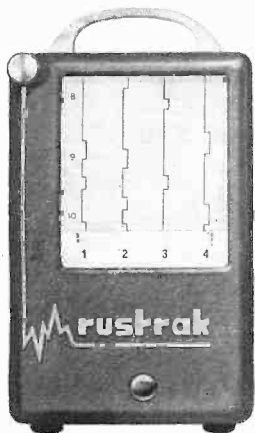
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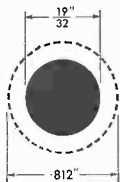
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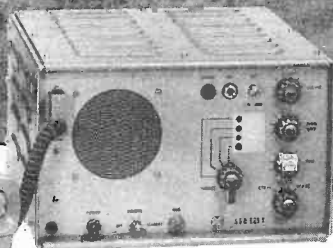
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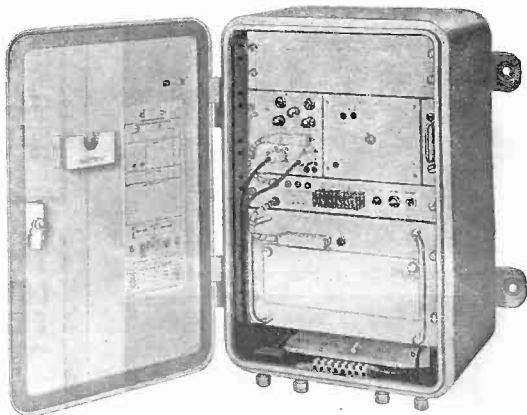
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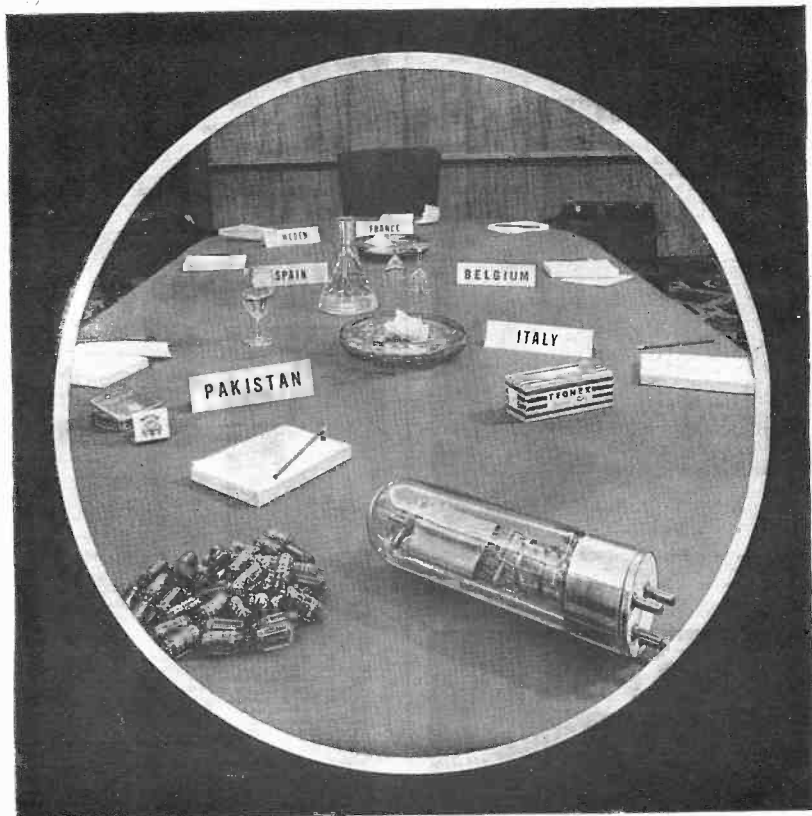
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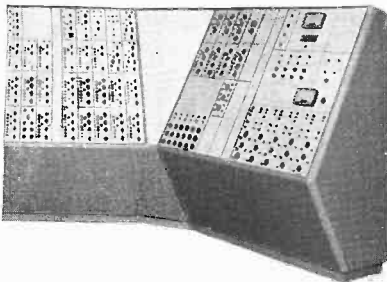


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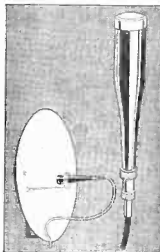
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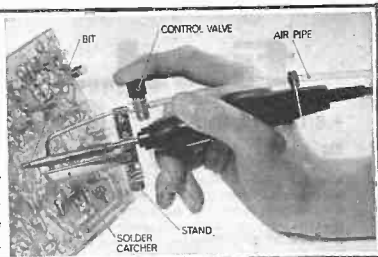
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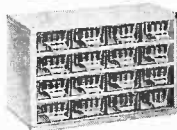
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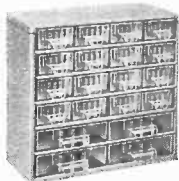
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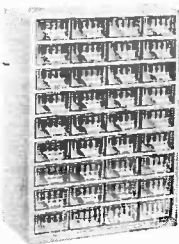


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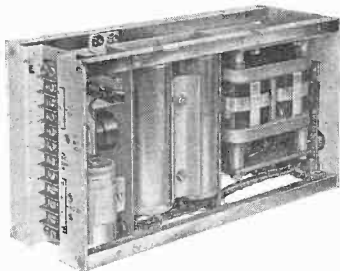
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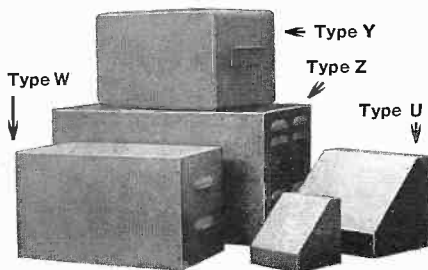
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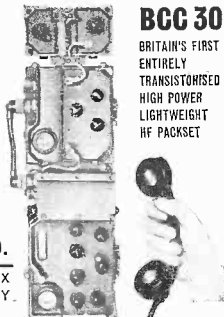
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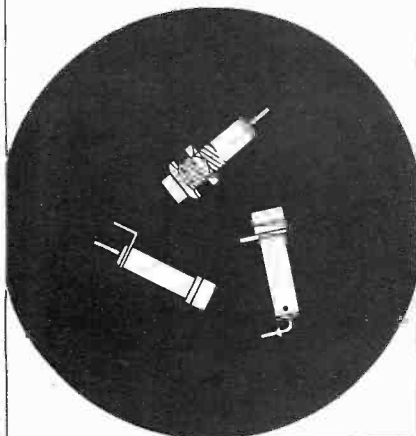
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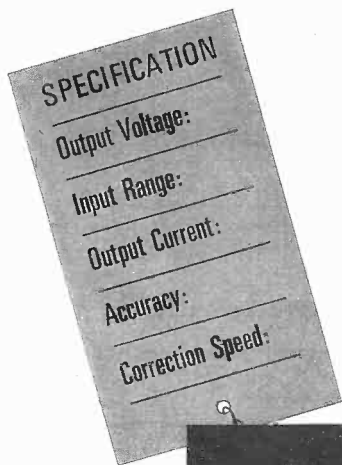
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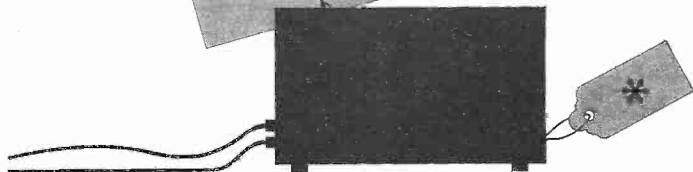
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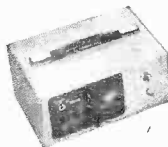
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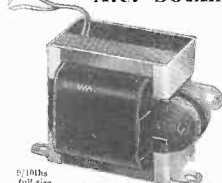


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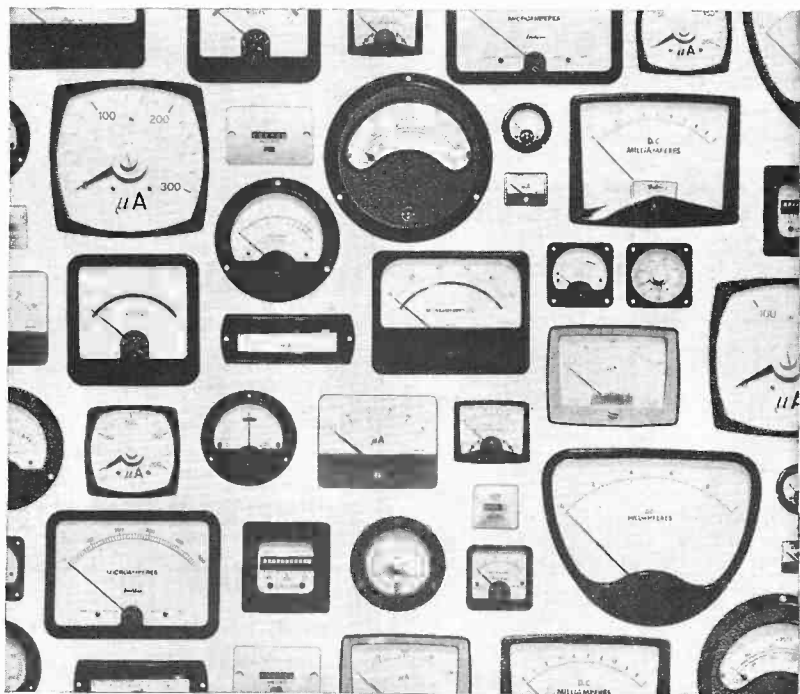
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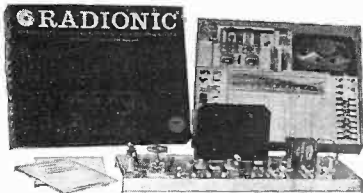
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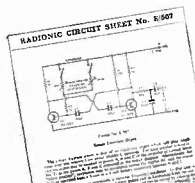
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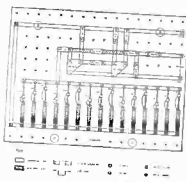
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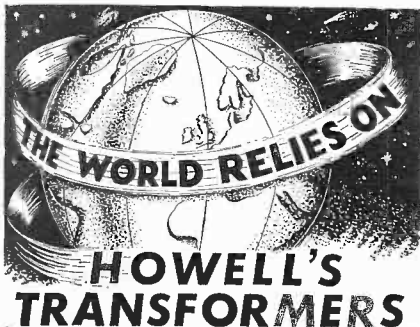
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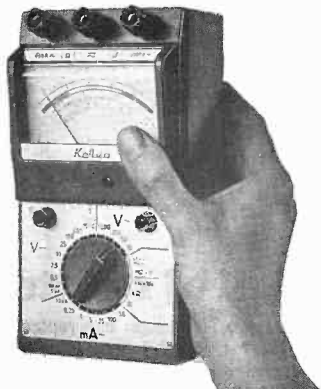
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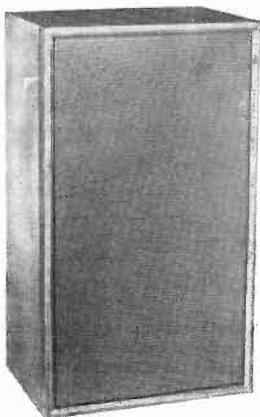
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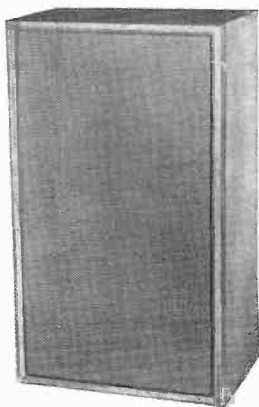
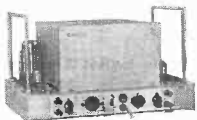
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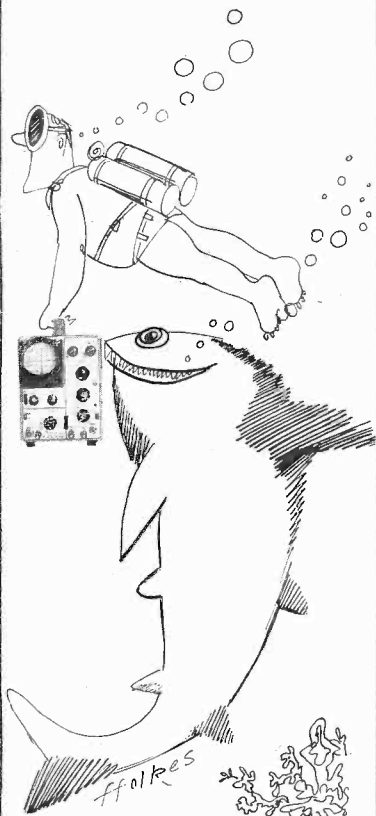


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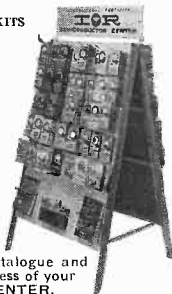


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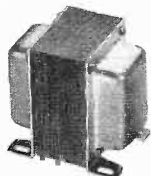
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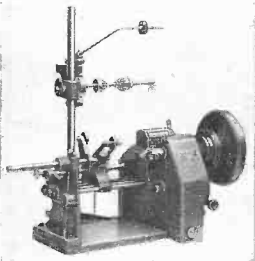
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ELECTRONICS, TELEVISION, RADIO, AUDIO

Crisis in the Electronics Industry?

INTEGRATED circuits are said to be creating a crisis in the U.K. electronics industry. Since very large scale production is needed to make i.c.s competitive with conventional circuits, there will be room for only the handful of big companies who can make the necessary capital investment for such production, and in the West these are likely to be American. Already U.S. companies have secured over half the British market, and our own manufacturers have their backs to the wall commercially. The real danger, as stated by British industrialists, is that if we cease to make our own i.c.s the U.K. electronics industry will degenerate into little more than a collection of assembly shops, wholly dependent on outside sources, with no incentive or facilities for developing new techniques and consequently nothing to sell in world markets. Indeed the situation could be worse, for i.c.s are not just components like transistors (which we already import in large quantities) but sub-systems—and sub-systems which are getting more and more comprehensive in function. As John Pickin, manager of Ferranti's Electronics Department, said recently "the equipment design of the future will be done on a slice of silicon, and if we do not have the ability to do this we will not have an electronic equipment industry in this country."

Is there really a crisis? One could argue that the U.K. industry is already highly dependent on imported products, licensing arrangements and so on, and that everyone has accepted the fact that it could not function without them. We have to recognize that other industrial countries also have to export electronic products and know-how in order to live and that consequently Britain must import a reasonable share. Our own ability to export depends on the continuing prosperity of these nations. This is the co-operative principle underlying the European Common Market, whether you agree with it or not. Our limited resources simply do not allow us to produce good and highly competitive products in every field of electronic manufacture, and our policy must be to take full advantage of those products which other countries design better and make cheaper. Are not integrated circuits just another phase in this process?

It could also be argued that electronic engineering is gradually moving towards, or even becoming part of, systems engineering, and that consequently there is much to be gained in efficiency by having your basic "components," the sub-systems, made for you by specialist manufacturers. In systems engineering even a computer is only a component. British firms have already done excellent business in selling plants and systems abroad—and these are firms which are virtually no more than "assembly shops." What is important here is knowing how to make the hardware do the job required.

But the basic problem is to ensure that Britain gets her fair share of the available cake—the world markets in electronics. With the resources we can muster for i.c. manufacture we cannot afford to fritter away our energies over the whole field of applications, but we *could* concentrate on a carefully selected range of circuits which we knew we could produce well and competitively. This implies rationalization of products among the manufacturers. One industrialist has already admitted that the U.K. has twice as many companies as necessary in i.c.s making half as much investment as necessary. In their present fresh and belligerent state of mind these companies are not likely to go into rationalization voluntarily. But if there is, indeed, a crisis, they may eventually be forced into it in self-defence—and this would be the wrong reason.

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PICKUP ARM DESIGN—1

DESIGN AND MOUNTING OF ARMS FOR MINIMUM DISTORTION DUE TO LATERAL TRACKING ERROR

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It is possible (as shown in this article) to design and mount a pickup arm 8 in long and of conventional shape so that the maximum distortion, due to lateral tracking error, of a fairly large signal (10 cm/sec r.m.s. recorded velocity) is less than 1%. In view of the small value, any attempt to reduce the tracking error further by means of additional levers, etc., will not only have a negligible effect on the distortion, but will increase the inertia of the pickup assembly, which is undesirable. Lateral tracking error distortion for a pickup which is poorly mounted so that this value becomes 5%, say, is still small compared with that from other sources. However, these other distortions (which are much greater for stereophonic records) are slowly being reduced with improvements in pickup arms, whereas tracking error is unaffected. In fact, the distortion from tracking error increases if elliptical stylus are used.

One may argue that a 1 or 2% reduction in harmonic distortion is insufficiently large to be worth bothering with. This is true if it involves a modification to the pickup arm resulting in greater expense. If, as shown, it merely involves slightly more care in offsetting the pickup and mounting the arm, the extra effort is certainly worth while.

ANYONE who constructs a pickup arm has to choose values for the offset angle of the cantilever assembly and the overhang. Anyone mounting a pickup need only consider the overhang in order to determine the optimum distance between the pickup arm and turntable pivots. There is a general belief that a pickup arm should be mounted for minimum angular tracking error and that the longer it is, the better. However, the distortion of a given modulation becomes greater as the groove speed decreases, and a pickup arm designed and mounted for minimum tracking error will have its maximum error on the worst possible occasion, namely at the innermost grooves of the record. As the length of an arm increases, its inertia (effective mass) at the stylus tip will usually increase, and this is undesirable.

Two designs are given. The first is for a general purpose record player suitable for 7 in, 10 in and 12 in discs, and the second is for a record player restricted to 7 in discs. In both cases, a range of arm lengths is considered, and the most suitable values for the offset angle and overhang are given. Also, for the first design, values of overhang are included for offset angles varying within limits of about $\pm 0.5^\circ$ from the recommended value. Provided that the offset angle of a pickup arm lies within this range, the mounting may be considered as optimum with the smaller angle slightly favouring 45 rev/min records and the larger angle favouring long-playing records.

The importance of mounting accurately is shown graphically. An error of 0.1 in in overhang or 2° in offset angle will more than double the distortion from long-playing records. In view of this, the distances between the stylus and turntable centre, at which the tracking error should be zero, have been given. These distances are independent of the length of the pickup arm and enable the accuracy of the mounting to be checked. A mounting procedure is given for reducing these errors. If the offset angle is fixed and slightly different from the recommended value, an alternative procedure is given for adjusting the overhang for optimum mounting. Neither of these methods requires the offset angle or overhang to be accurately determined. All that is required is an alignment protractor and such a device is obtainable from most hi-fi dealers.

Distortions in reproduction from discs

The quality of reproduction from discs is limited mainly by three factors. Firstly, tracking error, which is the error in alignment between the cantilever assembly and the groove in which the stylus is located. Lateral (horizontal) tracking error is the angle, in the horizontal plane, between the groove, or direction of motion, and the cantilever, and for a conventional pickup arm the lateral tracking error varies with distance from the turntable centre. Vertical tracking error, which only affects stereophonic records, is the difference between the angle which the stylus makes with the vertical when slightly lifted (the arm remaining stationary), and the value specified for the particular record being played. The second factor limiting the quality of reproduction is groove deformation due to the plastic record material being compressed by the stylus tip. However, this distortion tends to reduce the third and most serious form of distortion, tracing distortion due to the cutter and stylus being of a different shape. The flat edge of the V-shaped cutter faces the direction of record motion, and therefore an absence of signal corresponds to a large groove width. As a result of a signal causing the

cutter to move horizontally, as in the case of a monophonic signal or two stereophonic signals approximately equal in amplitude and phase, the groove width becomes narrower causing a stylus of spherical tip to rise. This is the pinch effect. The high-frequency response of a pickup is limited by the radius of the stylus tip since the finite width in the direction of record motion limits the ability to follow groove modulations. The quality of reproduction is reduced further if the tracking mass is insufficient for the stylus to contend with large high-frequency velocities (for which a small effective tip mass is favoured) and low-frequency amplitudes (for which a large compliance is favoured).

Elliptical styli

In articles on pickup design, lateral tracking error is usually only briefly considered since the distortion is generally very much less than that resulting from other causes. However, in the last few years, these other forms of distortion have been reduced considerably. With pickups of lower effective tip mass and higher compliance, lower tracking masses have been possible. This has enabled styli of smaller radius (0.0005 in and less) to be used with a subsequent reduction in tracing distortion. Tracing distortion has been reduced further by the adoption of elliptical styli, mounted so that the major axes face the direction of record motion. The "vertical" cutting angle has been standardized in the U.S.A. at 15° and it should not be long before this value becomes universal. The result will then be a considerable reduction in vertical tracking error and subsequent distortion when the latest stereophonic records are tracked by the latest pickups, i.e. pickups designed so that a line joining the stylus tip and cantilever pivot is inclined at 15° to the record surface.

Despite their many advantages, the use of elliptical styli results in an increase in distortion from lateral tracking error by causing a time-lag effect. With a spherical tip, the points of contact with the walls of unmodulated grooves and at the crests of waves are always the same as those of the cutter, i.e. a line through these points will be perpendicular to the direction of record motion. However, if the stylus is elliptical and tracking error exists, then one point of contact will be slightly ahead of the other. In these circumstances, tracking error cannot be ignored. Even with spherical styli, it is certainly worth spending a few minutes extra in mounting a pickup arm if, as a result, the distortion due to lateral tracking error is reduced to a minimum, perhaps even halved.

Distortion due to tracking error

The main form of distortion due to tracking error is second harmonic and is given by H. G. Baerwald¹ as:—

$$\epsilon = 100 \frac{V_0 \tan \eta}{u} \approx 100 \frac{V_0 \eta}{u} \dots \dots \dots 1$$

where ϵ = % 2nd harmonic distortion, V_0 = peak recorded velocity (cm/sec), η = tracking error in radians, and u = groove speed (cm/sec).

Third and higher harmonic distortions amount to less than 10% of this value and are considered negligible. Using the expressions:—

$$V_{rms} = \frac{V_0}{\sqrt{2}} = \text{r.m.s. or effective recorded velocity (cm/sec)}$$

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$$\phi = \frac{360\eta}{2\pi} = \text{tracking error (degrees)}$$

x = distance between stylus tip and turntable pivot (in)

s = turntable speed (rev/min),

$$\text{then } u = (2\pi \cdot 2.54x) \frac{s}{60}$$

and equation 1 may more conveniently be put in the form

$$\epsilon = 100 \frac{\sqrt{2}}{2.54 \times 60} \cdot \frac{V_{rms} \phi}{xs} = 9.28 \frac{V_{rms} \phi}{xs}$$

A typical maximum value for V_{rms} may be considered as 10. Therefore,

$$\epsilon_{max} = 92.8 \frac{\phi}{xs}$$

The sign of ϕ (positive or negative) is ignored as it has no effect on the distortion. In order to prevent an excessive groove amplitude in the bass and to improve the signal to noise ratio, a gain of approximately 4 dB per octave with increasing frequency is applied to the recorded sound before it reaches the cutting heads. A reverse characteristic is used for replay. As a result, harmonic distortions are effectively reduced by 4 dB per octave. The effective maximum second harmonic distortion ϵ'_{max} is as follows,

$$\epsilon'_{max} = 10^{-1/20} \epsilon_{max} = 0.631 \epsilon_{max} = 58.5 \frac{\phi}{xs}$$

since the variation in pickup voltage is proportional to the recorded velocity. Hence,

$$\begin{aligned} \epsilon'_{max} &= 1.76 \frac{\phi}{x} \dots \dots \dots 33\frac{1}{2} \text{ rev/min} \\ &= 1.30 \frac{\phi}{x} \dots \dots \dots 45 \text{ rev/min} \end{aligned}$$

The values for harmonic distortion in this article are for spherical styli; other distortions due to tracking error are considered in Appendix 1 (in the next issue). The important point is that the distortions depend on ϕ/x .

Therefore, in a design ϕ/x should be minimized, and not ϕ alone.

Design procedure

Fig. 1 shows tracking errors for a straight pickup arm. By offsetting the pickup the errors are reduced considerably as shown in Fig. 2. Using the symbols in Fig. 3, we wish to determine the optimum values of θ_0 and f when x lies between given limits. We therefore require a value of θ which will remain almost constant as x varies within this range.

From Fig. 3:—

$$d^2 = x^2 + l^2 - 2xl \cos \beta = x^2 + l^2 - 2xl \sin \theta$$

¹H. G. Baerwald, *J.S.M.P.I.E.*, p.591-662, 1941.

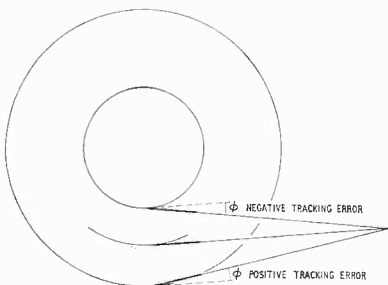


Fig. 1. Showing tracking errors for a pickup arm without offset.

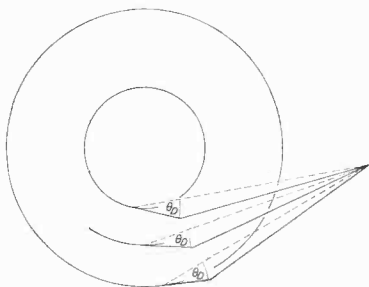


Fig. 2. Pickup arm with offset, θ_0 . By offsetting the pickup, tracking errors are reduced considerably. Note that the stylus moves very slightly in the direction of the record motion.

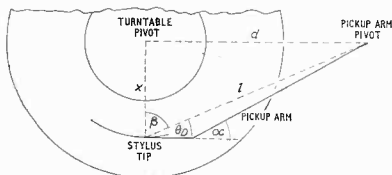
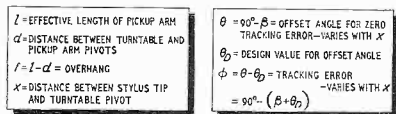


Fig. 3. Key to symbols used. The tracking error is zero when $\beta + \theta_0 = 90^\circ$. Optimum values for offset, θ_0 , and overhang f for x between giving limits are derived in the text. The offset angle of the pickup should not be confused with α , the angle through which the arm is bent.

$$\text{Rearranging, } 2l \sin \theta = x + \frac{l^2 - d^2}{x} = y \quad \dots (2)$$

where $d = l - f$.

We will consider first the determination of θ_D and f for minimum angular tracking error, and then show how these expressions are modified to reduce the maximum distortion of a given modulation, which is proportional to groove speed and therefore is inversely proportional to x .

Design for minimum angular tracking error

A value of θ is required which varies least in equation 2 when x varies between x_1 and x_2 ; l and d are constants. If we plot y against x we will obtain a curve which starts at infinity when $x=0$, reduces to a minimum value, and then increases to infinity at $x=\infty$ as in Fig. 4. The minimum at $x=x_m$ is given by differentiating equation 2 with respect to x , and equating to zero.

$$\frac{dy}{dx} = 0 = 1 - \frac{l^2 - d^2}{x^2}$$

$$\text{Hence, } x_m = \sqrt{(l^2 - d^2)}$$

Since l is given, x_m is varied by altering d . It is clear that x_m should lie between x_1 and x_2 if we wish the variation in $2l \sin \theta$ to be as small as possible for x varying between x_1 and x_2 . In particular, the variation will be a minimum if the values of y at $x = x_1$ and $x = x_2$ are equal. From equation 2,

$$x_1 + \frac{l^2 - d^2}{x_1} = x_2 + \frac{l^2 - d^2}{x_2}$$

$$\text{Hence, } x_1 x_2 = l^2 - d^2 = x_m^2 \quad \dots (3)$$

The minimum value of y then occurs at the geometric mean of x_1 and x_2 . Using equation 3,

$$d = (l^2 - x_1 x_2)^{1/2} = l - f$$

The overhang f is therefore given by:—

$$f = l - (l^2 - x_1 x_2)^{1/2}$$

The maximum and minimum values of y are given from equations 2 and 3,

$$y_{max} = x_1 + x_2 = 2/\sin \theta_{min}$$

$$y_{min} = 2(x_1 x_2)^{1/2} = 2/\sin \theta_{max}$$

The design value for θ_D is then as follows:

$$\theta_D = \frac{1}{2}(\theta_{min} + \theta_{max}) = \frac{1}{2} \left[\sin^{-1} \frac{(x_1 x_2)^{1/2}}{l} + \sin^{-1} \frac{x_1 + x_2}{2l} \right]$$

$$\text{or } \sin \theta_D \approx \frac{1}{2}(\sin \theta_{min} + \sin \theta_{max}) = \frac{(x_1^{1/2} + x_2^{1/2})^2}{4l}$$

Design for minimum tracking error distortion

In determining the pickup parameters for minimum lateral tracking error, we minimized the variation in $2l \sin \theta$ for x varying between x_1 and x_2 by including a turning point in the curve of $2l \sin \theta$ against x at $x = (x_1 x_2)^{1/2}$. $2l \sin \theta$ then has the same value at x_1 and x_2 . In this case, by minimizing the variation in $\sin \theta$ (l being constant), we automatically minimized the variation in θ , as required i.e. ϕ is as small as possible for $x_1 < x < x_2$.

Since for a given tracking error the distortion is inversely proportional to x , it is preferable for $\phi/x = (\theta - \theta_D)/x$ to be as small as possible. Most forms of distortion increase as the pickup approaches the turntable centre, so it is desirable for the distortion due to lateral tracking error to be a very small at the average minimum value of x . Let the tracking error be zero at $x = x_0$. We have mentioned Fig. 4 which gives $2l \sin \theta$ against x . Fig. 5 gives $2l \sin \theta/x$ against x and in addition $2l \sin \theta/x$ against x . The distance between these curves at a given

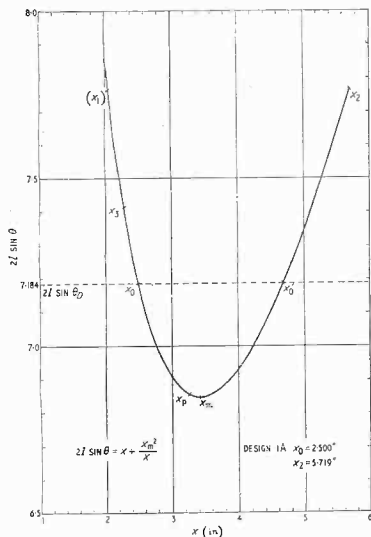


Fig. 4. Variation of $2l \sin \theta$ with x for a pickup arm designed for zero tracking error at x_0 and minimum distortion between x_0 and x_2 . As $x_0 < x < x_2$ the maximum harmonic distortion ($\propto |(\theta - \theta_0)|/x$) decreases to zero, increases to a maximum at x_0 (the same value of distortion as at x_2) and then decreases to zero at x_0 . Note that the largest negative value of $2l(\sin \theta - \sin \theta_0)$ occurs at x_m and the largest negative value of $2l(\sin \theta - \sin \theta_0)/x$ at x_p .

value of x is $2l(\sin \theta - \sin \theta_0)/x$ which we will minimize since it is approximately proportional to $(\theta - \theta_0)/x$ and allows a considerable simplification in the algebra.

Both $2l \sin \theta$ against x and $2l(\sin \theta - \sin \theta_0)$ against x have a minimum at x_m , whereas $2l(\sin \theta - \sin \theta_0)/x$ against x has a minimum at a value of x less than x_m , which we will call x_p . From equations 2 and 3:

$$2l \sin \theta = x + \frac{x_m^2}{x}$$

Hence,

$$\frac{2l(\sin \theta - \sin \theta_0)}{x} = \frac{x + \frac{x_m^2}{x} - \left(x_0 + \frac{x_m^2}{x_0}\right)}{x} = W \dots 4$$

W against x has a minimum at $x = x_p$. Hence, from Fig. 5, x_p is the value of x at which the continuous line is the furthest distance below the interrupted line. Differentiating equation 4 with respect to x and equating to zero:

$$\frac{dW}{dx} = 0 = -\frac{2x_m^2}{x^2} + \frac{x_0}{x^2} + \frac{x_m^2}{x_0 x^2} = \frac{1}{x^2} \left(x_0 + \frac{x_m^2}{x_0} - \frac{2x_m^2}{x} \right)$$

$$\therefore x_p = \frac{2x_m^2}{x_0 + \frac{x_m^2}{x_0}} = \frac{2x_0 x_m^2}{x_0^2 + x_m^2}$$

Ideally, we require $\frac{\theta_D - \theta_p}{x_p} = \frac{\theta_2 - \theta_0}{x_2} \dots 5$

$$\text{but since } \sin \theta_D - \sin \theta_p = 2 \sin \frac{\theta_D - \theta_p}{2} \cos \frac{\theta_D + \theta_p}{2}$$

$$\approx \frac{\pi}{180} (\theta_D - \theta_p) \cos \theta_D$$

$$\text{and } \sin \theta_2 - \sin \theta_0 = 2 \sin \frac{\theta_2 - \theta_0}{2} \cos \frac{\theta_2 + \theta_0}{2}$$

$$\approx \frac{\pi}{180} (\theta_2 - \theta_0) \cos \theta_0$$

we may replace equation 5 by

$$\frac{2l(\sin \theta_D - \sin \theta_p)}{x_p} = \frac{2l(\sin \theta_2 - \sin \theta_0)}{x_2}$$

This may be put in the form:

$$\begin{aligned} x_0 + \frac{x_m^2}{x_0} - \left[\frac{2x_0 x_m^2}{x_0^2 + x_m^2} + \frac{1}{2} \left(x_0 + \frac{x_m^2}{x_0} \right) \right] \\ = \frac{2x_0 x_m^2}{x_0^2 + x_m^2} \\ = x_2 + \frac{x_m^2}{x_2} - \left(x_0 + \frac{x_m^2}{x_0} \right) \end{aligned}$$

Simplifying,

$$x_m^2 \left(\frac{2x_0}{x_0^2} - \frac{2}{x_2} + \frac{1}{2x_0} \right) + x_m^2 \left(3x_0 - \frac{2x_0^2}{x_2} \right) \frac{x_0^2}{2} = 0$$

$$\text{Hence, } x_m^2 = \frac{x_0^2 x_2 (5.828x_0 - 4.828x_2)}{x_0^2 + 4x_0 x_2 - 4x_2^2}$$

This reduces to:

$$x_m^2 = \frac{x_0^2 x_2}{0.8284x_0 + 0.1716x_2} \dots \dots 6$$

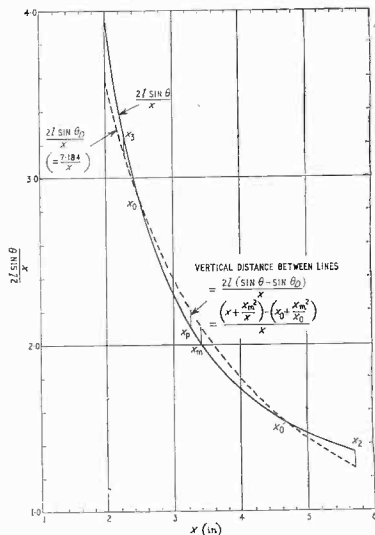


Fig. 5. Variation of $2l \sin \theta$ with x . The vertical distance between the lines is proportional to the harmonic distortion of a given modulation.

The design value θ_D for a given value of l is given from:

$$\sin \theta_D = \sin \theta_0 = \frac{1}{2l} \left(x_0 + \frac{x_m^2}{x_0} \right) = \frac{1}{l} \frac{x_m^2}{x_0} \quad \dots \quad 7$$

since $\theta_D = \theta_0$. The overhang f is given by:

$$f = l - (l^2 - x_m^2)^{1/2} \quad \dots \quad 8$$

All we require are values for x_0 and x_m . x_m^2 is then determined using equation 6 and the value substituted in equations 7 and 8 to determine θ_D and f for a given value of l .

The value of x less than x_0 at which the harmonic distortion (of a given modulation) is the same as at x_0 and x_2 we will call x_3 . This is given as follows:

$$\frac{\left(x + \frac{x_m^2}{x} \right) - \left(x_0 + \frac{x_m^2}{x_0} \right)}{x} = \frac{\left(x_2 + \frac{x_m^2}{x_2} \right) - \left(x_0 + \frac{x_m^2}{x_0} \right)}{x_2}$$

$$= \frac{\left(x_0 + \frac{x_m^2}{x_0} \right) - \left(x_p + \frac{x_m^2}{x_p} \right)}{x_p}$$

From the first two expressions,

$$\frac{1}{x_2} \left(x_0 + \frac{x_m^2}{x_0} - \frac{x_m^2}{x_2} \right) x^2 - \left(x_0 + \frac{x_m^2}{x_0} \right) x + x_m^2 = 0$$

The two solutions for x are x_2 and x_3 .

$$\therefore x_2 + x_3 = \frac{1}{x_2} \left(x_0 + \frac{x_m^2}{x_0} - \frac{x_m^2}{x_2} \right)$$

$$\text{Hence, } x_3 = \frac{\left(x_0 + \frac{x_m^2}{x_0} \right) - \frac{x_m^2}{x_2}}{x_2}$$

The tracking error changes from positive to negative between x_2 and x_p . Let it be zero at a value of x which we will call x'_0 .

$$x + \frac{x_m^2}{x} = x_0 + \frac{x_m^2}{x_0}$$

$$\text{Rearranging, } x^2 - \left(x_0 + \frac{x_m^2}{x_0} \right) x + x_m^2 = 0$$

Solutions are x_0 and x'_0 .

$$\therefore x_0 + x'_0 = x_0 + \frac{x_m^2}{x_0}$$

so $x'_0 = \frac{x_m^2}{x_0}$, as expected.

Note that whatever design method is used to obtain θ_D and f , the tracking error at a given value of x is given by:

$$\phi = \sin^{-1} \left[\frac{x + f(2l-f)}{2l} \right] - \theta_D \quad \dots \quad 9$$

i.e. $\phi = \theta - \theta_D$

where θ is given from equation 2 rewritten in the form:

$$\sin \theta = \frac{x + f(2l-f)}{2l}$$

Equation 9 was used to compare the distortion of a given modulation arising from a pickup mounted as suggested here with that obtained if Bauer's method is used. Results will be given in Appendix 11.

Summary of design formulae

As x decreases, we have shown that the tracking error changes from positive to negative and back to positive again. The tracking error per unit length is zero at x_0 and x'_0 and the maximum negative value (at x_p) is equal to the positive values at x_2 and x_3 , as indicated in Fig. 5.

The values of x are related as follows:

$$x_0 + \frac{x_m^2}{x_0} = x'_0 + \frac{x_m^2}{x'_0}$$

$$\left(x_2 + \frac{x_m^2}{x_2} \right) - \left(x_0 + \frac{x_m^2}{x_0} \right) = \left(x_3 + \frac{x_m^2}{x_3} \right) - \left(x_0 + \frac{x_m^2}{x_0} \right)$$

$$= \frac{\left(x_0 + \frac{x_m^2}{x_0} \right) - \left(x_p + \frac{x_m^2}{x_p} \right)}{x_p}$$

$$x_p = \frac{2x_m^2}{x_0 + \frac{x_m^2}{x_0}}$$

where $x'_0 > x_0, x_2 > x_3$.

We must specify two values of x to obtain a solution.

If we specify x_0 and x_2 (as in the given designs), then:

$$x'_0 = \frac{x_m^2}{x_0}$$

$$x_p = \frac{2x_0 x_m^2}{x_0^2 + x_m^2}$$

$$x_3 = \frac{\left(x_0 + \frac{x_m^2}{x_0} \right) - \frac{x_m^2}{x_2}}{x_2}$$

$$= \frac{x_0^2 x_2}{0.8284x_0 + 0.1716x_2}$$

where $x_m^2 = \frac{0.8284x_0 + 0.1716x_2}{x_0^2 + x_2^2}$

Alternatively, we can give values for x_2 and x_3 , the limits of x between which we require the tracking error per unit length to be minimized. Then it is easily shown that:

$$x_0 = \frac{x_2 x_3}{0.8536x_2 + 0.1464x_3}$$

$$x'_0 = \frac{x_m^2}{x_0} = \frac{x_2 x_3}{0.1464x_2 + 0.8536x_3}$$

$$x_p = \frac{2x_2 x_3}{x_2 + x_3}$$

where $x_m^2 = \frac{8x_2^2 x_3^2}{x_2^2 + x_3^2 + 6x_2 x_3}$

The offset angle θ_D and overhang f for minimum tracking error per unit length are then as follows:—

$$\sin \theta_D = \frac{1}{l} \cdot \frac{x_m^2}{x_p}$$

$$f = l - (l^2 - x_m^2)^{1/2}$$

To be concluded.

Acknowledgement is due to Alfred Imhof Ltd. for the use of their facilities in the production of our front cover.

Transistor Electronic Organ

1—ELECTRONIC ORGAN DESIGN TRENDS

By T. D. TOWERS, M.B.E.

THE electronic-organ-building bug is a curious one. Bitten by it, many thousands have become afflicted with *dementia organistra*. The first clinical signs displayed by the victim of the disease (now assuming epidemic proportions) is a delusion that without substantial mechanical, electronic and musical skills he can construct one of these complex mechanical, electronic, musical instruments. This article is written with compassion for those who feel the first stirring of the *dementia organistra* fever in their blood, and cannot bring themselves to accept the sage advice from the pundits (usually reserved for those contemplating marriage!) . . . "Don't." It tries to give the happy victim some helpful guidance by setting out the main design features of present day electronic organs.

WHAT AN ELECTRONIC ORGAN IS

Organs have been around for a long time now. My favourite illustration of this is the ancient organ pictured in Fig. 1(a). This appeared in a famous medieval Psalter, the "Utrecht," and different musicologists have ascribed various dates from A.D. 600 to A.D. 1000 to the instrument. It may seem a far cry from it to the modern domestic electronic organ, which nowadays often takes a form somewhat like Fig. 1(b). Yet both belong to the same organ class of instrument; i.e., they are keyboard instruments in which a note sounds when a playing key is depressed and continues to sound until the key is released.

"Electronic" v. "wind" organs.—How an electronic organ differs in fundamentals from a classical "wind" organ is illustrated schematically in Fig. 2.

In the wind organ of Fig. 2(a), a bellows or pump-type compressor, A, forces air under pressure through a pipe, B, into a high-pressure reservoir ("wind-chest"), C. By operating one of the keys in the keyboard, E, the player can let compressed air pass through ducts D and F on the start of its way to make the organ pipe, I, sound or "speak." However, the "stops," G, in their off position stop the flow of wind to the pipe, I, even though a note is depressed on the keyboard. By putting on a selected stop, the performer opens the way for air to flow to a selected set or "rank" of pipes designed to produce a certain type of sound. The organ pipe is like a large whistle blown by the wind pressure released into it when the appropriate stop and key are actuated. The stop selects a set of pipes of one timbre, and the playing key selects the pipe of the proper pitch in that set.

Some ranks of pipes are mounted in free air, but others are enclosed in a sound-insulated enclosure, J, in Fig. 2(a), known as a "swell cabinet." Sound from these enclosed pipes emerges through a sort of venetian blind, K, and its apparent loudness depends on how far the blind is opened. A foot-operated "swell" or "expression" pedal, L, controls the opening of the venetian blind, and thus the apparent volume of sound from the "swell" ranks of pipes. The player manipulates the organ from the console (shown dotted) which presents to him the keyboards (usually more than one), the stop actuators ("pistons" or "tabs") and the swell pedal.

The block schematic of an electronic organ is given in Fig. 2(b), laid out to parallel the functions of its various sections with those of the wind organ above it, so as to highlight the differences. An electrical supply,

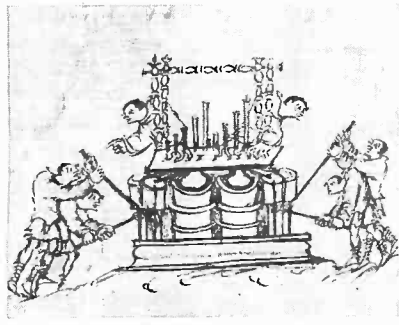
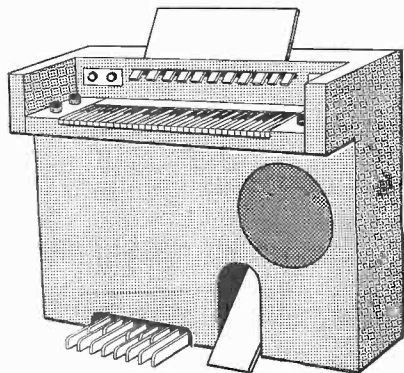


Fig. 1. The organ over the years: (a) illustration from early Psalter of pre-A.D. 1000 organ, well-known to organ antiquarians; (b) sketch of typical modern domestic electronic organ.



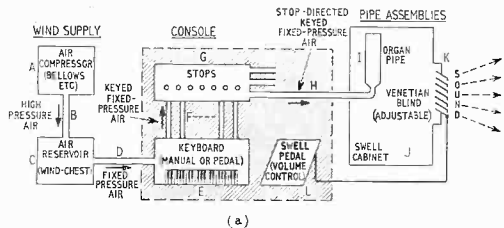


Fig. 2. Diagrammatic illustration of the essential differences between a traditional "wind" organ (a) and an electronic organ (b).

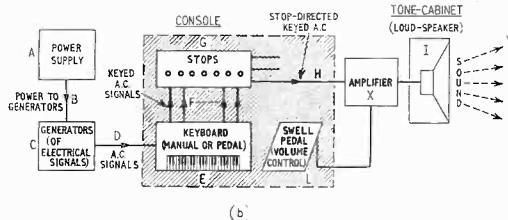


Fig. 3. Basic differences between two main classifications of electronic organs: (a) continuously-running generators and (b) switched generators.

A, provides power for the generators, C, of electrical a.c. signals of various frequencies. As with the wind organ, the console provides keyboard(s) E, stop controls, G, and a swell pedal, L.

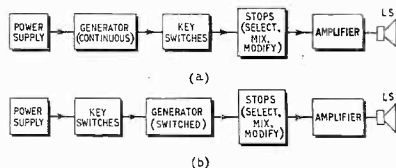
The playing keys on the electronic organ keyboard select the appropriate frequency a.c. signals from the generator, and the stop tabs direct these selected signals into corresponding tone shaping circuits. Subsequently, these keyed, stop-directed, shaped, a.c. signals at H pass into the amplifier, X. There they are amplified and used to produce a sound output by driving a loudspeaker, I. The output level is controlled by an amplifier volume control actuated by the foot pedal, L.

The traditional organ nomenclature of keyboards ("manual" and "pedal"), "stops" (tone colours), and "swell pedal" (volume control) have been carried over into electronic organs, because they achieve the same results, even if in a vastly different way.

Distinguishing mark of the electronic organ.—In essence, while a wind organ emits sound directly from the vibrations of a column of air in a pipe, an electronic organ produces it indirectly from a.c. electrical signals by turning them into vibrations of a loudspeaker cone. (You should not, by the way, confuse the electronic organ with the electric organ in which electrical power is used to drive the mechanical movements).*

TONE GENERATION IN AN ELECTRIC ORGAN

Electronic organs fall into two main classes according to whether their a.c. signal generators run all the time, or are switched on and off by the playing keys. The distinction is illustrated diagrammatically in Fig. 3. In the continuous-generator case at Fig. 3(a), a power supply drives the generators continuously, but their outputs do not pass towards the loudspeaker until a playing key is



depressed; i.e., the keyboard switches come between the generators and the output busbar. In the switched-generator case at Fig. 3(b), the generators are inactive until the depressing of a playing key feeds the power supply into one of them; i.e., the keyboard this time comes between the power supply and the generators.

Generator types in common use.—Anything that will generate stable audio-frequency a.c. signals can be used as an electronic organ generator. The path of development up to date is strewn with weird relics in consequence. Fig. 4 gives a selection of some of the more common tone generator principles that have survived into current commercial organs.

At Fig. 4(a) is a sketch of the principle of the time-proved rotating disc electromagnetic generator. This has been used now for some 30 years in the Hammond organ, which is regarded by many as the "Rolls-Royce" of electronic organs. A rotating iron disc with undulating rim, set up opposite a permanent magnet, gives rise, in a coil wound round the magnet, to a.c. signals at a frequency set by the rotational speed and the number of wave cycles round the rim.

Another common generator type is the wind-blown, electrostatic pick-up, reed system illustrated in Fig. 4(b). Derived from harmonium practice, this has been widely used by such firms as Farfisa. The metal reed is continuously vibrated by a fixed-pressure air supply, and the

*What does one call a pipe organ with a semiconductor logic switching action? (See "W.W.", Jan. 1965, p. 16).—Ed.

capacitance variation between the pick-up electrode and the earthed metal reed modulates the output voltage to give an a.c. signal at the reed resonant frequency.

Fig. 4(c) illustrates a third principle, the rotary-disc, capacitor pick-up system which has been the hallmark of the Compton electronic organs for over three decades. Here the voltage on a fixed pick-up electrode is modulated by the varying capacitance between the electrode and an earthed cyclically changing pattern on the face of a rotating disc. (In the Compton version, actually, the pick-up rotates and the cyclic pattern is on a fixed stator, but the principle is the same.)

A slightly different type of generator is the key-vibrated, electro-static pick-up system illustrated at Fig. 4(d), used in "electronic pianos," such as the Wurlitzer. Strictly it is not an organ generator, because the vibration of the reed dies steadily away after the impact of the key hammer. However, it is included because the electronic piano is in many other ways very like an electronic organ.

These first four examples of organ generators given above have been all electromechanical. There is a tendency for such types to be superseded by purely electrical oscillators, of which common examples are given in Figs. 4(e) to (h). LC oscillators such as Fig. 4(e), valve, and Fig. 4(f), transistor, have their frequency set by their own LC tank circuits. By contrast, the divide-by-two circuits at Fig. 4(g), a transistor triggered blocking oscillator, and Fig. 4(h), a transistor bistable divider, merely

provide an output signal at a frequency one half of the input trigger signal frequency.

A.C. signal waveshapes.—What sort of waveshapes do the various generators produce? This is important because, if, as is sometimes desirable, the instrument has to simulate the sound of a traditional organ, it will have to feed to the loudspeakers highly complex periodic waveforms.

One way of producing such complex a.c. output signals is by Fourier synthesis, i.e., by mixing a harmonically related series of sinewaves in varying proportions. This is the working principle of the Hammond. The basic sinewave shape, as shown in Fig. 5(a), comprises only a fundamental frequency without harmonics, of course. Most electric organs nowadays, however, approach the production of a complex output waveform by taking a basic waveshape rich in harmonics and working on it to tailor the strength of the individual harmonics to give the desired tone colour.

The two commonest harmonic-rich waveforms used in organ generators are the sawtooth, Fig. 5(b), and the squarewave, Fig. 5(c). It is worth noting the harmonic content of these two types. The sawtooth of basic frequency f has material overtones in both the even ($2f, 4f, 6f, \dots$) harmonic series and the odd ($3f, 5f, 7f, \dots$) series. The squarewave has odd harmonics only. The importance of this lies in the fact that the sawtooth with

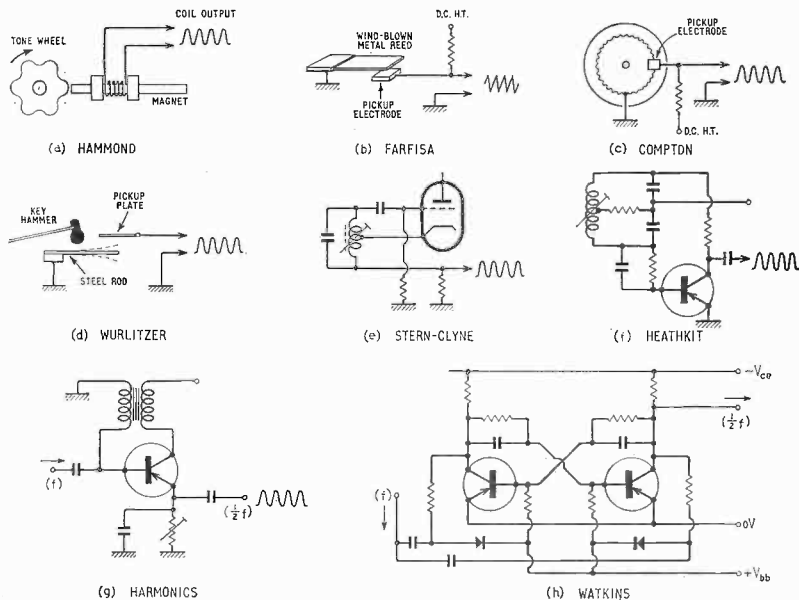


Fig. 4. A selection of some of the more common tone generator principles used in commercial electronic organs: (a) rotating disc, magnetic pick-up; (b) vibrating reed, electrostatic pick-up; (c) rotating disc, electrostatic pick-up; (d) struck rod, electrostatic pick-up; (e) sinusoidal oscillator, valve; (f) sinusoidal oscillator, transistor; (g) blocking oscillator, divide-by-two, transistor; (h) bistable multivibrator, divide-by-two, transistor.

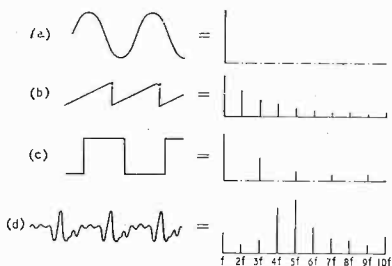


Fig. 5. Indications of the relative strength of the harmonics in basic waveshapes commonly produced by electronic organ generators compared with the complex waveshape required for a typical organ oboe stop: (a) sinewave; (b) sawtooth; (c) squarewave; (d) complex (oboe).

its full range of harmonics can be more easily reshaped to produce the normal "diapason" classical organ sound than can the squarewave. In the latter, the absence of even harmonics makes diapason tone difficult, although it has compensations in that it can more easily produce hollow tones such as the clarinet sound.

Some commercial organs use basic sawtooth and some basic squarewaves. The tendency nowadays is to adopt the squarewave because it is cheaper in production and less "temperamental" than the sawtooth. Some pundits claim that they can tell blindfolded whether an organ uses sawtooth or squarewaves, but with a well-designed instrument I personally have not found this to be so.

For the designer, the choice of sawtooth or squarewave is important, because the tone-forming filter circuits controlled by the stop tabs are substantially different for the two types. This is why any circuit for a tone filter should have on it an indication of the basic waveshape it controls.

To put point to this discussion of generator waveshapes, the complex waveshape of an oboe sound is shown in Fig. 5(d) with a spectrum of its harmonic content. This particular waveshape is characterized by a low level of fundamental and a predominance of fourth and fifth harmonics. You can see that the harmonic distribution of Fig. 4(d) can be relatively easily made up by adjusting the amplitudes of the harmonics of the sawtooth, Fig. 4(b). But, in the case of the squarewave, Fig. 4(c), even harmonics have to be produced separately and added before the desired complex oboe waveform can be achieved.

ACTION OF PLAYING KEYS

In the electronic organ each playing key operates a set of ganged switches, usually four or five in number, to produce different harmonic pitches. As mentioned earlier, these key-switches either select the appropriate outputs from certain of a set of continuously running generators or switch the actuating power to selected oscillators in a set of normally non-operating generators.

Locked- and free-phase oscillators.—The oscillators controlled by the key-switches can all be independently adjustable so that, apart from their being individually tuned as nearly as practicable to the appropriate scale-related note frequencies, they are "free-phase" oscillators working independently of each other. In such a system, initial tuning can be tedious, because anything from 60

to 600 oscillators have to be set up. It is rather like tuning a wind organ (where each pipe has to be checked).

The alternative (and nowadays by far the commonest) arrangement is the "locked-phase" oscillator system. In this there are only 12 independent tunable "master" oscillators, one for each note, (C, C♯, D, D♯, E, F, F♯, G, G♯, A, A♯, B) of the equal-tempered musical chromatic scale. All the frequencies required by the organ are octavely related to these 12 oscillator frequencies, and are produced by a string of divider oscillators phase-locked to the master oscillators (usually to subharmonics) and thus not independently adjustable.

The free-phase and locked-phase oscillator systems are illustrated diagrammatically in Fig. 6 for all the Cs to be sounded by an organ. In the locked-phase system of Fig. 6(a), the master C oscillator is tuned to C6=2,093.0 c/s; the Cs for lower octaves are produced by a string of divide-by-two circuits phase-locked to the master oscillator. This means that C1=65.4 c/s, for example, comes out coldly mathematically at an exact half of C2=130.8 c/s, and in phase with it. In the free-phase system of Fig. 6(b), each oscillator in the string of Cs is individually tuned as near as possible to the harmonic series frequencies, but each C will inevitably depart slightly from half of the frequency of the C above it. Also the individual oscillations will not now be in phase with each other.

There is no doubt that the free-phase oscillator system can more closely simulate the so-called "chorus" effect of a wind organ with all its pipes individually tuned and not in phase with each other. But you must pay for perfection, and many electronic organs now make do with the cheaper, easier-to-set-up, locked-phase divider system.

The problem of key clicks.—Whether locked- or free-phase oscillators are used has a large bearing on the keying system. Obviously in a locked-phase system the oscillators must run all the time and the keys can only switch their outputs. This means switching a.c. signals, in which it is very difficult to avoid "key-clicks" arising from the abrupt interruptions of the signal. With free-phase oscillators, on the other hand, it is possible to use the keys to switch the d.c. supply to the oscillators. The oscillator output a.c. signals then do not start and stop abruptly, and key clicks are much less of a problem. Also in this case the gradual start of oscillation stimulates the steady build-up of sound in a wind organ pipe when the playing key is pressed.

In the electronic organ, the "stops" are merely the switches for selecting the tone forming filter circuits through which the basic waveform signals are fed from the generators via the playing keys to achieve the desired final tone colour. There are three basic classes of stops: "speaking," "percussion," and "coupler."

Speaking stops are the switches that control the tone colour of the organ output. They select certain pitches of notes supplied by the playing keys, mix them and feed them to appropriate waveshape-modifying filter circuits. These stops are concerned primarily with the recurrent waveshape of the final signal to the loudspeakers, i.e., the steady, continuous, fixed-amplitude, fixed-frequency waveshapes.

Percussion stops.—Other switches, the so-called "percussion stops," are usually provided to control the envelope of the output signals. If the playing key switched the signal on and off abruptly, we should get a signal

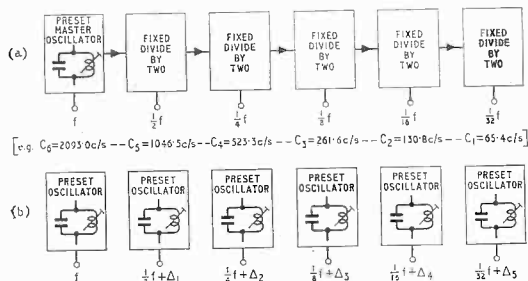


Fig. 6. Illustration of differences between locked-phase (a) and free-phase (b) oscillator systems in electronic organ. (The locked-phase system is also known as "divider" type or "synchronous" system.)

envelope like Fig. 7(a). In an attempt to simulate classical organ sound and to give variety to tonal effects, electronic organ designers use some or all of the variety of percussion (or envelope-distortion) arrangements illustrated in Figs. 7(b) to (g).

"Attack" percussion at Fig. 7(b) is the delaying of the build-up of the waveform signals on switch-on. "Decay" at Fig. 7(c) is the permitting of signals to fall off (or even cease) before the playing key is released. "Sustain" at Fig. 7(d) is letting the signal continue after key off. In all these types, the amplitude and frequency of the note are unchanged in the middle of the note sounding.

But there are other percussion stops which modify the waveshape characteristic throughout the whole note length. Of this class is the "tremolo" stop at Fig. 7(e) which varies the signal amplitude at a subsonic

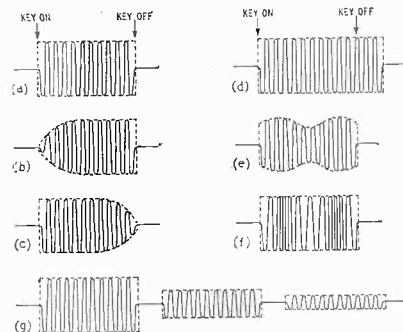


Fig. 7. Modifications of fixed-amplitude, fixed-frequency, waveshape to achieve various special "percussion" effects: (a) No percussion— instant start, instant stop, constant-amplitude, constant-frequency. (b) "Attack"—varying rate of increase of amplitude on switch on. (c) "Decay"—amplitude falling off before key off. (d) "Sustain"— signal continues after key off. (e) "Tremolo"—signal frequency constant but amplitude varied at low frequency (amplitude modulation about 6 c/s). (f) "Vibrato"—signal amplitude constant but signal frequency varies at low frequency (frequency modulation about 6 c/s). (g) "Reverberation"—signal repeated one or more times after key off, with progressively reduced amplitude.

frequency between 3 and 9 c/s to make an approximation to the "tremulant" effect in a wind organ. "Vibrato" at Fig. 7(f) frequency-modulates the signal at subsonic frequency again in an attempt to simulate the organ tremulant. But in the wind organ, the tremulant is a mixture of amplitude and frequency modulation, and electronic organ designers sometimes provide both tremolo and vibrato in consequence. (People who have difficulty in remembering which is vibrato and which tremolo may welcome my own little *aide-memoire*: Vibrato varies the Frequency, because "V" sounds not unlike "F.")

The last percussion stop is "reverberation" (also known as "echo") illustrated in Fig. 7(g). In a wind organ sited in a church or a large hall, part of the "organ" sound is due to the characteristic echoes from the walls. In effect, a tone burst is heard repeated one or more times with diminishing volume after the key is released. This is simulated in an electronic organ by feeding off part of the signal into a delay network and then releasing it back into the amplifier. The delay network can be acoustic (some people have even used a length of garden hose), mechanical (e.g. passing sound through a coiled spring), or moving-magnetic (e.g. signal recorded on a magnetic tape and picked up again later by a separate playback head).

Coupler stops.—Where an electronic organ has more than one keyboard, "coupler" stops are often provided to enable the separate keyboards to be coupled; i.e., speaking stops can be set up for one keyboard and played from another by switching in coupler stops.

COMPLETE ELECTRONIC ORGANS

If you now look back to the schematic block diagram of an electronic organ given in Fig. 2(b), you will see that we have now covered, albeit sketchily, most of the component sections of the system. But a few further remarks might not be out of place here.

The power supply is usually mains driven, but transistor models are available nowadays for battery operation. The amplifier is a little different from a standard hi-fi amplifier. Because of the possibility of sustained high level signals, its overload characteristics require careful design. Output powers range from 10 W to 100 W usually. The loudspeakers must not only have adequate wattage ratings, but must have a good response down to below 20 c/s to handle low pedal notes without distortion. The swell pedal (volume control) receives

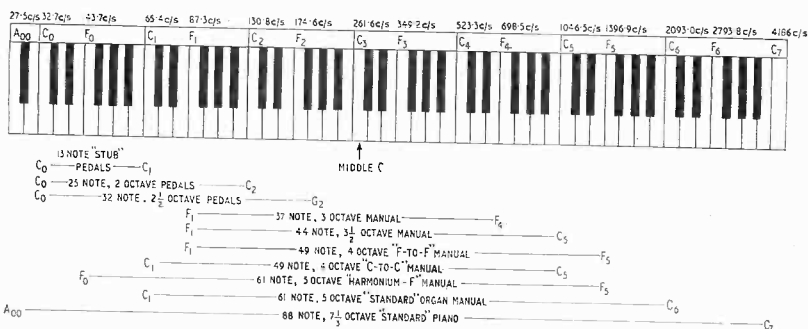


Fig. 8. Compasses of keyboards in various commercial organs compared with standard piano keyboard.

much hard wear. For this, therefore, often some arrangement of light- or voltage-dependent resistors is used to avoid using a conventional carbon or wirewound potentiometer which may soon develop contact noise.

Consoles.—A variety of layouts of the component sections of the electronic organ will be found in commercial models. The tendency nowadays is to pack everything into a single cabinet or "console." This was not always possible in earlier days, but modern miniature components make the self-contained organ the norm. The domestic electronic organ sketched at Fig. 1(b) illustrates this compactness tendency. Here you can see the stop tabs sited above the keyboard (their usual position), supplementary controls at each end, the swell pedal convenient to the player's right foot, the loud-speaker facing forward, and the pedal board at the bottom left.

Keyboards.—As to the keyboard itself, the student of the electronic organ will find a confusing variety in use. Commercial instruments will, of course, have at least one manual keyboard. More frequently they have two manuals, usually called the solo (upper) and accompaniment (lower). These are also occasionally referred to as the "swell" and "great" respectively to conform with traditional wind organ terms.

Where an electronic organ is intended to substitute for a traditional organ, three manuals are not uncommon, and these are usually given the standard organ descriptions of: "great," "swell" and "choir."

In the "pedal department," two arrangements have become fairly standard in the electronic organ. The first of these (used in the illustration of Fig. 1(b)) is the 13-note, stub-pedal common in the smaller domestic organs. The standard full 31-note classical pedal compass is not often used for electronic organs, although many models feature the second standard, a shorter two-octave, 25-note version.

When engineers first take an interest in electronic organs, they often find themselves confused by the plethora of keyboard specifications. For their benefit, Fig. 8 sets out the compasses of widely used keyboards against the spread of the familiar standard 88-note piano keyboard. The standard organ manual keyboard is the C₁-C₅, 5-octave, 61-note version, but you may come

across the alternative "harmonium, F-to-F" version with the same compass. The shorter manual keyboards listed in Fig. 8 have become very common in small domestic electronic organs, mainly because of their cheapness.

SOURCES OF FURTHER INFORMATION

By now the reader must have realized that electronic organ design is not without its complexities. If all this has not discouraged him too much, he may welcome suggestions as to where to look for further information.

For basic principles and general reading, I can heartily recommend "Electronic Musical Instrument Manual," Pittman, 1961, and "Electrical Production of Music," Macdonald, 1957, both by A. Douglas. No textbooks exist covering current practice in this rapidly changing field, but some useful information can be gleaned from such books as "Electronic Organ Handbook," H. E. Anderson, Sams, 1962, and "Servicing Electronic Organs," C. R. Pittman and E. J. Oliver, Sams, 1962.

And for the "antiquarian," interested rather in the development of the electronic organ, recommended reading "musts" are "Electronic Music and Instruments" by B. J. Miessner, *Proc.I.R.E.*, Vol. 24, No. 11, Nov., 1936, and "Electronic Musical Instruments and the Development of the Pipeless Organ" by G. T. Winch and A. M. Midgley, *Journal I.E.E.*, Vol. 86, No. 522, June, 1940.

OUR NEXT ISSUE

A 32-page supplement in our June issue will give a stand-by-stand preview of the Instruments, Electronics & Automation (I.E.A.) Exhibition to be held at Olympia from May 23rd to 28th. This supplement will be in addition to the normal quota of features and technical articles which will include the second of the series on electronic organs and a review of the latest developments in audio techniques seen at the London Audio Fair. We also plan to include a report on some of the new equipment seen at the Hanover Fair (April 30th-May 8th).

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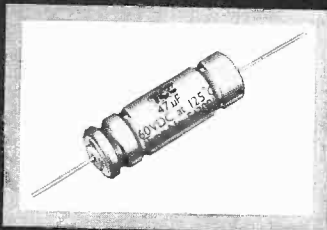


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WORLD OF WIRELESS

Aircraft Collision Avoidance System Proposed

IN 1955 the Air Transport Association of America asked the electronics industry to submit proposals for a form of collision avoidance equipment. Since then much work has been devoted to the development of a system suitable for installation in civil and military aircraft and compatible with the American air traffic control system, but to date a system has not been fully approved. However, one manufacturer, McDonnell Aircraft, of St. Louis, Missouri, has developed a system for use in specialized flight test operations with its aircraft. The system uses the time frequency, range altitude technique which entails the measurement of range, the rate of change in range (range-rate) and altitude. Each aircraft carries a computer which has a precise transmit time assigned to it; each computer also knows the transmit time of the other aircraft. All aircraft transmit their altitude on a common frequency at the allocated times and the difference in μ s, between the assigned time of transmission and the actual time of reception by the computer is used to determine distance to the transmitting aircraft. At the

receiving aircraft a frequency shift varying with the relative speeds of the two aircraft is measured to give their rate of closure. The computer divides distance by rate of closure to predict possible time of collision, advises what escape manoeuvre to use and then indicates when the potential collision has been avoided. Great precision in both timing and frequency stability is necessary. Accuracies in the order of one part in ten million are required. Each aircraft must carry an accurate frequency standard and the system requires a practical method for synchronizing the timing devices carried aboard all aircraft at appropriate intervals.

In its present form, the McDonnell system is not suitable for airline use, but the Air Traffic Control Committee of the A.T.A. has been directed to give all possible assistance to the company so that the system can be subjected to early flight evaluation by airlines.

Collision avoidance systems for aircraft are being discussed in our correspondence columns (see page 249).

Television Without Cameras

THE Canadian RCA Victor Company has developed a technique by which printed matter can be televised without the use of a camera; instead, for example, a typist operating a keyboard can type messages for direct television transmission.

The system, designated DIVCON—digital to video conversion—includes a display control logic sub-system programmed to store information from various sub-systems and to "read" the contents of a memory to a series of digital-to-video-generator sub-systems which translate the digital data into a video form synchronized with the television scanning periods. The system has all control functions normally found on an electric typewriter such as space,

shift and backspace and there are special facilities for erasing. There is also a facility to underline words or to make them flash. Stationary or travelling displays of information are also possible. Code conversion logic is provided so that the DIVCON system can work with data processing centres using different machine languages.

Data is stored in the random access core memory in an eight bit code and the actual size of the memory is dictated by the number of output video channels wanted and the amount of character symbols needed for each channel.

Readers who watched the B.B.C.'s televised results of the General Election will be interested to learn that the DIVCON system was used for these transmissions.

International Amateur Radio Conference

FROM May 23-28 at Opatjia in Yugoslavia about 70 delegates representing some 40,000 licensed radio amateurs from the national amateur radio societies of Austria, Belgium, Finland, France, Germany, Italy, Luxembourg, the Netherlands, Nigeria, Norway, Poland, the Soviet Union, Sweden, Switzerland, the United Kingdom and Yugoslavia will meet to discuss over 100 proposals and suggestions at the triennial conference of the International Amateur Radio Union, Region 1 (Europe and Africa) Division.

Subjects for discussion will include amateur teletype standards and frequencies, proposals for a European orbiting

satellite carrying amateur radio (OSCAR) and proposals by the R.S.G.B. that harmonically related frequencies should be used for s.h.f. working and authority sought for amateurs to use the amateur u.h.f. and s.h.f. bands for space communication. At present under international regulations amateur space communication activities are confined to the 144-146 Mc/s band.

The Conference is being organized by the honorary secretary of the I.A.R.U. Region 1 Division, John Claricoats, G6CL, who, until he retired at the end of 1963, had been secretary of the R.S.G.B. since 1930.

Australian Telecommunication Problems

THE Australian Telecommunication Development Association, a representative body for manufacturers of telecommunication equipment, is confronted with a major problem, for although the Post Office supports the indigenous electronics industry, the Defence Services do not. Last year, 80% of all radio transmission equipment purchased by the Commonwealth Department was imported. The Association admitted that some complex electronic devices could only be obtained abroad, but also claimed that the Australian industry could have supplied 50% of the required equipment. The prob-

lem is twofold; first, that since this industry is regarded as a major unit in the Australian economy a continued flow of orders overseas can only be detrimental to the economy; and, more seriously, A.T.D.A. maintain that with dependence on other countries for the supply of electronic defence systems, Australia could suffer if hostilities cut supply lines.

Objectives of A.T.D.A., prompted by the above problems, include (a) to encourage and maintain the growth of the industry and (b) to sustain effective co-operation with relevant Government departments and industry.

E.E.A. Annual Report

THE Electronic Engineering Association has recently issued its annual report which includes details of the general activities and work of the individual divisions during 1965.

One of the objectives of the Association has been to establish close liaison with the Ministry of Technology whose main function has been to implement the Government's programme of modernizing the economy and to maintain close collaboration with industry. However, the report indicates that the general economic inhibitions and cancellation of military aviation projects have created disturbing uncertainties for the whole of the electronic capital equipment industry.

During the year the Association has conducted a detailed

Canvas and Electronics.—Modern electronic equipment is a feature of a 300-ton topsail schooner, *Sir Winston Churchill*, which sailed recently on her maiden voyage. Equipment includes Marconi Marine "Raymarc" radar, presented by English Electric Ltd., a Marconi Marine "Kestrel II" transmitter-receiver provided with a radio goniometer and presented by The Marconi Company Ltd., a "Graphette" recording echo sounder and an octahedral radar reflector, presented by the Marconi International Marine Co. Ltd. The radar reflector has been fitted because sailing vessels are not always good radar targets; improved reflections will ensure that the schooner can be under radar observation by other vessels. The schooner is owned by the Sail Training Association.

Thirteen new relay stations will be built during the next three years by the Independent Television Authority. These stations will bring a new or improved service to over half a million people along the western side of the United Kingdom. All sites and channels have not been decided but provisional information indicates that three stations will be built along the Scottish coast to cover Central Ayrshire, Helensburgh and Rothesay and six small stations in Central Wales will cover the main towns of Abergavenny, Bala, Brecon, Ffestiniog, Llandrindod Wells and Llandoverly. Another station will cover Whitehaven (Cumberland) and, in the West of England, a station will be built between Hereford and Gloucester, another near Bath and one near Barnstaple and Bideford.

The International Publishing Corporation, of which the Iliffe group is a member, has acquired a 40% interest in Cahners Publishing Company, the Boston, Mass., publishers of 17 trade and technical journals. Among the journals published by Cahners is the monthly *Electronic Engineers' Design Magazine*, which has a controlled circulation of nearly 52,000. The latest acquisition brings the total number of trade and technical journals published by I.P.C. associated companies in this country, France, Italy, West Germany, Holland, the Far East and America to 213.

Yugoslav Post Office will issue a special 0.85 dinar stamp on May 23rd to commemorate twenty years of Yugoslav amateur radio, and the opening of the Region I (Europe and Africa) Conference of the International Amateur Radio Union in Opatija. Further publicity will be provided by the daily operation of the station YU0IARU, from the headquarters of the Yugoslav National Amateur Radio Society. Transmissions will be from 1200 to 1300 G.M.T. on 14.1 Mc/s.

Digital Instrumentation is the title of a one-day symposium to be held on May 19th at the Mid-Essex Technical College and School of Art, Victoria Road South, Chelmsford. Subjects to be covered include circuitry, displays, ergonomics, digital counters, measurement techniques and methods of digitalization. Further details are available from the Department of Electrical Engineering.

analysis regarding the expansion potential for production of radio communications, electronic control, navigational aid, aviation and broadcasting equipment. This has resulted in a document "Electronics and the Economy" which has been submitted to the Ministry of Technology and other Government departments in an attempt to show that if certain Government action were taken the total output and exports of the various analysed sectors of the industry could increase and the imports of manufactured equipment reduced. Detailed discussions with the Ministry of Technology on the document have not yet taken place but it has already provided a background to various committees and enquiries instituted by the Government.

Royal Society Awards.—Three awards of unstated value under the Royal Society and Nuffield Foundation Commonwealth Bursaries Scheme have been made. To Dr. A. R. H. Cole (University of Western Australia) to enable him to gain experience in high resolution electronic spectroscopy; to Professor E. O. Hall (University of Newcastle, New South Wales) to enable him to gain experience in the use of thin film electron microscopy and to Professor E. R. Love (University of Melbourne) to assist him to continue work on integral equations and integral transforms. An award under the Paul Instrument Fund Committee worth £3,250 has been made to Professor A. F. Gibson (University of Essex) for work on the application of infra-red lasers.

A visit of a "Spaceobile" to this country is being arranged by the British Interplanetary Society and the National Aeronautics and Space Administration of America. The "Spaceobile" is a vehicle containing models, films and other visual aid equipment for demonstrating astronautics and will be on a three-month tour of the U.K. Schools, colleges, universities, adult education centres, industrial, service establishments and other organizations who wish to be included in the tour should contact the Society at 12 Bessborough Gardens, London, S.W.1.

The second session of the **Aeronautical Radio Conference** convened by the I.T.U. commenced on March 14th for a period of eight weeks at the *Maison des Congrès* in Geneva. The main purpose of the Conference is to draw up a new world-wide plan for h.f. long-range communications for aircraft flying on national and international civil air routes. This will be considered in conjunction with the development of v.h.f. communications to reduce the use of high frequencies for short-range communication in many parts of the world.

Radio Nepal.—A gift of a 100 kW Marconi h.f. transmitter, costing with associated equipment about £145,000, is being made to the Nepalese government by the Ministry of Overseas Development as part of the £1M capital grant made by the British Government to Nepal in 1960. With the new transmitter, Radio Nepal will be receivable in all parts of Nepal.

This year's **National Mobile Rally** organized by the Radio Society of Great Britain will be held on Sunday, May 1st, at the Bedford headquarters of Texas Instruments Ltd. The talk-in stations will be GB3BS on 160 and 180 meters s.s.b. and GB2VHF on 2 and 4 metres.

An exhibition of educational equipment arranged by Educational Instruments Ltd., is being held on May 9th and 10th at the Embassy House Hotel, 31 Queens Gate, London, S.W.7. Admission to the exhibition is free to all those interested in educational equipment.

PERSONALITIES

Joseph W. King, Ph.D., 35-year-old physicist at the Radio and Space Research Station, Slough, Bucks, has received the 1965 Wolfe award for his work on "the analysis of results obtained from topside sounding satellites and their interpretation in terms of ionosphere theory; work which has enhanced his own reputation and that of the station." The £500 award is the eighth of ten annual awards to be made under the will of the late James Perrin Wolfe to the research worker who is considered to have made an outstanding contribution to the work of what was the Department of Scientific and Industrial Research, now the Science Research Council. Dr. King, who joined

Kenneth G. Budden, M.A., Ph.D., A.M.I.E.E., lecturer in physics at the Cavendish Laboratory, Cambridge, has been elected a Fellow of the Royal Society. The citation refers to Dr. Budden's "contributions to the theory of the propagation of electromagnetic waves in a non-uniform anisotropic medium, particularly of long radio waves in the ionosphere." A few years ago he was given a grant of £1,694 by the then D.S.I.R. for studying the theory of propagation of radiations from artificial satellites.

Professor James D. McGee, O.B.E., M.Sc., Ph.D., M.I.E.E., who is Professor of Applied Physics at Imperial College, London, is also among the newly elected Fellows of the Royal Society. Dr. McGee, who is 63 and a graduate of Sydney University, was a research physicist at E.M.I. Research Laboratories from 1932 until joining the academic staff at Imperial College in 1954. Reference is made in the Fellowship citation to Dr. McGee's work on the development of electron-optical instrumentation, television camera tubes, infra-red image converters and image intensifiers.

Sir Frederick White, K.B.E., chairman of the Commonwealth Scientific and Industrial Research Organization, Canberra, Australia, is also appointed an F.R.S. for his work in radio physics and radar and "for outstanding contributions to the development of Australian science."

C. H. Noton, M.Sc., Ph.D., M.I.E.E., who has been with the Plessey organization since 1959 and has latterly been resident general manager for France, has been appointed to the new post of group commercial executive of the Plessey Components Group. Dr. Noton, who studied at the Universities of Liverpool and Oxford, served at the Ministry of Economic Warfare during the second world war. He was in the chemical industry before joining Plessey.

A number of other executive appointments in the Plessey Components Group are also announced. The new group quality control manager is **E. P. Stanton, Ph.D., M.I.E.E.**, who has been with the company since 1957 and was chief inspector of Plessey-U.K. prior to his new appointment. Dr. Stanton, who is 56, previously held posts in the G.P.O. Engineering Department and was for 18 years in the Royal Signals. **R. H. Crampton**, who is 44 and has been with the company 27 years and has been regional production controller since 1964, has become production and

cost control manager. **H. P. Kavanagh**, who is 41, is appointed group data production manager. He has been with Plessey for two years. **J. A. Lee**, chief engineer of two Plessey divisions successively between 1958 and 1965, becomes group mechanisation engineer. He joined Plessey 25 years ago at the age of 20. **L. W. D. Sharp, M.I.E.E.**, who came to Plessey from E. K. Cole Ltd. in 1948 at the age of 26 and has held various divisional and group appointments, is appointed manager, new product development. He was a development engineer with E. K. Cole.

Lord Nelson of Stafford, chairman and chief executive of the English Electric Company, has been elected president of the British Electrical and Allied Manufacturers' Association. Lord Nelson, who succeeded to the barony on the death of his father in 1962, is a member of the National Electronics Research Council and of the Minister of Technology's Advisory Council on Technology and is also chancellor-designate of the new University of Aston, Birmingham.

H. S. Macadie, who in 1932 designed the Universal Avometer, has retired from the board of Avo Ltd. but is being retained by the company as a consultant. Mr. Macadie, who is 65, served his apprenticeship in mechanical engineering and then joined B.T.H., where he worked on meter production. In 1925 he started his own company to manufacture plug-in tuning coils for broadcast receivers. The coils were marketed



Dr. J. W. King

the Radio and Space Research Station in 1961, is a graduate of Rhodes University, South Africa. In 1951 he was awarded the Cornwall and York Prize by the Vice-Chancellors' Committee of the South African Universities for his first publication on ionospheric physics. He went to Emmanuel College, Cambridge, on a Rotary Foundation Fellowship for Advanced Study and was awarded the 1955 Hamilton Prize by Cambridge University for the research in ionospheric physics which he did at the Cavendish Laboratory.

C. C. Moore, who has been with the Bush organization for 35 years and for the past year has been managing director of Rank Bush Murphy Ltd., has been appointed to the newly created post of vice-chairman. The chairman is **John Davis**. Mr. Moore, who is 67, was appointed assistant managing director of Bush Radio in 1952 and held the same position in the Bush-Murphy merger. He succeeded to the managing directorship when **Dudley Saward** resigned in April last year. **J. P. Colfils**, who was named managing director designate several months ago, succeeds Mr. Moore. Mr. Colfils will continue also as managing director of Rank Audio Visual.



H. S. Macadie

by the Automatic Coil Winder & Electrical Company, which also marketed the d.c. Avometer and an automatic coil winding machine both invented by Mr. Macadie's father. In 1927 Mr. Macadie, jr., joined the Automatic Coil Winder Company; the name of which was changed in 1957 to Avo and he became a director.

Group Captain E. Fennessy, C.B.E., director of the Plessey Electronics Group, has also been appointed chairman of the reconstituted board of British Telecommunications Research Ltd. The company was established in



J. Lawton

1946 jointly by A.T. & E. and B.I.C.C. and became a subsidiary of Plessey in 1961 with the acquisition of A.T. & E. A minority interest is retained by B.I.C.C. Under the reorganization J. Lawton, M.Sc., M.I.E.E., has become director of research. Mr. Lawton, who took an honours degree at University College, London, was in the Post Office Research Department from 1939 until joining B.T.R. in 1950. He became a director in 1956 and since 1962 has been responsible for the management of the company.

G. B. B. Chaplin, M.Sc., Ph.D., M.I.E.E., chief scientist at Plessey-U.K. Ltd., has been appointed to the chair of electrical engineering science at the University of Essex. Dr. Chaplin, who is 42, graduated at Manchester University, where he subsequently served for three years as lecturer in electrical engineering. From 1955 he was for four years in the Electronics Division at A.F.R.E., Harwell. He joined Plessey in 1959 as technical manager of their Rokke Manor Research Laboratory.

Captain F. J. Wylie, director of the Chamber of Shipping's Radio Advisory Service since it was established in 1948, has retired and joined Marconi International Marine Company as a consultant. Capt. Wylie's Naval appointments included Fleet Wireless Officer Mediterranean, 1929-31; officer in charge of the experimental department of H.M. Signal School, 1935-37; deputy director, Admiralty Signal Department, 1941-43; and director of radio equipment, Admiralty, 1944-46. The Radio Advisory Service was established to give authoritative advice on all aspects of marine radio communications and navigation and to investigate and evaluate all types of marine electronic equipment for the benefit of the members of the Chamber of Shipping and of the Liverpool Steam Ship Owners' Asso-

ciation. Capt. Wylie is succeeded as director of the Chamber of Shipping's Radio Advisory Service by Captain R. G. Swallow who has been deputy director since his retirement from the Navy in 1954. Capt. Swallow, who was a communications specialist throughout his Naval career, was successively deputy director of the Admiralty's radio equipment department and of the electrical department and director of the Signal Division. From 1952 until his retirement in 1954 Capt. Swallow was director of the Tactical School, Woolwich.

J. M. C. Dukes, M.A., D.I.C., A.M.I.E.E., who six months ago joined A. C. Cossor Ltd. as technical director, has been appointed deputy managing director. Until joining Cossor last year Mr. Dukes had been for six years with the Plessey Company first as technical manager of the Telecommunications Division and since 1962 as chief engineer of the Electronic and Equipment Group. Prior to joining Plessey he was for 12 years with Standard Telephones & Cables.

G. E. Gilbert, A.M.I.E.E., who has been in charge of the B.B.C.'s engineering buying section since 1936 is retiring on April 30th. He joined the Corporation in 1935 having previously served with Marconi's and the Western Electric Company. The new engineering buyer is T. R. Boys who joined the B.B.C. in 1939 at the Daventry transmitting station. He later served in the departments concerned with research, planning and design and since 1954, has been assistant engineering buyer.

Professor Charles W. Oatley, O.B.E., M.A., M.I.E.E., who has occupied the Chair of Electrical Engineering at Cambridge University since 1960, has been appointed to the board of directors of English Electric Valve Company. For 12 years before the war he was a member of the staff of the Physics Department, King's College, London University. For some time during the war he was in charge of basic work on radar transmitters and receivers at the



Prof. C. W. Oatley

Radar Research and Development Establishment of the Ministry of Supply. Since 1945 he has been a Fellow of Trinity College, Cambridge, and was lecturer in electrical engineering prior to receiving his professorship.

Reginald J. Watson, works manager of Mullard's Southampton factory since 1957, has joined Standard Telephones & Cables (Transistors) Ltd. the newly formed subsidiary of S.T.C. and has been appointed director of operations for the E.F.T.A. area of the semi-



R. J. Watson

conductor activities of the I.T.T. group. Mr. Watson will be at the Footscray, Kent, headquarters of the new company which will co-ordinate the semiconductor operations at both Footscray and Harlow (Essex).

OBITUARY

Major-General Sir Leslie Phillips, K.B.E., C.B., M.C., director of signals at the War Office from 1943 to 1946, died on March 19th aged 74. Sir Leslie, who was knighted on his retirement in 1946, was throughout his 35 years' Army career a signals officer. He was promoted to brigadier and appointed Commandant of the Signal Training Centre in Jubbulporc, India, in 1939 but was recalled to this country in 1940 to become command signals officer, Eastern Command in London and as such he was largely responsible for the provision of communications for this country's defence. For two years prior to his War Office appointment in 1943 he was signals officer-in-command Home Forces.

Leonard A. Norman, who had been with Philips Electrical Ltd. for 37 years, died on March 17th after a long illness. After war service in the R.A.F. he rejoined the company in 1946. Six years ago he was appointed sales manager of Stella Radio and Television Company and later rejoined the parent company as a field manager, southern region.

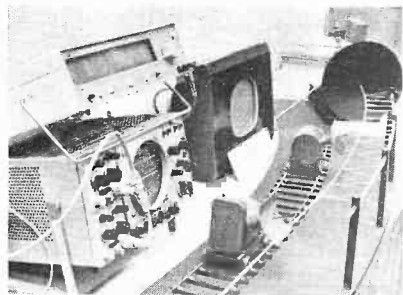
Physics Exhibition 1966

ELECTRONIC AND MEASURING TECHNIQUES UNDER DEVELOPMENT

THE Institute of Physics and the Physical Society held this year's exhibition of scientific instruments and apparatus (the 50th) among the decaying Victorian splendours of the Alexandra Palace in North London—where at least there was plenty of room to move around. The scientific character of the event was well maintained by careful selection of exhibits, and in the following report we have picked out a number of items which will be of interest to people in electronics and communications.

MICROWAVES

Railway communications. An exhibit by the British Railways Board demonstrated the possibility of communication with a moving vehicle for the exchange of information. The exhibit comprised a length of Goubau line fed via a conical launcher at a frequency of 3 Gc/s and terminated similarly, the line supporting a travelling surface wave. A small probe attached to a cavity resonator (see illustration) is brought into the field of the surface



An exhibit illustrating how communication with moving trains can be achieved using guided surface waves of about 3 Gc/s.

wave (the resonant frequency incidentally being varied by a variable-capacitance diode) and it was shown that variations in resonance produced corresponding variations in the echo power at the transmitter. If the line is continuous over long distances then the wide bandwidth possibilities of the line could also be exploited for point-to-point communications.

Gunn oscillators. Semiconductor sources of microwave radiation relying on the Gunn effect have received a considerable amount of publicity recently but their development still continues.

Such devices were shown by a number of exhibitors, but the first public demonstration of their potential was given by Dr. C. Hilsam at the R.R.E. stand.

A Doppler radar system was shown in which the speed and time of travel of a model locomotive were measured. The minute Gunn oscillator replaced the usual klystron and operated at 8 Gc/s (X-band) with about 5 mW of power. The outstanding point of the exhibit was the size of the radar equipment which could be held in the hand.

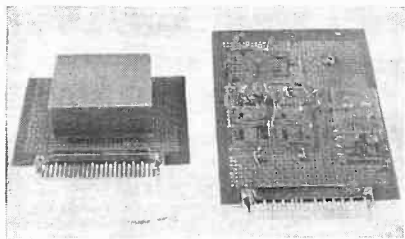
The gallium arsenide device is prepared on an epitaxial layer of GaAs 10 μ m thick and of 1 Ω cm resistivity, which is grown on to a GaAs substrate 0.2 mm thick and 0.001 Ω cm in resistivity. The substrate is alloyed to a copper heat dissipator, this being the positive contact, and a small tin spot is fused on to the epitaxial layer to form the negative contact. The device operates at 12 V and 100 mA delivering about 5-20 mW of power. The frequency is stable to within 1 Mc/s per day and can be adjusted over the range $\pm 10\%$ by cavity mounting.

CIRCUIT TECHNIQUES

Circuit construction techniques of special interest included a new screen-printed film method based on thin mica wafers, introduced by Johnson Matthey. On display was a range of 0.6-in square mica wafers, some with films of resistive material fired on to one side and others with capacitor plates silvered on to both sides. Round the periphery of each wafer are 16 regularly spaced conducting tabs, to some of which the R or G elements are connected. To build up a complete passive circuit the wafers are stacked on top of each other, the complete stack is fired to fuse the layers together, and the edges of the conducting tabs at the sides of the stack are soldered together as appropriate to form required interconnections. Other passive or active elements can be soldered to the stacks, which can be encapsulated in various resins. Some prototype circuits, which were, of course, a good deal smaller than conventionally constructed circuits, were shown in operation.

A high-density method of mounting and interconnecting monolithic integrated logic circuits was displayed by I.C.T. Unencapsulated semiconductor chips, each containing one or more gate circuits, are mounted on a ceramic substrate, and this carries a thin-film multi-layer conductor network by which all connections to the chips are made. The power and earth conductor patterns are standard for a given chip. Superimposed on these, and separated by an insulating layer, is the logic "wiring," consisting of two sets of orthogonal conductors. The insulation separating the conducting strips is present only within the cross-over areas, so that interconnection between the sets of conductors can be made at any cross-over by depositing a metal link. Fifty integrated circuits can be accommodated on a 2in \times 1in substrate. The complete units are intended to be encapsulated and mounted in flat packs.

RC time-varying networks. Frequency selective networks without inductors are obviously desirable for use



An example of frequency selective RC time-varying networks. The thin-film unit shown is a single sideband demodulator developed by A.E.I.

in integrated circuits. A type of active filter (discussed briefly in *W.W.*, March 1966, p. 129) of the time-varying variety is under investigation at the A.E.I. Research Laboratories at Blackheath. One disadvantage of these networks is the high sensitivity to changes in component values and much effort is directed in reducing this.

These networks involve the use of resistors and capacitors and time-varying elements such as modulators or sampling switches. Apart from the more usual type where low-frequency low pass filters are simulated, work is also directed at developing all-pass phase-shift networks used to produce a highly selective filter characteristic at a high frequency.

Single sideband demodulators were exhibited, one in thin film form (illustrated). In these the signal is fed to two modulators driven by carrier signals in phase-quadrature and derived from a carrier phase shift circuit. The modulator output signals are fed to a wide band 90° phase difference network and then added. The output is filtered to remove signal components and then amplified. The carrier phase shift is derived digitally from two bistable circuits thus permitting the carrier frequency to be varied and giving a filter characteristic which can be translated to any position of the frequency spectrum.

LASERS AND ELECTRO-OPTICS

Hologram reconstruction was demonstrated by the Marconi Company, who are one of the many firms investigating applications. Holography relies on a method of optical production which does not use lenses and which was described in 1948 by Dr. D. G. Gabor. The outstanding feature of holography, however, is not that lenses are not required but that three-dimensional images can be formed.

Light from an object illuminated by laser light is allowed to fall on a high quality photographic plate without the aid of a lens so that a focused image is not formed. Light directly from the laser source is also allowed to fall on the plate and as a consequence interference fringes are formed, in a similar fashion to the fringes of the classic Lloyd's mirror demonstration, which are recorded on the plate. Thus the amplitude and phase of the object light is recorded, the fringe pattern being modulated in accordance with the nature of the object. The image is viewed by looking through the plate when this is illuminated by laser light.

One of the interesting features of the hologram is that if the plate is broken, the image can be reconstructed

from only a small piece of the original, but with some loss of detail. This basic process may well find application in a wide variety of fields, much as the laser has done. Developments are still taking place and one of the latest is that it is now possible to view a coloured 3-D image with white light.

Metrology by holography has been investigated by the N.P.L. and A.W.R.E. and was demonstrated on the British National Committee for High Speed Photography stand. It is suited particularly to the detection and measurement of small changes in surface contour for rough surfaces.

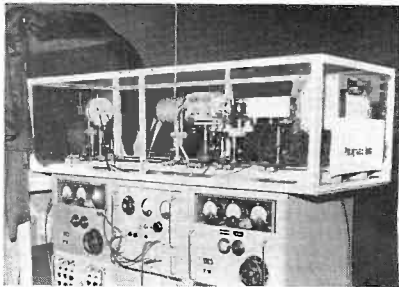
A hologram is made of an object (see previous item), but instead of viewing the image with the object removed, the object is left in place, resulting in the image being superimposed on the object. If the surface of the object then moves slightly interference takes place between corresponding points on the object and its reconstruction giving rise to fringes across the object surface. The degree of distortion can be calculated from the fringe pattern.

If the plate is exposed twice, that is once with the object in its original position and the second exposure after the object has been distorted, then a permanent record of the fringe pattern is made.

The technique can be used to compare production line components for shape against a master; for determining the thermal expansion of components; and for investigating surface vibration modes.

Laser light is not necessary for image formation; the demonstration used filtered mercury light.

Radar photography allowing clear pictures to be taken, or distances of objects to be determined, through a light-scattering medium such as fog or dust was demonstrated by the Atomic Weapons Research Establishment (operating in their new "swords into ploughshares" role). The principle is to avoid reflection of light back into the camera from the scattering medium, as would occur with continuous illumination (e.g., driving a car in fog with headlights on). This is achieved by propagating a short "pulse" of light from near the camera to illuminate the object and then opening the camera shutter at the correct instant to receive only the light reflected from the object. Thus, although the scattering medium absorbs some of the incident and reflected light, the camera shutter is closed at the instant when the incident

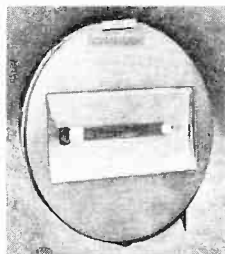


"Business end" of U.K.A.E.A. radar photography equipment—the entrance of the tunnel enclosure being visible on the left.

light pulse is temporarily causing the unwanted reflection from this medium.

Application of the principle depends on the use of electronic methods to achieve sufficiently short light pulses and rapid shuttering. In the demonstration a 20 ns high-intensity light pulse was generated by a Q-switched ruby laser and propagated down a long tunnel enclosure at the far end of which was placed the object to be photographed. The light pulse was therefore about 20 ft long, and at any instant during its travel only the section of tunnel within this 20 ft span was illuminated. Net curtains within the tunnel provided the scattering medium. The "camera" was an electronic image converter (Mullard 6929) and was pulsed from a line controlled by a spark gap to give an exposure time of 20 ns. The "camera" was correctly synchronized with the light source by triggering the spark gap from the firing of the Kerr cell in the laser unit. Suggested applications include photographing industrial processes through clouds of fumes, vapour or particles, as in furnaces.

Electronographic cathode-ray tube. To obviate the inefficient process involved in recording a photographic image from the face of a cathode-ray tube, the General Electric Company suggested a new technique. The conventional faceplate and phosphor are replaced by a thin mica window (3 μ m thick) which is bombarded by the electron beam. The beam, with an energy of about 25 keV penetrates the window thus reaching a photographic film placed directly on the window.



A small electronographic c.r.t. which uses a mica faceplate to enable direct photographic recording of images and eliminating the losses associated with a phosphor screen (G.E.C.)

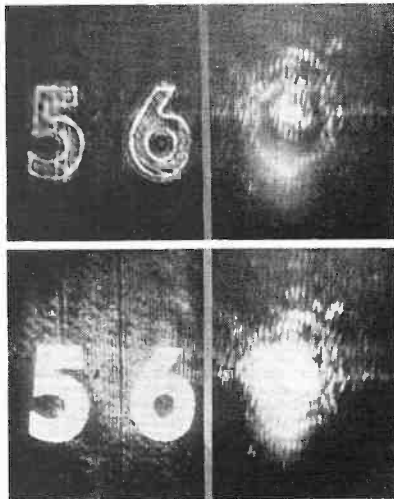
ing good resolution and line widths of 0.0003 in are readily obtainable.

Difficulties exist in obtaining the mica screen and the exhibit showed only a small area of screen.

Contrast enhancement for automatic pattern recognition.

It is known that, due to diffraction effects, the image of a point source produced by a converging lens is surrounded by the Fourier transform of any aperture placed in the plane of the lens. If this is a transparency then the image plane contains the transform of the pattern on the transparency. The spatial components of the pattern are displaced radially about the optical axis and opaque stops can be introduced into the image or Fourier plane to suppress selected frequency components. If the low-frequency components of the Fourier transform are suppressed, areas of constant density in the pattern are eliminated and if the high-frequency components are suppressed the fine granular noise is eliminated.

Thus, this method allows contrast to be increased and



Showing enhancement of boundaries by introducing stops to suppress selected frequency components a diffraction pattern. The N.P.L. suggests that the technique may have application in automatic pattern recognition.

noise to be decreased, and may find application in automatic recognition and processing of optical patterns. An attribute of the system is that several patterns may be processed simultaneously.

Laser Doppler velocity meter has been installed in a factory for the purpose of measuring the speed of aluminium sections emerging from an extracting machine, and the developers, Decca Radar, gave a demonstration of its principles. Light from a low-power gas laser was focused on the moving surface, a rotating disc with its drive shaft inclined at an angle to the incident beam. The back-scattered energy, together with a sample of the incident laser light, was focused on to a balanced photomixer to produce a difference-frequency signal. A tracking unit followed the frequency of this signal and displayed it on an indicator as velocity in ft/min, for comparison with a similar indicator showing the disc surface velocity measured directly by a tachogenerator. Features of the system are its high sensitivity, sufficient to give good Doppler signals from a diffuse scattering surface, and reduction of local oscillator noise in the design of the homodyne balanced mixer. Other possible applications of the principle include measuring the length and vibration of surfaces and monitoring the surface finish of materials.

INFORMATION PROCESSING AND STORAGE

Correlation techniques for measurement and detection (see April 1966 issue, p. 190) appeared in a number of applications. In an "automatic "go, no-go" servo tester

introduced by Sperry, for example, the test signal is a random binary signal and the cross-correlation coefficient between the excitation and the resulting response of the servo is measured for a number of different delays to give the cross-correlation function—which is identical to the servo's impulse response if the delay parameter is considered as real time. The tester runs automatically through 11 preselected delays. At the end of each measurement a voltage representing the c-c coefficient is examined and if it is within required acceptance limits the tester passes on to the next measurement. If all 11 tests are passed the tester indicates "go." If any test is failed it can be repeated, and if the servo continues to fail it "no-go" is indicated. Acceptance limits are fed in by a programme plug.

In a cross-correlator for electroencephalograph signals shown by J. C. Shaw (M.R.C.) and K. R. McLachlan (Southampton University) simplicity is achieved by a digital mode of operation. One of the input signals is quantized into a number of levels, then the binary numbers generated from the different levels are delayed in shift registers and used to gate the second input signal. Another digital correlation technique, shown by Plessey, is intended for extracting sequences of pulses or events with fixed intervals between them from a noisy environment. Nearly all the functions required are performed by one simple circuit unit.

Elliott-Automation Radar Systems demonstrated a radar technique using correlation which achieved the effect of transmitting short pulses of high peak energy, giving good range resolution without actually requiring high peak powers. A long 210-ns pulse of r.f. energy is coded by phase reversal into the 7-bit sequence 1110010 before transmission (using a circulator with diode switches in a transmission line connected to one of its ports). On reception the reflected pulse sequence is fed into a correlator, where it is processed into a shorter pulse of 30 ns containing substantially the same energy and having, consequently, about seven times the amplitude. This would give a radar range resolution of 5 metres.

A correlation type of detector was also used at the receiving end of a system of communication using noise type signals demonstrated by Marconi. If a purely angle-modulated carrier is multiplicatively mixed with wideband noise, selection of one of the resulting sidebands gives a signal which to all normal detectors is pure noise. Thus secrecy of transmission can be obtained. At the receiver the signal is applied to one input of a correlation detector while the original noise, also transmitted separately, is applied to the other input. The original modulated carrier is then obtained at the detector output. As noise is not cyclic several channels can be transmitted by this method, the output of the noise source being delayed a different amount for each channel and separate correlation detectors being used for them. In the demonstration two speech channels were transmitted, using frequency modulated inputs and a 15 Mc/s noise source. The system is said to be reasonably immune from jamming and interference.

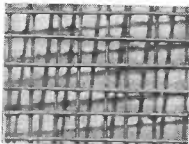
Non-destructive delay-line store, not requiring continuous circulation to hold the information, was demonstrated by Plessey. Intended for use as a register, the experimental device stores 20 bits, but capacities of several hundred bits are envisaged. It comprises a length of 0.0075 in diameter beryllium-copper wire plated with a thin film of magnetic material. Current signals corresponding to the binary pattern to be stored ("on" for 1, "off" for 0) are passed through the conductor, setting up

magnetic fields, while at the same time mechanical stress pulses are transmitted by a PZT transducer—the acoustic delay being 60 μ s.

The effect of the stress pulses is to "freeze" the time-varying magnetic signals into a spatial magnetic pattern on the line. Each stress pulse reduces the coercivity of the magnetic film over the small section of line occupied by the pulse during its travel. The magnetic fields set up by the current signals magnetize the film, but the value of remanence left at a particular section of line after cessation of a "1" current signal depends on the coercivity of that section, which is determined by the mechanical stress pulse. If, therefore, a "1" current passes through the conductor when the stress pulse is occupying a particular section of line, that section will be left with a higher value of remanence than the rest of the line and so will magnetically store a "1" digit. Similarly with successive stress pulses acting at other positions along the line.

Read-out is performed by transmitting another stress pulse, and the resulting changes of flux as it moves down the line induce 60 μ V pulses in the conductor. Signal/noise ratios of up to 17 dB in the read-out signals can be obtained with a digit rate of over 400 kc/s.

Memory stores by electrophoresis. The technique of electrophoresis has been well used in the past and recently the possibility of using this method for the formation of ferrite memory stores has been studied by the British Scientific Research Association and the Royal Radar Establishment. It should be possible to achieve very thin ferrite sections, allowing faster switching times and lower



Illustrating the deposition of ferrite computer stores by electrophoretic deposition.

drive currents, by this method. Another advantage is that the laborious threading of toroids would become unnecessary.

One variety of ferrite has been used so far which has to be sintered at about 1400°C to give the requisite square-loop characteristic, and has so far restricted the conductors to platinum.

Read times of 50 ns have been achieved with 0.005 in wires with an interrogate current of 300 mA.

Because deposition can easily be controlled and the ferrite can be deposited on electrodes of almost any shape, desirable element shapes, which have previously been thought to be uneconomical, may well prove to be easily fabricated.

MEASUREMENT AND ANALYSIS

Direct measurement of transfer functions of electrical networks and mechanical devices was demonstrated with new solid-state equipment developed by Wayne Kerr from their earlier Transfer Function Computer. The new equipment is more flexible in that it allows two types of measurement—in one the system excitation is provided by a built-in signal generator and in the other it is applied separately or is already present in the system. The equipment also permits conventional frequency response

(continued on page 233)



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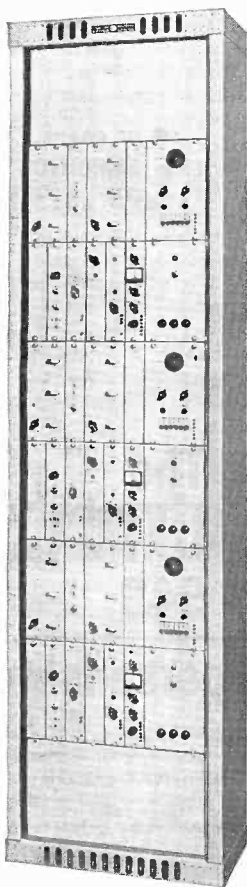
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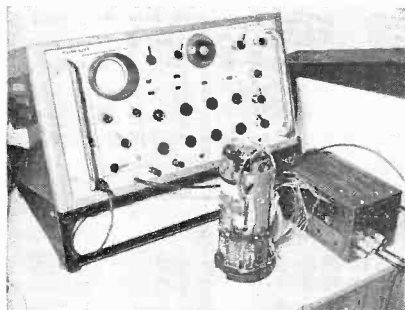
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RECEIVER



BREAKTHROUGH

Marconi telecommunications systems



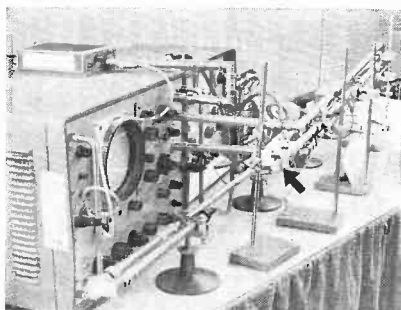
Wayne Kerr instrument for direct measurement of transfer functions, with electro-pneumatic servo unit in front.

measurements (magnitude and phase) to be made. The basic principle is to set up a model of the system using d.c. analogue computing elements (mostly integrators and potentiometers), apply a common excitation to the system and the model, connect the response signals of the system and model to a c.r.t. null indicator, then adjust the model until the null indicator shows that the model is equivalent to the system. Excitation can be a sine wave, a ramp or some other repetitive waveform, typical frequencies being from 0.01 to 500 c/s. Transducers are used where the excitation or response is mechanical.

The transfer function is presented as the ratio of the response to the excitation, derived from the linear differential equation describing the behaviour of the system and expressed in operational form. Adjustment of the model is a matter of adjusting calibrated potentiometers which vary the equation coefficients in the numerator and the denominator, and once a null balance is obtained the coefficients can be read from the dials. Demonstrations included the use of a self-contained instrument for educational purposes, measurements on an electro-pneumatic positioning servo and tests to determine the dynamic properties of elastomers. Also shown was a commercial instrument using the technique for operational performance testing of servos.

Transistor measurements by reflectometry, operating in the time domain, were demonstrated by the College of Aeronautics. The method is an extension of the technique for locating and analysing discontinuities in transmission lines by propagating impulses or step functions along them and examining the waveforms of the resulting reflections on a c.r.o. If the discontinuity takes the form of an inserted network the network elements can be determined from an analysis of the reflection waveform—and, of course, such a network can be a transistor. The function of the transmission line is then merely to separate the incident and reflected waves in the time domain so that they can be conveniently displayed.

In the demonstration the transistor was incorporated in a 50- Ω coaxial line system fed from a pulse generator, and voltage waveforms at points equidistant from the transistor network ports were recorded by a two-channel sampling c.r.o. and an x-y plotter. Five response curves were plotted—the transmitted and reflected pulses for



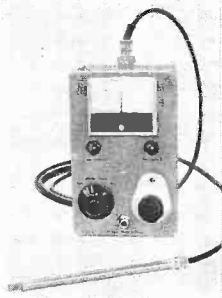
Apparatus for reflectometry transistor measurements (College of Aeronautics, with black arrow showing coaxial line link containing transistor under test.

both directions of transmission and a reference pulse obtained by substituting a direct link for the transistor under test. From these curves, which contain information characterising the transistor network over the entire frequency spectrum provided by the excitation pulse (up to 1 Gc/s), the required data are obtained by computer analysis—though this was not actually done at the exhibition.

The corresponding frequency-domain parameters can be derived, and conversion to the h , y or z parameters can be performed by standard procedures. The technique promises to be simpler and quicker than the conventional bridge method of parameter evaluation—assuming that a digital computer is available—but its full potentialities and limitations have yet to be explored.

A gallium arsenide thermometer has been developed by the International Research & Development Company. It is suited for the range 4-300°K and should find application in research cryostats and refrigeration systems, although it is not suitable for accurate calorimetric investigations at the lower end of the scale, owing to the comparatively high power required.

The instrument has an almost linear scale and relies on the temperature sensitivity of the forward voltage drop across a gallium arsenide junction (at constant current). Variation of the current through the junction alters its sensitivity and it is necessary to increase current in order to increase sensitivity.



Gallium arsenide thermometer using the temperature dependence of the forward voltage drop across a GaAs junction and developed by I.R.D.C.

NEW TECHNIQUES IN MEDICAL ELECTRONICS

AS SEEN AT THE PHYSICS, AND BIOMEDICAL ENGINEERING EXHIBITIONS

As opposed to the numerous systems available for investigating man closely under highly organized circumstances (space flight, operating theatre, extreme environmental conditions, etc.), there were, until quite recently, few methods by which the physiological parameters of a normal man in normal conditions could be measured and recorded. This was simply because instruments for such research must necessarily be of very small weight and size, inconspicuous when worn, be capable of storing acquired data for long periods and be inexpensive; in fact they have to be "socially" acceptable. The biomedical engineering division of the National Institute for Medical Research has scrutinized many methods of recording such data, which were rejected for reasons such as power consumption, cost, or even size (in spite of miniaturization).

However, electro-chemical methods of data storage have been used with encouraging results. One instrument (Fig. 1) employs a commercially available electro-chemical integrator of very small size, and high charge sensitivity. Here, the use of three such devices permits division of data in time, so producing a table of results not unlike the analysis of a continuous recording. Other instruments shown were a heart beat totalizer, and a temperature excursion integrator. The size of such instruments varies, but lying between the size of a matchbox, and a packet of 20 cigarettes, they are still conveniently small, and with power consumption of the order of $400\mu\text{W}$.

A functional stimulator with myo-electric control has been developed for patients who still have some muscle control although they may have extensive muscle paralysis. Feedback of the output signal into the amplifier

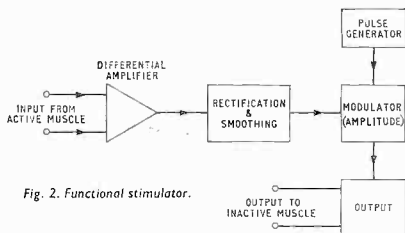


Fig. 2. Functional stimulator.

via the patient's body is inhibited by isolation transformers in the circuit and separate battery supplies. The apparatus, shown by Devices Ltd., is suitable only for conditions where paralysed muscles have not degenerated.

An ophthalmoscope that employs ultrasonics for pulse echo scanning of the interior of the eye was exhibited. It is intended to serve as a diagnostic instrument for the quick location of foreign matter, the detection of a detached retina, the measurement of its extent, and for the location of tumours. This ultrasonic examination can be carried out safely and quickly even with eye conditions such as opacity, or cataractization.

Marketed in this country by T.E.M. Sales Ltd., it has a frequency range of 3-20 Mc/s, and results are displayed on the screen of a 5-in diameter c.r.t. The output from the transmitter is controlled so that only sufficient energy is fed into any of the examination probes. Changing frequency is effected by changing the probe, no further switching being required. A reject control permits the operator to blank out the smaller echoes, thus obtaining a cleaner pattern. Enlarging of the response pattern is continuously controllable within a wide range; at minimum enlargement the range amounts to 13mm of the eye displayed on the screen, and at maximum enlargement the range is 130mm.

A correlator is being developed by the Clinical Psychiatry Research Unit at Graylingwell Hospital, Chichester, Sussex, to assist in the spatial analysis of electro-encephalogram signals. The form of e.e.g. is governed by the site on the scalp from which it is recorded, thus the topography of the e.e.g. is regarded as important. Since the cross-correlation function is a measure of the relationship between signals from two sites, this is measured over each contiguous second of a sample of e.e.g. record, thus determining the inherent variability of the record.

The design of the correlator has been simplified by using a digital method, in which one signal is quantized into six quantization levels. Time displacement of the two signals to be correlated is carried out by shift registers, and multiplication reduced to a gating process. The signals pass through the correlator, once for each time-displacement value. The output from the correlator is stored on punched paper tape, each number on the tape representing a point on the correlogram.

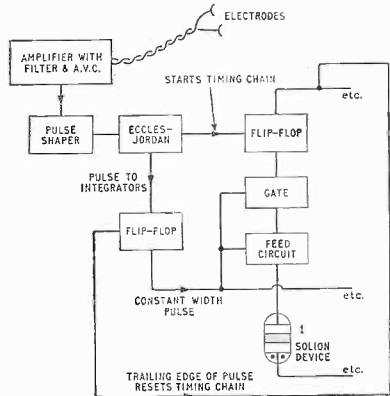


Fig. 1. Heart rate integrator.

Public Address Equipment

SOME APPARATUS SEEN AT THE ANNUAL A.P.A.E. EXHIBITION

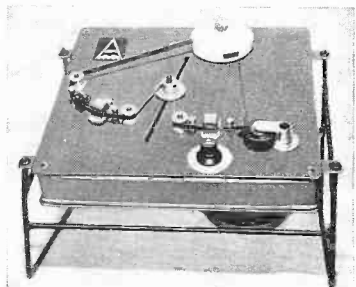
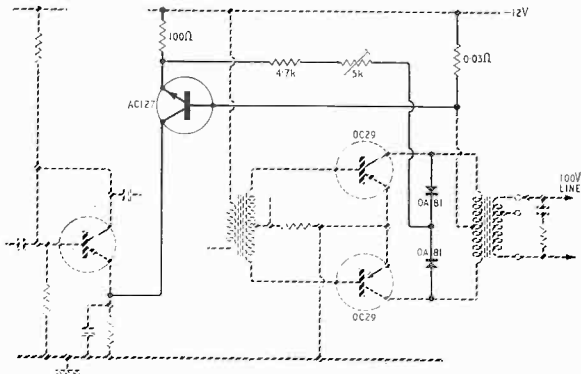
ONE of the most outstanding demonstrations for those who visited the Association of Public Address Engineers' Exhibition at the Kings Head, Harrow, was given by one of the Association's officers, Mr. Haydon Warren. A Western Electric 618 moving-coil microphone, introduced around 1931 and forerunner of the S.T.C. 4017, fed a permanent magnet version of the 555 loudspeaker (a 1928 design) via a 250 mW transistor amplifier concealed in the microphone stand. The resulting sound level was surprising and quite usable for small halls, the secret being the loudspeaker—with an electrical to acoustic efficiency of 50%!—the most efficient loudspeaker yet produced?

Amplifiers

A number of manufacturers have introduced more transistor amplifiers this year. Wharfedale (now, incidentally, known as Rank Wharfedale) announced their new 50 W design. This four-channel amplifier has a sensitivity of 50 μ V and will provide about 50 W (75 W music power) into a 100 V line, or into a low impedance load, with 4% harmonic distortion. Three Philips transistor amplifiers were introduced by Peto Scott (10, 20 and 35 W) and the new Sound Coverage range includes 15, 30, 60 and 120 W designs with a sensitivity of 200 μ V and a distortion of 3%. Output transistors (in push-pull) are used in parallel for the high powers. The new 50 W Gramplan (type 662) amplifier uses four 2N2869 transistors in the output stage, each having a fuse in the collector leads so that in the event of failure of one transistor, operation is still possible. To protect the transistors should the accumulator supply be incorrectly connected, a diode and supply relay are included.

A 50 W amplifier design was introduced by Mullard

Protection circuit for Mullard 50W amplifier design to give protection against short circuited load. Rising current in the output stage causes the a.f. drive to be reduced by tending to cut off a transistor (OC71 on left).



A tape transport mechanism developed by Planet Projects. This illustrates a simple method of obtaining a variable delay between recording and playback.

for operation from a 12 V battery. A single matched pair of output transistors (OC29) are used and distortion is less than 10%. To avoid destruction of the output transistors when the load is short circuited a protective circuit is included (see accompanying circuit). Should the output be short circuited the output transformer primary current rises and the AC127 turns on, tending to cut off the OC71, and thus reducing the drive. The pre-set variable resistor is adjusted at maximum output so that the OC71 collector current just starts to fall.

CTH Electronics augment their transistor amplifiers covering the range 15-1000 W with a 25 W unit which



includes the usual feature of automatic switching from mains to battery in the event of mains failure.

Armstrong have entered the p.a. field by modifying their 227 tuner-amplifier for 100 V line working. A version is available with a balanced 30 Ω input. As a high quality amplifier the distortion will be considerably less (0.5% at 8 W) than the average p.a. amplifier.

Microphones

Although a number of new types of microphone are under development (e.g. semiconductor and electret types) it is not expected that these will appear for some time in the p.a. field. Two of the new mikes shown this year were the Amplivox Telemike and the first microphone of the recently formed London Microphone Co. This last type is a low-cost moving-coil (illustrated) and forms a very neat and inexpensive microphone for various uses including transistor tape recorders—some manufacturers have been importing low cost Japanese moving-coil microphones for this purpose. Sensitivity at 500 Hz is 300 μ V/ μ bar and the frequency response is very smooth down to about 200 c/s and at 50 c/s the response is down to 10 dB. The Telemike is a two-way transducer and was briefly described in the April issue (p. 210).

Some other new microphones are the Reslo CPD pencil type†, the Shure 581 series, a number shown by Fi-Cord (including Beyer types), the A.K.G. D500 series, the Philips EL6033 and the Acos Mic 70. A Neumann condenser microphone (KTM) was exhibited by Bauch (who have, incidentally, changed their address to Holbrook House, Cockfosters, Barnet). This incorporates an f.e.t. followed by a common-emitter mode silicon transistor. A 9 V battery powers a simple built-in d.c. converter to supply the polarizing voltage.

† See *Public Address Engineers Journal*, March 1966.

The first microphone to be produced by the London Microphone Co. This moving-coil microphone is suitable for domestic tape recorder use and costs about £3 (retail).



A photograph which arrived too late for last year's display of early p.a. equipment. The transmitter-size amplifier (a W.E. No. 1 amplifier) was rated at 40 W and dates back to 1919.

Some portable equipment

Of the mixers shown, the EMT104, shown by F. W. O. Bauch, was perhaps the most interesting. This robust portable transistor mixer includes a battery cassette taking either primary or secondary cells and a charging unit. Some of its features are a peak programme meter, monitor loudspeaker, tone generator, rumble filter on each channel and a switchable limiter and compressor.

Two new portable transistor p.a. systems by the French firm of Bouyer were shown by Claude Lyons and Associates. In the smaller (5 W) version which measures about 22 x 5 x 3 in the amplifier is enclosed together with batteries and a number of loudspeakers. A stand mounting version is also available with a longer loudspeaker column.

A small sound level meter was shown by Cosmocord covering the range 60-95 dB s.p.l., with "A" weighting.

Communication Receiver Survey

Amendments

IN our survey last month Eddystone was listed under Stratton and Co. Ltd. but should, of course, have been under Eddystone Radio Ltd. which is a subsidiary of The Marconi Company Ltd.

Under K.W. Electronics Ltd. the HQ180A should read HQ170A. Details for the HQ180A which were omitted are as follows:—

Country of Origin U.S.A.

Type of Circuit Double and triple superhet.

Frequency Bands 540 kc/s-7.85 Mc/s (double conversion); 7.85-30 Mc/s (triple conversion); bandspread 3.5-30 Mc/s (calibrated for amateur bands).

Receiving Modes s.s.b., a.m., c.w.

Impedances 52-600 Ω (I/P) 3 and 500 Ω (O/P).

Sensitivity and (S+N)/N Average 1.5 μ V for 10:1 s/n on a.m. and 0.7 μ V on s.s.b. and c.w.

Selectivity Selectable sidebands, u.s.b. 1, 2 and 3 kc/s; l.s.b. 1, 2 and 3 kc/s; d.s.b. 0.5, 2, 4 and 6 kc/s at 6 dB down.

Image Rejection 55 dB.

I.F. 1st 3.035 kc/s; 2nd 455 kc/s; 3rd 60 kc/s.

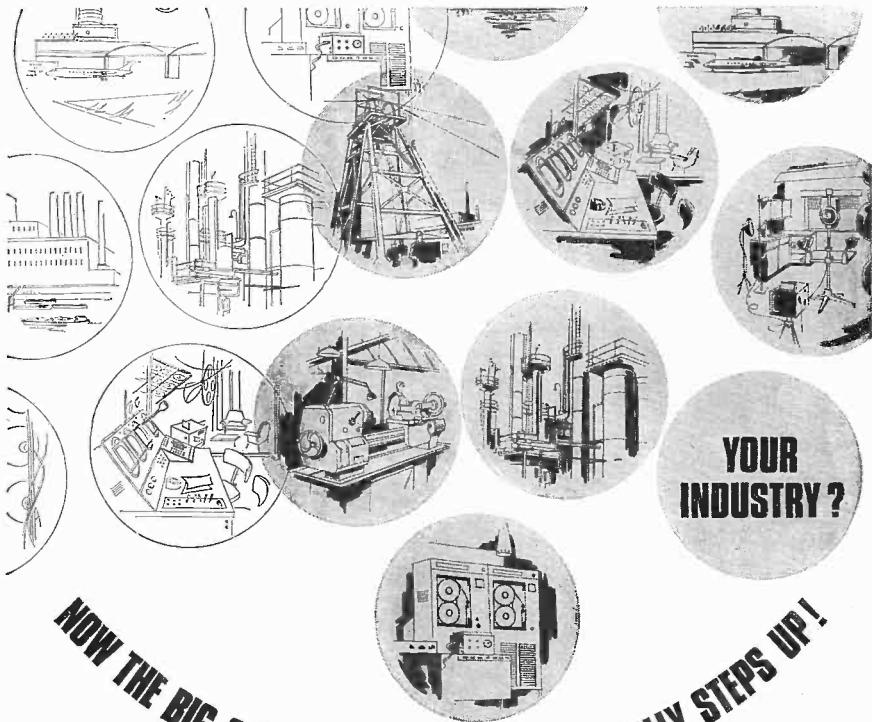
Local Oscillator 1st tunable; 2nd 2,580 kc/s (xtal); 3rd 395 kc/s. The receiver also incorporates a xtal filter (on 0.5 kc/s selectivity); b.f.o.; a.n.l.; xtal calibrator; power unit.

Gain Controls n.l., r.f.

Valves 1B.

Semiconductors 2.

Additional Information Product detector, vernier tuning, aerial trimmer, calibrated "5" meter, 14 controls.



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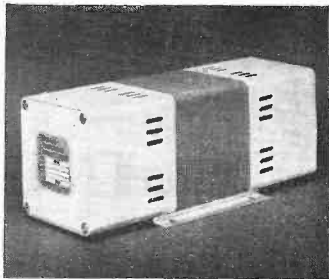
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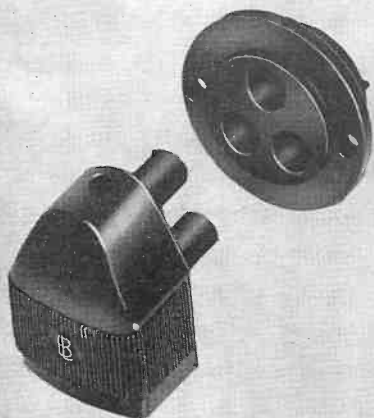
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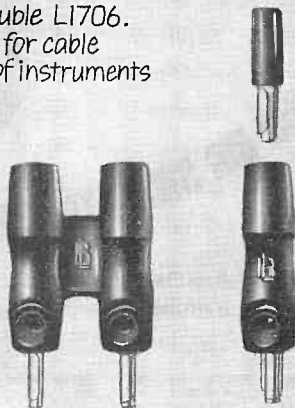
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A.F. Amplification with the Cascode

AN OUTLINE OF THE ADVANTAGES OF THE TWIN-TRIODE OVER THE PENTODE FOR A.F.

By G. A. STEVENS

PRIOR to the development of Band III television tuners the cascode was very little used, its main use being in low noise r.f. stages, and it was used as a voltage amplifier in a stabilised power supply, where high gain was required.

The properties that enhance its use for r.f. amplification also indicate its suitability as an a.f. voltage amplifier. These advantages are as follows:—

1. High gain.
2. Low noise.
3. Low inter-electrode capacitances (particularly between input and output).
4. Capability of being designed with low phase shift (in feedback amplifiers).

The characteristics of the cascode that give rise to the above advantages are best illustrated by a simplified analysis of the circuit (a fuller analysis will be found in the "Radio Designers' Handbook" by Langford-Smith, Ch. 12, Sect. 9 XI).

In the basic cascode circuit as shown by Fig. 1, two halves of a double triode are connected in series, the bias for the lower valve, V1, being derived by grid current through the high value of grid resistor, the upper valves grid being held at a fixed potential V_{g2} .

Since the anode potential of V1 is held constant by the cathode follower action of V2, the change in anode current of V1 caused by an alternating voltage $V_{i,p}$ on its grid is:—

$$\delta I_a = V_{i,p} \cdot g_{m1}$$

where g_{m1} is the g_m of V1.

Providing V2 does not draw grid current, all current entering at its cathode must pass through the anode load R_L .

Hence,

$$\begin{aligned} V_{o,p} &= \delta I_a \cdot R_L \\ &= V_{i,p} \cdot g_{m1} \cdot R_L \\ G &= \frac{V_{o,p}}{V_{i,p}} = g_{m1} R_L \dots \dots \dots (1) \end{aligned}$$

Hence the circuit has the same gain as a pentode with the same g_m as the lower valve.

Also, since for any valve that has an impedance R_k in its cathode, its anode impedance is increased by a factor: $(1 + g_m R_k)$.

We can write for the cascode:—

$$r_a = r_{a2} (1 + g_{m2} r_{a1}) \dots \dots (2)$$

since the anode impedance of V1 is in series with the cathode of V2. Therefore, from equations (1) and (2) we can derive:—

$$\mu = r_a g_m = r_{a2} \cdot g_{m1} (1 + g_{m2} r_{a1}) \dots (3)$$

From equation (2) it can be seen that for a twin triode

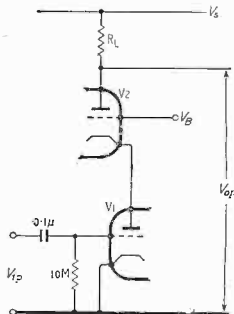


Fig. 1. Basic cascode circuit.

with $r_a = 100 \text{ k}\Omega$ and $\mu = 70$ then its composite anode impedance is:—

$$r_a = 100 (1 + 70) \text{ k}\Omega \approx 7 \text{ M}\Omega$$

i.e., much more than a similar pentode.

The similarity of characteristics to a pentode is shown even more clearly by considering the knee voltage area. As stated above all the current flowing into the cathode of V2 must pass through the load R_L providing no grid current is drawn. However, once this point is reached, increasing the input drive only produces more grid current, so producing a very sharp pentode-like knee point as V2 bottoms. So far, it can be seen that the characteristics are those of a semi-idealized pentode, certainly better than the normal amplifying pentode. There are, however, two factors which make the cascode superior to the pentode for audio amplifier use.

Firstly, since there is only one grid in a triode, this taking no current, the partition noise generated by a pentode is absent, so reducing the noise level by a factor of three or more.

Secondly, in a pentode, the g_m is dependent on the anode current, which in turn depends on the load resistor



G. A. Stevens has for the past five years been a maintenance engineer with Rediffusion Television Ltd. From 1932 until being called up for National Service in the R.A.F. in 1939 he was in the Research Laboratory of E.M.I. which he joined on leaving school at the age of 19.

SYMBOLS

$g_{m1} g_{m2} g_{m3}$	Mutual conductance of V1, V2 and combination respectively.
$F_{a1} F_{a2} F_a$	Anode impedance of V1, V2 and combination respectively
$\mu_1 \mu_2 \mu$	Amplification factor of V1, V2 and combination respectively
$G_1 G_2 G$	Voltage gain of V1, V2 and combination respectively
G	Overall gain after application of negative feedback.
$V_{i,p}$	Signal input voltage
$V_{o,p}$	Signal output voltage
R_a	Anode load
V_{c3}	The d.c. voltage at the cathode of V3
Z_a	The output impedance at the anode of V2
C_M	Effective Miller capacitance at the grid of V1
β	The fraction of the signal voltage at the cathode of V3, which is fed back to the grid of V2

and the supply voltage available. So that, for a given supply voltage, increasing the anode load decreases the g_m and a point is reached where increasing the load actually lowers the gain. In the cascode, however, the gain G is the product of the anode load of V2, and the g_m of V1.

Suppose now in Fig. 1 that a resistor was inserted between the anode of V1 (and cathode of V2) and the h.t. rail V_S . This would mean that V1 could be taking a reasonable current and so have a good g_{m1} whilst V2 could have a large value of anode load with its consequent low anode current.

This arrangement can, and does, give very high gain providing certain points are kept in mind.

1. Since V2 can now be cut-off without affecting the working conditions of V1, the bias of V2 (V_{a2}) must be derived *via* potential divider of some sort from the anode of V2.
2. The anode load for V1 is the cathode of V2 and so is equal to $1/g_{m2}$ and when the mutual conductances of both valves are equal, the gain of V1 is equal to -1 and the Miller capacity at the grid of V1 is only about 3 pF (for an ECC83). However, with the top valve drawing only a small current its g_{m1} is very low and presents a considerable load to V1. This means that the gain is shared between the two valves and the Miller capacity can be high, although the screening of input to output is still good, due to the cathode grid of V2.

Usually, a circuit with this gain would only be needed at the input of an amplifier, and should therefore be fed from a low-impedance source which swamps the effect of the Miller effect.

Since the grid of V2, and hence its cathode, is stabilised by the action of a potential divider from its anode, the effect of changes of bias on V1, due to signal level variation is, to a large part, compensated for, so allowing the use of grid current bias.

If, as is usual, the negative feedback for the whole audio amplifier in which the cascode is incorporated is applied to a small resistor in the cathode of V1 (about 100 Ω , which has little effect on the parameters of the circuit); then, the 0.1 μ F capacitor and 10 M Ω resistor in the grid circuit are outside the feedback loop, and so do not affect the phase shift over the cascode. Similarly, since V2 is drawing very low anode current the voltage

feed, V_{a2} , to its grid can be supplied by another 10 M Ω resistor and 0.1 μ F capacitor, so giving virtually no i.f. phase shift down to very low frequencies.

The bias voltage, V_{b2} , can be supplied in one of two ways:—

1. By direct potential divider from the anode of V2. This method has two disadvantages, the maximum value of resistance usually made in standard ranges is 10 M Ω , and with this value as an anode load, the potential divider has a resistance of this order which shunts the signal, and so reduces the gain.
2. The anode load and potential divider can, if the value of R_a is too high cut off V2 by reducing its anode voltage to the point where no current flows through the valve.

The above drawbacks limit the straight potential divider at high values of R_a , although when R_a is around 1 M Ω or less there is nothing to choose between this and the next bias method to be described.

If the potential divider, instead of being applied direct, is in the cathode load of a cathode follower; the grid of which is connected to the anode of V2 the above drawbacks disappear.

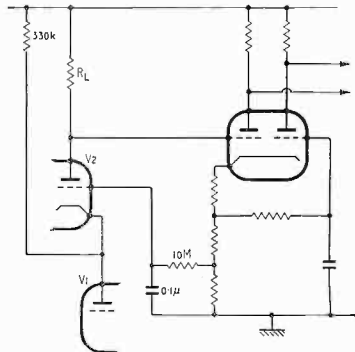
Usually in modern audio amplifiers the trend is to follow the input voltage amplifier with a directly coupled phase splitter, of the cathode-coupled variety, in this case the cathode resistor of the phase splitter may be replaced by two other resistors so forming the necessary bias network for V2, as illustrated by Fig. 2.

Practical cases

In order to investigate a practical case the two circuits shown in Figs. 3 and 4 (with and without the cathode follower bias network respectively) were constructed. The value of the 330 k Ω current bleed resistor was found experimentally for an anode load of 10 M Ω and on altering the load the value still seemed about optimum and so no experiments were conducted using values other than this.

S1 and R_2 were used to determine the output impedance of the circuit, R_2 being altered until closing S1 reduced the

Fig. 2. Bias arrangement of the cascode circuit to feed a cathode-coupled phase inverter.



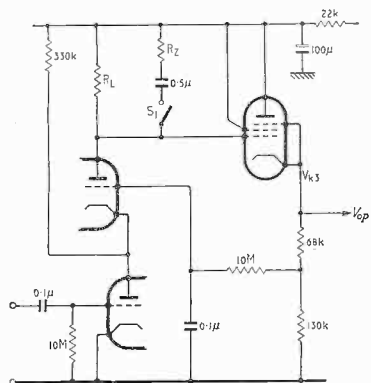


Fig. 3. Circuit used for measurements incorporating the cathode biasing method. Results are shown in Table I and Figs 6 and 8.

output signal by 6 dB, R_a then being equal to the output impedance (the series capacitor having negligible reactance).

In the experiments only the anode resistors were altered, together with slight adjustments to the stabilized power supply to give 300 V after the decoupling resistor.

The gain from the input to the anode of V1, was also measured for different loads, this giving the measure of Miller capacity on the input, and also the gain of the cascade if:—

1. The grid of V2 is not decoupled to earth; or
2. The effect of putting a frequency sensitive network between the cathode of V3 and the grid of V2 to act as an equalizer circuit.

The results are set out in Table I and in the graphs of individual parameters which follow later. The odd

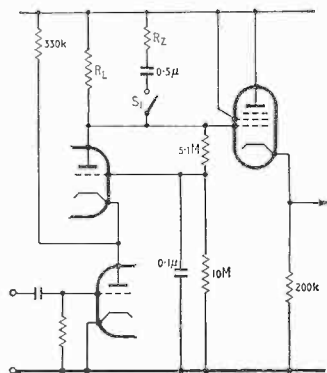


Fig. 4. Test circuit used for measurements. S_1 and R_a were used to determine the output impedance. Results are shown in Table I.

values of R_L being the measured values of standard resistors.

As can be seen, when the anode load is in the order of 10 MΩ the output impedance is 3.5 MΩ. In the test circuit 3 the stray capacity at the anode of V2 was 8 pF and so the response was 3 dB down at about 6 kc/s, the gain-bandwidth factor is given by the ratio G/Z_o , and as can be seen, increases with lower anode loads. The early roll-off associated with the highest gains can in fact be a desirable feature in feedback amplifiers, where it is necessary to reduce the loop gain below unity before the phase shift reaches 180° in order to avoid instability. When the cathode-coupled phase splitter uses a pentode in the first stage, so avoiding large values of Miller capacitance loading the anode of V2, the total stray capacitance should not exceed 16 pF and so the 3 dB upper

TABLE I

R _L	FIG. 3 Cathode follower bias							FIG. 4 Potentiometer bias					
	V _{k2}	C _M	G ₁	G	G(dB)	Z _o	G ₂	G/Z _o	G	G(dB)	Z _o	G ₂	G/Z _o
16 MΩ	153V	103pF	68	4,800	73.5	3.8 MΩ	70	1.26			Cut-off		
13 MΩ	153	103	68	4,800	73.5	3.67 MΩ	70	1.26	940	59.5	2.58 MΩ	13.8	0.365
10 MΩ	153	90	60	4,400	72.8	3.52 MΩ	73	1.25	2,350	67	2.2 MΩ	39	1.02
6 MΩ	154	75	50	3,750	71.3	2.48 MΩ	75	1.52	2,500	68	1.73 MΩ	50	1.44
3 MΩ	157	53	35.6	2,850	69.0	1.72 MΩ	80	1.64	2,100	66	1.35 MΩ	59	1.55
1.5 MΩ	161	34	22.5	1,850	65.2	1.13 MΩ	82	1.62	1,480	63.5	940 kΩ	66	1.58
1 MΩ	165	24.5	16.3	1,325	62.5	790 kΩ	81	1.67	1,130	61	693 kΩ	69	1.63
620 kΩ	173	16	10.8	900	59.0	540 kΩ	83	1.67	815	58	493 kΩ	77	1.66
400 kΩ	184	11.5	7.6	632	56.0	370 kΩ	83	1.71	610	55.5	334 kΩ	80	1.83
253 kΩ	195	8.5	5.6	423	52.5	235 kΩ	75	1.80	418	52	203 kΩ	75	2.03
137 kΩ	217	5.5	3.6	251	48.0	130 kΩ	70	1.93	250	48	125 kΩ	70	2.0
91 kΩ	232	4.4	2.9	176	45.0	87.5 kΩ	61	2.01	175	45	86 kΩ	61	2.03
52 kΩ	253	3.1	2.1	98	40.0	51.1 kΩ	47	1.92	98	40	50 kΩ	47	1.96
30 kΩ	260	2.5	1.7	56.5	35.0	28.8 kΩ	33	1.96	56.5	35	28 kΩ	33	2.02

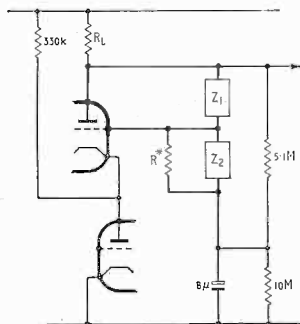


Fig. 5. Use of the cascode for equalization by frequency selective feedback. R_b is only required to provide bias if Z_2 is blocked to d.c.

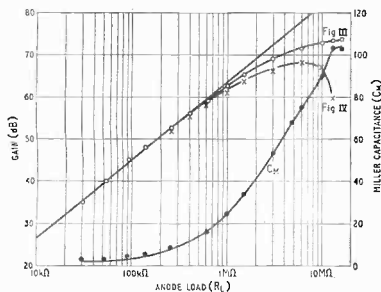


Fig. 6. Some of the results of Table 1 obtained from the circuits of Figs 3 and 4.

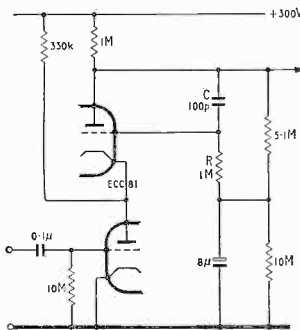


Fig. 7. Particular case of Fig. 5 for use as a tape playback equalizer.

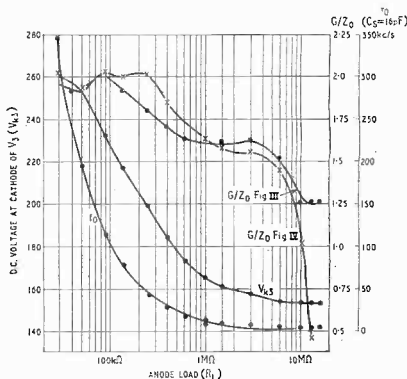


Fig. 8. Some of the results, listed in Table 1, obtained from the circuits shown in Figs 3 and 4.

frequency limit should be at about 15 kc/s with an anode load of $1\text{ M}\Omega$ giving a gain of over 60 dB. The gain-bandwidth product may be obtained from the factor G/Z_0 by multiplying it by 10^{-1} (assuming 16 pF strays).

An EF86 pentode with an anode load of $220\text{ k}\Omega$ and a following grid resistor of $1\text{ M}\Omega$ has a gain of about 200 (46 dB), the effective anode load due to the above two resistors is $180\text{ k}\Omega$, and the cascode gives a gain of 320 (50 dB) with a load of this value, while increasing the pentode load to $1\text{ M}\Omega$ and running under starvation conditions gives a gain of 400 (52 dB), the cascode giving 1,200 (62 dB), an even more marked improvement over the standard pentode circuit.

The ECC83 has a normal heater, as opposed to the bifilar type of the EF86, and so the hum introduced by the circuit could be expected to be greater than that of the pentode. With the $10\text{ M}\Omega$ load in circuit, (as with this value the hum and noise would be expected to be at their worst), and one side of the heaters earthed the total hum and noise at the output was 5 mV r.m.s., and when the heaters were on a d.c. supply the value fell to 1 mV r.m.s.—these correspond to an input level of 3.15 and $0.67\text{ }\mu\text{V}$ r.m.s. respectively over a frequency 25 c/s —6 kc/s. The measurements were taken on Fig. 3 with the input shorted, and includes hum and noise introduced by the output cathode follower.

As can be seen in Fig. 4 the gain ν load characteristic is linear up to $2\text{ M}\Omega$ (for Fig. 3) and then falls off and becomes asymptotic to the 75 dB co-ordinate. This corresponds to the value of μ at low currents as given in the published curves of the ECC83.

These results show the superiority of the cascode circuit over the conventional pentode in audio applications, and whilst the investigations were concerned only with the ECC83, the newly-developed ECC807 should show an even more marked improvement in cases where better hum or gain figures might be needed, although the excellent results obtained with the older valve type would usually make higher gain unnecessary in all but a few cases.

Finally, the effects of valve changes on the gain and d.c. level at the cathode of V_3 (V_{c3}) were investigated by

putting seven different new valves in the V1 and V2 position. The total gain variation was 10%, and the output voltage variation was over an 18V range.

APPENDIX

Equalization with the cascade

If the grid of V2 is not decoupled to earth, but is included in a frequency sensitive network as shown in Fig. 5, then the overall gain is a function of frequency.

The circuit now behaves with V1 as a pre-amplifier with gain G_1 , and V2 as a frequency-selective feedback amplifier. The gain of V2 as a feedback amplifier is given by:—

$$G'_2 = \frac{G_2}{1 + G_2 \beta} \text{ where } \beta = \frac{Z_2}{Z_1 + Z_2}$$

N.B. The application of feedback increases the input impedance of V2 and so also the gain of V1, this causes a reduction in the total lift supplied by the overall circuit, but for values of R_f greater than about 1 MΩ this effect is not serious, but in any case a little experimentation will soon determine the correct values.

Then the total gain of the circuit will be:—

$$\begin{aligned} G' &= G_1 \cdot G'_2 \\ &= \frac{G_1 \cdot G_2}{1 + G_2 \beta} \\ &= \frac{G}{1 + G_2 \beta} = \frac{G(Z_1 + Z_2)}{Z_1 + Z_2(G_2 + 1)} \end{aligned}$$

If Z_1 is not blocked to d.c. a blocking capacitor would have to be inserted to maintain the correct bias level on V2.

An example of a tape equalizer circuit is given in Fig. 7, the values of C and R given provide a time constant of 100 μs with a total top cut of about 36 dB, the i.f. gain being 60 dB, dropping to 24 dB at h.f.

Ceramic Cathode-ray Tube

Cathode-ray tubes can be a relatively unreliable part of the electronic equipment of many aircraft, space and defence systems, and in order to achieve great reliability a ceramic c.r.t. (for magnetic deflection) has been produced. The device has been developed by Elliott-Automation to meet a need for tubes which will withstand high acceleration and severe vibration in military systems.

In conventional tubes the gun assembly is supported within the tube neck by wires and in consequence the assembly can move relative to the screen. This is avoided in the ceramic tube since gun and electrode parts are integral with the tube (see illustration) and consequently greater display

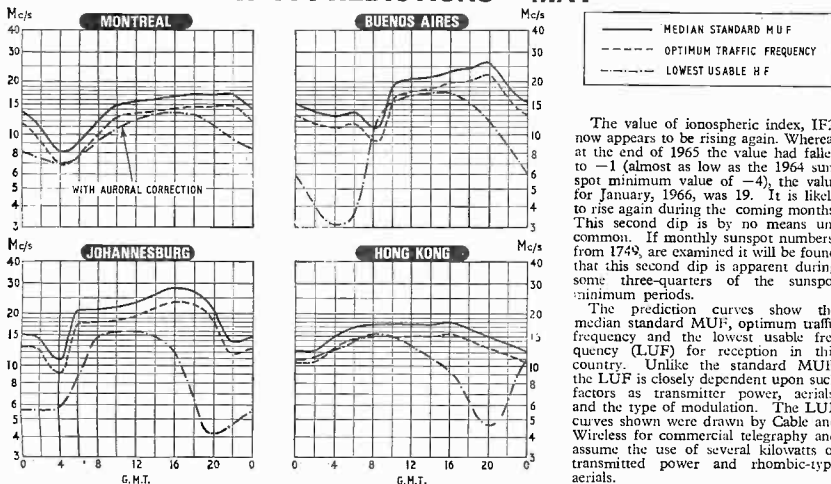


Magnetically focused ceramic c.r.t. developed by Elliott-Automation.

precision is possible. Left of the exposed view is the heater-cathode-modulator assembly. The metal modulator is brazed on to the ceramic material (alumina). This is followed by separate focusing and accelerator sections. The screen is glass, but, if necessary, this can be made from sapphire—a translucent form of alumina. The tube shown has a 1 in dia. screen, but 2 in tubes are in production and it should be possible to manufacture larger diameters up to 5 in or so.

The use of ceramic material enables the tube to be baked at a higher temperature during manufacture, achieving a better vacuum and consequently longer operational life.

H. F. PREDICTIONS—MAY



NEWS FROM INDUSTRY

IMAGE INTENSIFIERS

A JOINT £300,000 development programme by two English Electric companies, Marconi Instruments and the English Electric Valve Co., has resulted in new X-ray image intensifier equipment for medical purposes. The equipment allows medical staff to make detailed examinations of patients without exposing them to the comparatively high levels of radiation required for conventional screening; the dosage necessitated by previous methods has been reduced to about one-sixth and the resultant faint image undergoes an intensity amplification of about 40,000 times.

Basically the system entails the use of an X-ray tube whose rays, directed through the patient, form an image on a flat fluorescent screen, similar to that used for direct viewing fluoroscopy. The screen image is then optically focused on to the photo-cathode of a 4.5in image orthikon Type P825 camera tube, developed by the English Electric Valve Co. The electrical charge image produced by the tube from the optical image is then scanned and the tube output signal, in the form of a television waveform, is amplified and processed before being fed to high definition television monitors operating at 1,024 lines with triple interlaced scan. In addition to direct viewing the monitor display can be filmed or tape recorded.

The major advantage of the intensifiers is that the patient is subjected to less X-ray exposure and the medical staff can view a stronger picture and if necessary monitor a bodily process. An



A "patient" demonstrates the new Mariconette 10in X-ray image intensifier.

automatic beam control incorporated in the camera head and connected to the head amplifier stage of the image orthikon tube, adjusts the brightness of the picture so as to ensure that the best image is obtained with a wide variation of X-ray levels. The automatic beam control facility frees the operator from manually compensating for variation in fluorescent screen brightness during the scanning of a patient.

New S.T.C. Subsidiary.—The activities in the semiconductor field of the S.T.C. plants at Footscray and Harlow are now controlled by a newly formed subsidiary company, Standard Telephones and Cables (Transistors) Ltd. The range of products will include transistors, thyristors, signal diodes, silicon rectifiers, Zener diodes and integrated circuits.

Integrated Circuit Prices Down.—Due to process improvements Electrofil Ltd. have been able to effect price reductions in the Signetics range of integrated circuits. All 14-lead flat package items in the SE1000 military D.T.L. (Diodes Transistor Logic) are reduced by between 30% and 40% and those in the NE100 industrial D.T.L. range by approximately 40%. Of the other military ranges, all packages in the SE400J low-power T.T.L. (Transistor Transistor Log 16) range have been reduced by between 25% and 35% and all except the SE501 in the SE500 linear range have been reduced by between 30% and 40%.

Domestic Entertainment Equipment.

—Figures released by the British Radio Equipment Manufacturers' Association and based on net sale or return figures of deliveries by manufacturers to the home market show that sales of radio receivers (including car radios), radio-grams and television receivers fell during 1965. Figures for the last three years are listed in the table below. Radio receivers also include car radio receivers.

Year	Radio Receivers	Radio-grams	Television
1963	2,616M	0,274M	1,678M
1964	2,139M	0,282M	1,903M
1965	1,73M	0,233M	1,681M

Impetron Ltd., have formed an associated company in the United States with Ad. Auriema, whom they represent in the United Kingdom. The company, called Impetron Inc., will operate from 85 Broad Street, New York, N.Y.4. The function of the new company is to market European components in the United States.

The Electronic Industries Association of America have announced peak sales of colour, and monochrome television sets for 1965. In 1964 monochrome sales totalled 7.685 million, while in 1965 they rose to 8.028 million. Colour sets sold in 1964 totalled 1.366 million, this figure rose to 2.764 million in 1965.

The Australian Department of Civil Aviation has ordered a Solartron 24-target air traffic control simulator. It has a "playing area" of 400 miles square, with an altitude of 50,000 feet, and air speeds of up to 3,000 knots for each aircraft. It will also be used for ATC procedure evaluation, at airfields such as Tullamarine (Melbourne), Mascot (Sydney), Brisbane, Adelaide and Perth. The order is worth about £40,000.

Annual Report of B.E.A.M.A. records that exports by the British electrical industry in 1965 increased by 4½% over the previous year to £379M.

West Germany was chosen by the Electronic Valve and Semi-Conductor Manufacturers' Association (VASCA) for its first Inward Mission, organized in conjunction with the Export Council for Europe. The eleven delegates, who were with three exceptions from Government organizations, visited Ferranti (Chadderton), Joseph Lucas (Sutton Coldfield), M-O Valve (Hammersmith), English Electric Valve (Chelmsford) and S.T.C. (Paignton) during their five-day tour.

Pye of Cambridge Ltd. announce that large losses have been incurred in the radio and television side of the Group, particularly in respect of amounts due from a group of companies to which television sets have been supplied on a large scale. Other main sections of the business, however, are operating profitably.

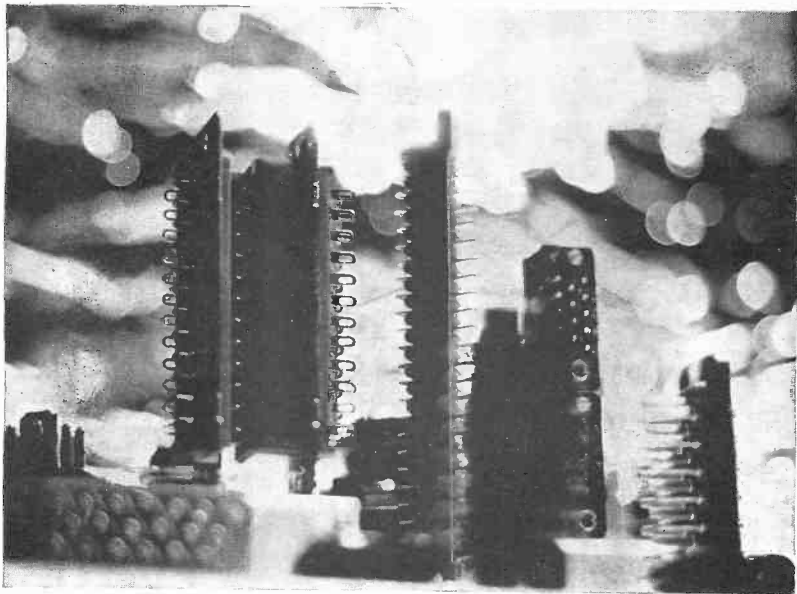
The **Ministry of Aviation** has placed a contract, worth £230,000, with the Marconi Company for a tropospheric scatter radio system. This will be used in Malaysia to provide the Royal Air Force with 24 voice channels between Singapore and Penang. A 10kW transmitter operating at 900Mc/s will be used at either end of the 370-mile link. The tropospheric scatter terminals will be supplied by Radio Engineering Laboratories, a division of the Dynamics Corporation of America.

An agreement for an exchange of patent licenses and technology in the integrated circuits field has been signed between Westinghouse Electric International Company, N. V. Philips of the Netherlands and North American Philips Company.

Under the terms of the contract, Philips and Westinghouse will effect a complete interchange of device specifications as well as process specifications and manufacturing technologies for their lines of digital and linear integrated circuits.

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TAS323

By
H. O. CODON

Communication Theory and Colour Television

Some degree of stability has been achieved in the U.K. colour television situation by the T.A.C.'s recommendation of the PAL system for Britain (and the P.M.G.'s acceptance of it), but this does not mean that the subject is now closed. In engineering the ultimate solution is never reached. This article examines the principal colour systems in the fundamental terms of communication theory, particularly from the point of view of economical utilization of the frequency spectrum, and ends with the intriguing question: what would we do if we could wipe the slate clean and start again?

TELEVISION has always demanded a wider channel than any other service, with the possible exception of high-definition radar. In broadcasting it also demands wide coverage, which implies (a) the use of the lowest possible frequency and (b) the use of frequency channels exclusively over a wide geographical area and with such high power that interference could be caused to other services using the same frequency in adjacent geographical areas. In addition, the cost of producing television programme material is so high that it is desirable to transmit programmes by landline or other networks between broadcast transmitters. The problem has been exacerbated by the demand for colour, which requires the transmission of additional information.

There have been many attempts to reduce the bandwidth occupied by the basic signal, but no scheme has been put into practical use: on the contrary, the transmission of television by pulse code modulation has been developed^{1,2,3}. This requires several times the bandwidth of the original signal, though in one case² the digits were sent through physically separate channels so that the bandwidth required for each channel was that of the original signal. The advantage gained is the regenerability of p.c.m., which allows a large number of repeaters to be used in cascade without imposing too stringent conditions on the performance of each. On the other hand, considerable ingenuity has been shown in transmitting coloured pictures within the same bandwidth as monochrome pictures.

Time analysis of the picture—It has been customary to work on the basis that the field scanning rate is a pre-assigned quantity and that the required bandwidth is then proportional to the number of "picture elements" in each field. The first assumption is justified in the absence of image storage at the receiver since the field rate is fixed from considerations of flicker. Increasing picture brightness makes flicker more noticeable, so that since modern television receivers give a much brighter picture than the early ones it is fortunate that the standards originally

adopted, 50 c/s in Europe and 60 c/s in America, allowed sufficient margin. For comparison, cinematograph flicker rates were 32 c/s on silent films and are 48 c/s on sound films. Synchronization with the power frequency was originally adopted (a) to permit the use of receivers with mechanical (mirror-drum) scanning and (b) to minimize the visibility of "hum bars" which are caused by inadequate smoothing of the power supplies of the receiver. Neither reason is now relevant, and the scanning in colour systems is *not* synchronized with the power frequency.

Since flicker results from brightness variation of a substantial part of the field of view, an immediate halving of the potential bandwidth requirement was obtained by adopting interlaced scanning so that two fields are required to build up a complete picture. Higher order line interlace has not been attempted in monochrome transmission, but there are some "dot interlace" schemes which span more than one line; and it is found that when the rate of re-painting a given spot on the picture is comparable with the response time of the eye there are objectionable stroboscopic or "crawling" effects. One would also expect that if re-painting of the whole picture took longer than the response time of the eye, then the representation of rapidly moving objects would be unacceptable.

It is then argued that if one has n scanning lines one must be able to resolve $r n$ points in each line, where r is the aspect ratio, so that if p is the picture repetition rate one must be able to transmit $p \times r n^2$ picture elements per second. It has been further argued (fallaciously) that one point or picture element can be approximated by a half sine wave so that the bandwidth required is $\frac{1}{2} p r n^2$. It is,

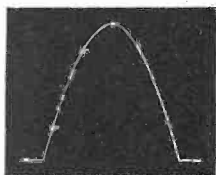


Fig. 1. A half cycle of sine wave.

however, obvious that the Fourier transform of an isolated half sine wave, Fig. 1, corresponding to an independent picture element, will include much higher frequencies as well. Gabor has pointed out⁴ that, subject to suitable definitions of effective pulse width, a Gaussian pulse gives the minimum value of $\frac{1}{2}$ for the product of time-bandwidth. A fairly close approximation to the Gaussian shape is the "raised cosine" of Fig. 2, which is the graph of $(1 + \cos x)$ from $-\pi$ to $+\pi$. Several mathematical analyses^{5,6} have confirmed that whereas the half-cycle basis allows resolution of a series of light and dark alternations, it does not suffice for an isolated change of brightness. It may be argued that in practice one is not concerned with isolated points; and empirically the

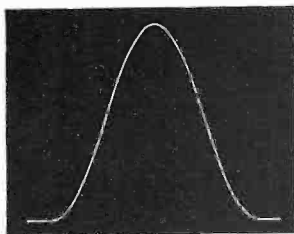


Fig. 2. A raised-cosine pulse.

required bandwidth is taken to be $(1/k) \frac{1}{2} p \cdot m^2$ where k is the "Kell factor" which is of the order of 0.7, various estimates of it ranging from 0.53 to 0.85.

Double-sideband amplitude modulation in principle involves two-fold redundancy, but true single-sideband transmission is not practicable for television because the presence of quite low frequencies and the necessity to preserve the phase of these components would make filter design an impossible task. The "vestigial-sideband" system has therefore been adopted in which one set of sidebands (usually the upper) is transmitted completely while only the lower frequencies are retained in the other. Goldman has shown⁴ that, given a limited bandwidth, it does not make too much difference to transient (step function) response where the carrier is placed, but that resolution of a repeated pattern is better if the carrier is near one end. The final decision to use the transmitter pass-band sketched in Fig. 3(a) with the complementary pass-band of Fig. 3(b) in the receiver, was influenced considerably by receiver cost.

The difference between "information" and "detail" points to the next source of redundancy in the television picture. A chequer-board pattern in which each light or dark unit occupies one picture element appears to be the most exacting test of a television system. Yet its information content is negligible, for any pre-arranged signal could have been used to turn on a pattern generator at the receiving end; or with interlaced scanning it would suffice to transmit one frequency at the edge of the video band with phase reversed on alternate frames. The latter implies square-wave phase modulation at 50 c/s, but even so the effective bandwidth is only a few hundred c/s and most of the video band remains empty. In a practical picture the detail is less regular (the example usually cited is a crowd of spectators at a football match) and the Fourier spectrum accordingly more diffuse, but usually only a part of the field of view contains fine detail.

The amount of detail can be objectively assessed by applying to the scanning waveform a voltmeter which measures the magnitude of the rate-of-change of signal voltage: *magnitude* must be used because there must be equal positive and negative slopes, so that the algebraic sum of all the changes would be zero. A "detail meter" is therefore an instrument which reads from the waveform the average amount of amplitude change in the picture, which may be expressed mathematically as

$$D = \frac{1}{T} \int_0^T \left| \frac{de}{dt} \right| dt$$

This figure occasionally reaches a few per cent of the

maximum ("chequer-board") value which it could have in the given video band.

Reduction of redundancy.—So far communication theory has taught us three things about television signals: (i) the maximum information content does not correspond merely with the highest video frequency; (ii) the transmission of an independent picture point for each half cycle of bandwidth is barely theoretically possible, and certainly unrealizable in practice; (iii) the television channel is very poorly occupied, or in the language of communication theory the signal is highly redundant. What, then, can be done to improve the utilization of the channel?

The most obvious approach in the light of the low value of D is to vary the scanning velocity in such a way that the local value of D remains approximately constant, i.e., the scanning spot must move relatively slowly where there are rapid changes in brightness and quickly over the "flat" parts of the picture. This idea was publicized in 1953^{5,6}, but has never made practical progress. The first difficulty is that the bandwidth of the control loop at the transmitter must be considerably greater than the video bandwidth. Developments in electronic components since 1953 have made this easier, but such circuits would still be costly if they had to be provided also in the receiver. The alternative is to transmit a separate velocity signal; but this would use up some of the bandwidth saved from the video signal, and still add to receiver cost. The more serious difficulty is that of ensuring synchronism of movement of the receiving scan with that at the transmitter, and this is probably an insuperable obstacle. In practice it is not permissible to allow the field repetition rate to vary with the average detail in the picture, because of problems of flicker and reproduction of motion.

A digitized substitute for variable-velocity scanning is known as "run-length coding." In this the picture amplitudes are quantized, and the number of picture elements between each successive change of level is counted. Instead of transmitting a continuous video signal, there is transmitted at each change of level a pair of signals indicating (a) how many picture elements are covered by the constant-amplitude run since the last change, and (b) what the new level is. Ideally one might transmit only the sign of the change in level (as in delta-modulation) if the resolution were good enough. But if the change from one element to the next may be more than one level, and in any case to prevent drift in level due to cumulative errors, the new level must be specified independently.

Limb and Sutherland⁷ have calculated that a bandwidth reduction of 2:1 might be obtained by run-length coding of a picture quantized into only 16 levels. (Other schemes for applying p.c.m. to television^{1,2} have suggested at least 32 levels.) The trouble is that long runs at uniform brightness are not so common: the orders of magnitude

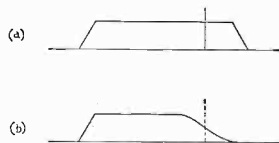


Fig. 3. (a) Transmission characteristic of vestigial-sideband transmitter; (b) transmission characteristic of vestigial-sideband receiver.

found by Limb and Sutherland for a picture of average detail were 10% of runs of length 3, 1% of length 8 and just over 0.1% of length 16. Some economy of bandwidth is obtainable if the system is made completely asynchronous, with special signals indicating end-of-line and end-of-frame¹⁰, but if the frame frequency is to be kept constant the system must be designed to meet the demands of a high-detail picture and little saving of bandwidth results. (If the system is set up for medium detail and fixed frame frequency, then a high-detail picture would not have been completely scanned by the conclusion of the frame period and the premature fly-back would leave a black space at the bottom of the picture.)

The redundancy of the television signal is most readily expressed in terms of the correlation between different parts of the same frame and between successive frames, and proposals have been made to exploit this correlation. The frame-to-frame correlation is complete for a stationary picture. It is therefore high for most pictures, in which there is a moderate amount of movement against a fixed background, though one can imagine that for a stage performance such as a Cossack dance there would be more movement than background. The engineering difficulty is that a complete frame must be stored if the correlation is to be exploited. There is also a very serious practical problem: even though the background is stationary, the picture is not constant unless the camera is fixed. This implies that all pan and zoom operations of the camera would be banned or discouraged, and sharp cuts from one scene to another would be impossible.

The fundamental difficulty is that either the communication channel must suffice for the maximum information rate of the video signal, which is the present practice, or the signal must be stored for a long enough period to allow transmission to occur at an average rate. The period needed to obtain a meaningful average is more like minutes than seconds, and though this could obviously be achieved by video-tape recording, it appears out of the question to incorporate a video-tape machine in every receiver.

The video spectrum.—A completely different approach which leads ultimately to the use of wide-band sub-carrier for colour, is via the Fourier spectrum of the video signal; and the characteristic of methods based on this is that any defect will appear as "noise" spread over the whole picture, rather than as failure to follow movement or as the complete loss of part of the picture. The nature of the picture spectrum was first analysed by Mertz and Gray¹¹, and the spectrum of television pictures has been examined by Bell and Swan^{6, 12}. The salient points are that (i) the amplitude of the Fourier components falls approximately as the inverse of the video frequency; (ii) multiples of the line repetition frequency are the dominant components; (iii) each of these line harmonics is fringed with "sidebands" spaced from each other by the frame frequency; and (iv) to the extent that there is movement in the picture, these last components spread out from lines into more diffuse components occupying to some extent the gaps between them. There are, therefore, no completely unoccupied frequencies but some parts of the spectrum are more occupied than others.

Mertz and Gray¹¹ suggested interposing other signals in the negligibly occupied parts of the spectrum of picture-telegraph signals, and this principle has been successfully used in compatible colour television. In particular, a signal exactly half-way between two frame-multiple component represents a flicker which should average out over the field period. This idea was first exploited for colour (see below) and has also been tried for bandwidth reduction¹³ according to the scheme indicated sche-

matically in Fig. 4. (The number of components in a television signal would of course be very much greater than is shown.) At (a) are shown 6 line harmonics, each accompanied by a few frame components; and at (b) the upper half of the band has been folded back into the lower half, the upper components (now shown dotted) being spaced half-way between multiples of frame frequency.

Unfortunately the eye does not remain fixed on one picture element, and wherever there is an edge in the picture the eye tends to switch from one element to the next as they change in successive frames, i.e., the observer experiences a dot-crawling effect. Consequently this scheme has not been pursued beyond the experimental stage. Dot-interlace has also been tried as a means of

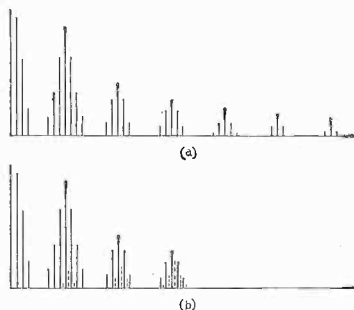


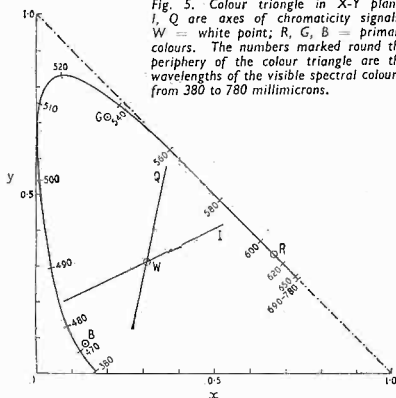
Fig. 4. (a) Schematic spectrum of television signal; (b) result of "folding" the spectrum in order to reduce bandwidth.

reducing the bandwidth required for colour¹⁴, but appears to have been abandoned.

Addition of colour.—Before considering methods of transmitting colour, it is advisable to consider what information has to be transmitted. Basically three co-ordinates must be given to specify an element of picture, and they can be represented in a three-dimensional model of which the section in the $X-Y$ plane is the colour triangle (Fig. 5) while distance parallel to the Z axis represents brightness. There are then two possible sets of co-ordinates: the brightness-hue-saturation scheme uses a kind of cylindrical polar or r, θ, z co-ordinates in which an axis parallel to Z is set up through the centre or white-point of the triangle, hue is represented by the angle θ of a radius vector to the point representing the light and saturation is represented by the length r of this vector relative to the radius to the periphery of the triangle. The alternative arrangement is to use the Z axis plus a pair of axes in the $X-Y$ plane.

It is essential in a compatible colour television system that the modulation of the main carrier should represent luminance; and it follows that the quantities transmitted on the sub-carrier must represent chromaticity (position in the colour triangle) rather than intensities of primary colours. To derive position in the colour triangle, one uses the fundamental law of colorimetry that the luminance of a "mixed" colour is equal to the sum of the luminances of its components, and normalizes the primaries having luminances say A, B and C , to chromati-

Fig. 5. Colour triangle in X-Y plane. I, Q are axes of chromaticity signals; W = white point; R, G, B = primary colours. The numbers marked round the periphery of the colour triangle are the wavelengths of the visible spectral colours, from 380 to 780 millimicrons.



city components $a = A/(A + B + C)$, $b = B/(A + B + C)$ and $c = C/(A + B + C)$. But it is obvious that $a + b + c = 1$ so one need specify only two chromaticity components; and these can be represented by the x and y co-ordinates in Fig. 5 and may be regarded as the normalized contributions from two of a set of imaginary primaries X, Y, Z such that the triangle XYZ encloses the whole of the approximately triangular space of real colours.

The television camera and reproducing systems use three real primaries; and typical primaries for a colour picture tube are specified by:

$$\text{Red } x = 0.67, y = 0.33$$

$$\text{Green } x = 0.21, y = 0.71$$

$$\text{Blue } x = 0.14, y = 0.08$$

At the transmitter the luminance signal is made up of a weighted sum of the luminances of the primaries:

$$E'_Y = 0.30 E'_R + 0.59 E'_G + 0.11 E'_B$$

(The primes indicate that the components are gamma-corrected, i.e. are ready for direct application to the colour tube in a receiver.) The two components of modulation applied to the sub-carrier are:—

$$E'_Q = 0.41(E'_R - E'_Y) + 0.48(E'_G - E'_Y)$$

$$E'_I = -0.27(E'_R - E'_Y) + 0.74(E'_G - E'_Y)$$

These three formulae can be regarded as three simultaneous equations which can be solved in the receiver for red, green and blue luminances in terms of E'_Y, E'_Q and E'_I .

Transmission of colour by sub-carrier.—The eye has less resolving power in colour than in luminance (brightness) just as it has less sensitivity. (The loss of colour discrimination at low illumination is known as the Purkinje effect, and moonlight falls within the intensity range in which there is no colour discrimination.) An example of the eye's insensitivity to detail resolution in colour can be seen in the remarkably satisfactory impressions given by monochrome photographs which have subsequently had colour superimposed by hand; the areas of uniform colour are often much larger than would have been acceptable in a painting. The principle of "mixed highs" was proposed to take advantage of this, the colour components being separated at low frequencies

but allowed to mingle as a combined white signal at high video frequencies. In Dome's proposal¹³, using two colour sub-carriers, the basic signal to 4 Mc/s was to have been the green component while the red component extended to 1 Mc/s and the blue to 0.2 Mc/s. Since the red and blue signals were not mutually overlapping in the combined video signal, the space that could be allocated to each was strictly limited. In the N.T.S.C. system the two colour signals occupy the same band (see below) so the bandwidth limitation is not so stringent.

In all compatible transmission systems which have so far been proposed, the first step is to provide a colour sub-carrier at a frequency where it will cause least interference with monochrome reception. The three factors which are used to minimize interference are: (i) the sub-carrier is placed mid-way between frame-frequency multiples, so that its effect should average out over alternate frames; (ii) it is placed near the upper end of the video signal, where luminance components are small so that low-frequency patterns will be of low intensity and beating with the strong components at the bottom of the video band will produce fine-grained patterns; (iii) the average amplitude of the sub-carrier is kept down; and (iv) the bandwidth of the modulation of the sub-carrier is kept down.

Compression of both colour signals into one frequency band is achieved by using double modulation of one sub-carrier: the sub-carrier is regarded as a two-phase channel and the two colour components are modulated on the sine and cosine components of the sub-carrier. The two colour signals are then known as the "in-phase" (I) and "quadrature" (Q) components. (This is equivalent also to having the carrier simultaneously modulated in amplitude and phase.)

It is specified in the N.T.S.C. system that the E_I component, sometimes known as the hue component, should have a coverage of at least 1.3 Mc/s, but must be at least 20 dB down at the main vision carrier. The E_Q component must cover at least 400 kc/s but have at least a 6 dB cut-off at 600 kc/s. Since the sub-carrier is placed at approximately 3.6 Mc/s in a nominal 4 Mc/s channel, E_Q may be approximately double-sideband but E_I must be vestigial sideband.

The separation of the I and Q components in the receiver requires an accurate phase reference; and even with a burst of sub-carrier added to the synchronizing signal at the end of every line, the receiver must be set up rather carefully to avoid colour distortion. Several schemes have been devised to overcome this phase sensitivity, known as SECAM¹⁶ (from France), PAL¹⁷ (from Germany), and a scheme tentatively known as NIR or SEQUAM¹⁸ (from Russia). All make one line of colour information serve for two lines of picture, and therefore require the receiver to incorporate a delay line capable of storing one complete line of signal.

In the SECAM system the two colour signals are sent alternately, a line at a time, and each received line therefore takes one of its parameters from the current modulation of the sub-carrier and the other from the stored signal of the previous line. Since only one parameter is transmitted at a time it suffices to amplitude-modulate the sub-carrier and there are no demodulation problems. The difficulty is that this very simplicity of demodulation makes SECAM completely incompatible with N.T.S.C. The PAL system is very similar to N.T.S.C., but the same video information is used for the in-phase component on two successive lines, but with reversal of phase of modulation in alternate lines. (E_I tends to convey hue, rather than saturation, because it is based on the difference between red and blue

signals. It is E_0 which has the smaller bandwidth of the two in N.T.S.C.) It follows that E_1 vanishes from the sum of successive lines, leaving E_0 , while similarly E_0 alone is recovered from their difference. If a phase-sensitive detector is used, an error ϕ in the reference phase will only reduce the amplitude of the colour signal by $\cos \phi$.

It is one of the practical difficulties of communication theory that one has no theoretical basis for deciding the fraction of the channel capacity that should be devoted to synchronization of one sort or another; one only knows that in practice it is an appreciable fraction. With normal teleprinter machine telegraphy, for example, it is about 25%; and in monochrome television the combined line and frame synchronizing pulses (and flyback times) account for about 20% of total signal time. It appears that the Russian NIR system devotes 50% of colour sub-carrier to synchronization by the device of transmitting sub-carrier reference phase throughout the whole of alternate lines (instead of only as bursts during synchronizing pulses) and then uses the N.T.S.C. double-modulation of the sub-carrier for the other lines. A one-line delay store is thus able to bring together simultaneously the modulated sub-carrier and the reference oscillation, instead of having to rely on the "memory" of a local oscillator to provide the same phase during the line of picture signal as was indicated during the preceding synchronizing pulse.

Having once accepted the idea of storing a line of colour signal, one has only two signals to transmit simultaneously, namely luminance and one or other of the two colour signals. But two signals can be transmitted on one carrier, and the SEQUIN system¹⁶ proposes to transmit the two as in-phase and quadrature components of the main carrier. The proposal includes the use of negative modulation, and saturated colours would be represented by both amplitude and phase changes of the carrier, while unsaturated colours would differ from the corresponding saturated colours by having smaller quadrature components but the same in-phase (black-white). With negative modulation the synchronizing pulses provide maximum-power pulses of carrier, and therefore a carrier phase synchronizing reference; an automatic phase-lock circuit is suggested; and this appears desirable because the frequency of the main carrier is some twenty to a hundred times greater than that of the conventional colour sub-carrier, the phase-synchronization of which is apt to cause trouble.

On grounds of communication theory, the objection to SEQUIN is that it puts the colour information into the low-frequency end of the spectrum which is already heavily loaded: in engineering terms this means that when the carrier is subjected to double modulation there must be limits on the depth of modulation of either component. In the proposed system this is covered by the reduction in amplitude of components which involve a large angle modulation.

Other possibilities.—So far we have reviewed existing developments in the light of information theory, but if we could wipe the slate clean and start again would we make any drastic changes? The most important factor seems to be that the picture is so variable in amount of detail, in position of the centre of interest in the field of view* and in amount and speed of movement that the only alternatives are either to provide for the most severe conditions on instantaneous transmission, and so have a poorly utilized channel for most of the time, or average

*It has been suggested that only the centre of interest of the picture should be scanned in full detail: a spiral scan could be made to give fine detail in the centre of the picture and coarser coverage of the outer parts.

over such a long sequence of fields that video tape recording seems the only available means of storage.

On the choice of modulation for radio transmission, there are ample theoretical studies showing that the present vestigial-sideband system is a good compromise between the minimum redundancy of a single-sideband system and the requirements of simple receiver design. From the point of view of monochrome signals only, communication theory suggests that most of the video band should be subjected to a pre-emphasis of nearly 6 dB per octave, so that the signal transmitted should have a flat spectrum, rather than a falling one. Looked at in another way, this is approximately equivalent to differentiating the video signal before transmission, to be followed by integrating in the receiver, which (a) in the transmitted signal gives prominence to the edges and boundaries which convey most information and (b) in the receiver reduces random noise. However, this would be incompatible with present colour systems, which only work because the monochrome spectrum is of low intensity at the frequency of the sub-carrier.

There is a degree of asymmetry in present colour systems, since horizontal resolution is reduced by a factor of two or three for E_1 and about six for E_0 , but vertical resolution is unchanged for colour in N.T.S.C. or reduced by two in SECAM and PAL. Since flicker is such a powerful factor in setting frame frequency, one wonders whether whole fields could somehow be interlaced in colour. Obviously the transmission of a colour signal cannot be delayed for a whole frame, since this would lead to colour trails behind moving objects, so one then comes to the dot-interlace scheme of colour television¹⁸. Just as line interlace is normally used, so each line can be broken up into groups of say three dots, which are transmitted in only one colour at a time, red, green or blue. Thus if one used in three successive scans the sequences RGB, GBR and BRG, super-position of the three scans would give all three colours for all three points.

The main trouble is that in this form the system makes use of the wrong tolerance in colour: it reduces bandwidth by allowing the formation of colour to be extended in time, instead of by degrading the spatial resolution in colour. It is also rather inflexible in the sharing of bandwidth between the three colours; unequal allocation is possible by giving two scans of green to one each of red and blue, but only such integral proportions are possible. The practical stumbling block is that dot systems are not immediately compatible with simple monochrome reception, and the more one refines the dot system the worse is the compatibility problem. To set against the pattern which can be created by the sub-carrier in the N.T.S.C. system, there is the break-up of the picture into dots, though this is theoretically at a higher frequency.

One is forced to the conclusion that although sub-carrier colour systems have no fundamental merit, and work only by virtue of the inefficient use of the channel by monochrome signals, they have great practical merit. Whether communication theory will guide anyone to a system having sufficient advantages to justify abandoning direct compatibility with existing receivers must remain an open question.

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LITERATURE RECEIVED

Two charts entitled **Oscilloscope Accessories and Mounting Arrangements for Oscilloscope Cameras**, are available from Marconi Instruments. These are intended as guides to accessories for use with Marconi scopes and, in particular, oscilloscope cameras.

WW 322 for further details

Oscilloscope input accessories is the title of another Marconi Instruments' publication (10 pages) introducing the 81 series of passive probes, and also discussing an active probe and a pre-amplifier.

WW 323 for further details

The **Lofting photomechanical process for reproducing drawings and graphic designs on flat surfaces**—all metals, laminates, plastics, and so on—is described in a 4-page leaflet available from Lee-Smith Photomechanics Ltd., of Lyon Way, Hatfield Road, St. Albans, Herts. The industrial applications of this process and details of the equipment needed for the process are included.

WW 324 for further details

"Abstracts of application notes and other Literature on Semiconductors available from General Electric Semiconductor Products Department April 1965" is the title of a publication (90.0/200.0) listing the company's semiconductor application notes, article reprints, manuals and papers. Requests should be made to the General Electric Company, Distribution Services, Building 6, Room 208, 1 River Road, Schenectady, N.Y. 12305, U.S.A.

WW 325 for further details

Newmarket Packaged Circuits ABC describes the range of packaged a.f. amplifiers and power units produced by Newmarket Transistors Ltd. Seven types of a.f. amplifier are presented, from 150 mW to 3 W, and circuits, dimensioned drawings and specifications are given for each item. A high-to-low impedance matching pre-amplifier and three power supply units are also detailed in this 12-page brochure.

WW 326 for further details

The second issue of the 312-page main catalogue of Brüel and Kjaer, of Naerum, Denmark, is divided into 14 sections, the largest covering special acoustic equipment. Others include a.f. generators, frequency analyzers, microphones and accelerometers. The final section contains abbreviated information on international standards for acoustical and mechanical measurements.

WW 327 for further details

The latest **Heathkit catalogue (86/1)** is available from Daystrom Ltd., of Gloucester. The catalogue lists available kits, including two recently introduced items—a "salamine" loud-speaker enclosure and the OS-2.3 in oscilloscope. A short leaflet listing some of the available American kits is also available.

WW 328 for further details

Crossed field amplifiers are discussed in the latest issue of *Micronotes* (vol. 3, no. 8). This American publication, intended for those interested in microwave technology, is available from Microwave Associates Ltd. at Cradock Road, Luton, Beds.

WW 329 for further details

Autospec Success is the title of publication SP149 available from the Marconi Company. The Autospec principle of automatic error correction for radio telegraph and data transmission systems is described briefly and examples of its use in various installations are given, together with some typical error-rate figures. The Autospec system was described in detail in *Wireless World* February and March 1964 issues.

WW 330 for further details

Transect Variable Ratio Transformers is the title of a 6-page folded leaflet describing a new range of transformers manufactured by Smith Hobson Ltd., Hershams Trading Estate, Walton-on-Thames, Surrey.

WW 331 for further details

British Insulated Callender's Cables Ltd., 21 Bloomsbury St., London, W.C.1, announce a new publication No. 508, **Television Distribution Cables**. Physical and electrical data of nearly forty different conductor sizes is presented in three tables. General supporting details cover cable construction, installation, screening efficiency and impedance uniformity.

WW 332 for further details

An engineering bulletin on the subject of 0.01% **A.C. Calibration Equipment** is available from Dynamco Instruments Ltd., Salisbury Grove, Mychett, Aldershot, Hampshire. The bulletin describes the general problems associated with a.c. calibration, drawing attention to the fact that there is no absolute standard for a.c. similar to the Weston Standard Cell used for d.c. calibration, and then details the method of the Dynamco system which uses the thermal transfer technique.

WW 333 for further details



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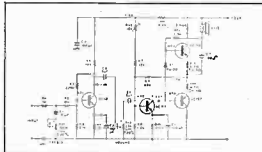
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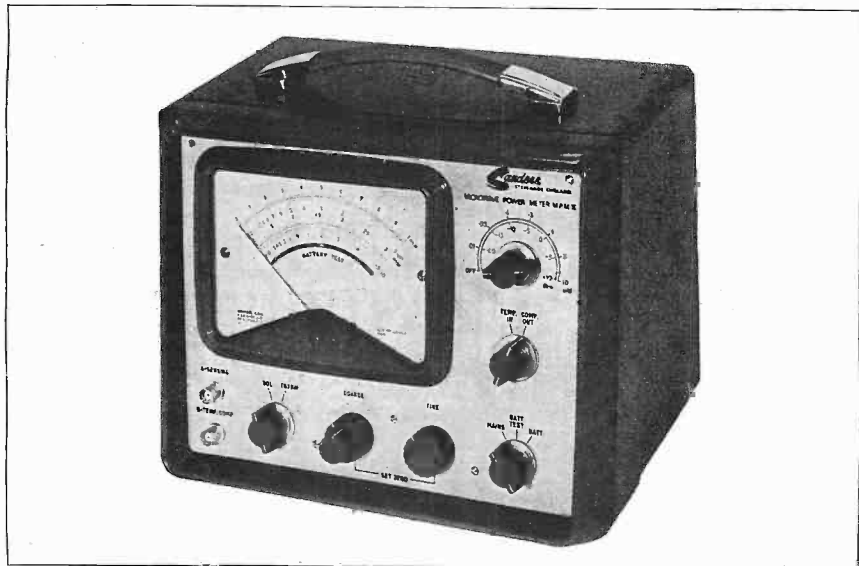
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LETTERS TO THE EDITOR

The Editor does not necessarily endorse opinions expressed by his correspondents

Collision Avoidance Systems

I READ with interest Mr. Gilmour's letter (April issue) about collision avoidance systems for aircraft. His theories, however, contain some serious defects:

(a) For aircraft climbing and descending (a prominent collision situation), he proposes the aircraft transponder to transmit "heights intending to be passed through." Two points here. First, the transponder has no crystal ball to foretell this, so it has to be informed of the intention by the already overworked pilot. Something else to do wrong. Second, the rate of descent is all-important, as well as the heights concerned.

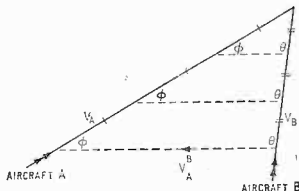
(b) The successful operation of the equipment depends on complete serviceability of both units in both aircraft. Things like this go "on the blink" with monotonous regularity. One fault in either system and all is useless.

(c) Also assumed is that both aeroplanes have the equipment fitted. Some small airlines have not yet had I.L.S. installed in their aeroplanes, a system years old and widely considered vital for safe air operations. What chance is there that they will fit this new wizardry? Also, a light aeroplane can do just as much damage in collision as another airliner; yet, with the best will in the world, these do not have room to fit the proposed equipment.

(d) Aircrew cannot tell readily from a height/airspeed/heading display whether or not another aeroplane constitutes a collision risk. They want it on a yes/no basis. Given that, which implies more than the "simple modification" proposed by Mr. Gilmour, it is impossible to "discuss avoiding action" with another aeroplane owing to the diversity of frequencies in use.

A better basis for starting to look at this problem is that of simple relative velocities.

Two aeroplanes on a collision course have velocities such that their relative velocity is along the straight line joining themselves. Let me make this more clear diagrammatically.



This is the old "parallel-axes" theorem; the relative velocity (represented by the intercepts) is always at a constant angle (i.e. no angular component of polar velocity) if the flight speeds are steady.

The following system therefore commends itself. A self-contained radar tracking system in the aircraft (of the gyro-based guided missile variety) picks up an aircraft within a certain range and "locks on" to it using proven missile techniques. The rest is simple. If the radar

aerial moves transversely (i.e. spherically) in following the stranger, the two aircraft are designed to miss each other. If the aerial stays geometrically stationary in following the other aeroplane the relative velocity is through the two aircraft with no angular component, and a collision course exists.

Warning of an impending collision could either be presented to the pilot or arranged to tweak the autopilot so that the relative velocity changes into one with an angular component subtended at the aerial and collision is avoided. The presentation of range and velocity information to the pilot as a bonus presents no problem, using the Doppler effect mentioned by Mr. Gilmour.

Where several aircraft are within range the locking radar would investigate each in turn, the nearest first. Once the aerial has locked on and moved while following its target it has satisfied itself that no danger exists and goes on to the next trace. This would only take a few seconds; no problem at 100 miles range even at closing velocities of 1,200 m.p.h.

This system, though idealistic, has none of the immediate snags prevalent in Mr. Gilmour's—

(i) The system is positive for all straight line flying, and takes account of rates of climb and descent.

(ii) The system is entirely self-contained. One's own careful airline has no need to trust the servicing and operating integrity of a small airline operating across the Atlantic at the "lowest fare ever."

(iii) If both aircraft have the system, one unserviceable is still a safe state of affairs.

(iv) The pilot is not bothered until something threatens, which is as it should be.

In conclusion, I am delighted to see this subject coming up in other than aeronautical circles. The evolution of the supersonic transport and proposed narrower air corridors demand the rapid development of such a device.

Officers' Mess,
R.A.F. College, Cranwell

G. L. PERRY

British Electronics Abroad

AS an electronics engineer and a resident in Hong Kong for the last six years, I wholeheartedly agree with your Editorial in the December issue and the subsequent letter from Mr. D. W. F. Milligan in the February issue.

We have just finished here in Hong Kong what is reputed to have been a very successful British Week. In conjunction with this, an eleven-day Engineering Exhibition was held to show off the cream of British products. After spending a couple of hours looking around the various displays, I concluded that British manufacturers of electronic equipment either had nothing to show, or they were just not interested. As I read most of the leading electronics journals and visit the various London exhibitions, I know that it is not the former—therefore I must assume it is the latter.

Of course, some of the larger British companies with Far East offices in Hong Kong were represented, but even they had nothing spectacular or inspiring on show.

One point in particular; although within 18 months Hong Kong is to have a broadcast television service, not one manufacturer of TV sets thought his products worthy of showing.

In the manufacturing field, Hong Kong is blessed with a large labour force and a low wage scale. This, together with a low rate of taxation, has attracted many American and Japanese manufacturers to set up their own or joint-venture plants in the Colony. In contrast, only one British company has followed suit—it is a subsidiary of a large American organization!

One American company started two years ago producing semiconductors on one floor of a small factory building. Last week they opened the doors of their own new 11-storey factory.

The general opinion of the industry here is that British manufacturers have "missed the boat"; an opinion which I am almost ready to endorse.

However, before passing judgment I decided to do a little research. I looked around this wonderful Colony of ours and concluded that one branch of electronics was still open to all. I refer to the field of industrial electronics, where none has so far ventured, though no doubt they will if left much longer.

As a start, a small group of electronics engineers have just formed a company for the design and manufacture of electronic devices for use in industry. Within ten days of advertising the fact, we have been swamped with queries from local manufacturers; so many that two of our group are permanently assigned as consultants. If the response locally is so good, then what better place to start a plant. No doubt the opportunities are as great in the other countries around us.

So, you manufacturers back home, why not take a trip out here and look over the market potential. I assure you the natives are friendly. Should manufacture be out of the question for you, there are sales to be made—but not from the sanctuary of your U.K. office.

Kowloon,
Hong Kong.

DON WEBSTER

"Flip-flop" a Misnomer?

THE letter from Mr. P. S. Pinder in your April issue can only produce my complete agreement, but what a shame that the printers let him down at a most crucial point in his argument.*

There is, however, doubt creeping into my mind. Perhaps the computing engineer uses the terms "flip" and "flop" to designate a "1" or "0" respectively being generated. In this case one can, I suppose, argue that their nomenclature is logical as this is the pattern of events occurring in a bistable, i.e., the first trigger pulse generates a "1" and the second pulse a "0." Presumably a bistable working in an "inverted mode" would then be called a "flop-flip"!

However, I will come down on Mr. Pinder's side—after all, why should the bistable be so privileged above members of its own family?

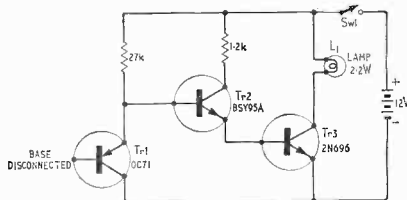
University of Keele,
Staffordshire.

E. W. FIRTH

*The fourth line from the end of Mr Pinder's letter should of course have read "a bistable a 'flip-flip' and an astable a 'flop-flip'."—Ed.

Automatic Car Parking Light

READERS may be interested in details of a circuit for an automatic car parking light which makes use of the light-sensitive properties of the OC71 when its point



is scraped off. The base is not connected. The circuit was designed as shown in the figure and field trials carried out on different makes of cars to prove its reliability. At night maximum drain on the car battery is approx. 196mA. When the device is inoperative the maximum drain is 0.5mA unless of course a switch is incorporated to disconnect it from the battery. The lamp is automatically switched on at dusk and switched off as daylight returns. The lamp stays alight in street lighting. The components cost about £1.

The circuit analysis is as follows:—

$$\begin{aligned} &\text{maximum current taken by the 2.2W lamp} \\ &= \frac{2.2 \times 10^3}{12 - V_{SAT}} = \frac{2.2 \times 10^3}{11.7} = 188\text{mA} \\ &\text{current required to saturate Tr3} \\ &= \frac{188}{\beta(\text{min}) \text{ of Tr3}} = \frac{188}{20} = 9.4\text{mA} \end{aligned}$$

As Tr2 and Tr3 form a darlington pair and Tr2 remains saturated when circuit is in use, therefore current in Tr2 = $\frac{12 - V_{SAT} - V_{be}}{1.2} = \frac{12 - 0.3 - 0.7}{1.2} = 9.2\text{mA}$

This current is enough to saturate Tr3 which requires $\frac{9.2}{\beta(\text{min}) \text{ of Tr2}} = \frac{9.2}{30} \approx 0.3\text{mA}$.

Current of about 0.45mA is enough in the 1st stage. For Tr2 to saturate V_{b1} must rise to at least 1.6V with respect to negative earth. The impedance of Tr1 is about 3.5k in the dark. Therefore $V_{b1} = 0.45 \times 3.5 = 1.57$ volts. The values of resistors can be varied to suit requirements.

Tr1 is conveniently placed near the windscreen ensuring enough light falls on its glass covering. The paint on the glass covering is easily removed. If the car is to be used regularly S1 may be omitted as drain on the battery during daytime is negligible.

Basildon,
Essex.

S. K. CHAWLA

Temperature Control

I WAS interested in Mr. J. A. Selby's thermostatic switch unit in the February issue. Perhaps a circuit I have used in a number of versions may be of interest. It is basically an a.f. oscillator, of which the feedback is through a Wheatstone bridge with a thermistor as one of the resistors. Fig. 1 is the simplest practical version. Oscillation takes place above or below a certain temperature, depending on the way the transformer is connected, and the oscillator output is rectified and amplified to operate the relay.

Using an F type thermistor made by S.T.C., this circuit will control a water bath to $\pm 0.5^\circ\text{C}$ or better. The

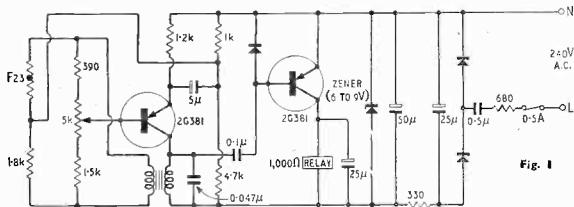


Fig. 1

resistors in the bridge should be wire-wound or metal oxide types, and their values depend on the temperature range required and the thermistor type. (The resistance of the FZ3 is $2,000\ \Omega$ at 20°C .)

There has been no trouble with the relay not switching over positively during gradual change of amplitude with temperature change.

An interesting variation is a proportional control version using a complementary monostable multivibrator as in Fig. 2. Over a small temperature range the on-off

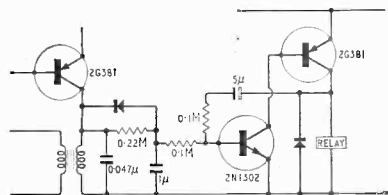


Fig. 2

ratio changes from zero to infinity. The recovery time is two or three seconds with the capacitor shown. Positive bias for the n-p-n transistor comes from the voltage drop of about 2V in the transformer winding.

The transformerless power supply shown calls for some comment on safety. The live mains wire should be connected to the capacitor, which must have an adequate rating (1,000V d.c.). The series resistor is to reduce transients, particularly when switching on. This is essentially a constant current supply and must not be open circuited (when it would become a voltage doubler), and the Zener diode must always be connected.

Bexhill-on-Sea,
Sussex.

C. Q. KEILLER

Television Sound Quality

SOME time ago I decided to improve the audio quality from my television-receiver in a similar manner to that described by Mr. T. M. George in your February, 1966, issue, but found one or two drawbacks. Most audio transformers proved unsatisfactory from the mains isolation point of view, since their insulation had not been designed to withstand 240V r.m.s. A much safer method is to isolate the receiver completely using a 1:1 mains isolating transformer*.

It is then a simple matter to bring the audio out from

* Make sure the mains supply is a.c.—ED.

the same point as used by Mr. George. If this is fed into a conventional valve pre-amplifier, and the interconnecting lead is reasonably short, there are no impedance problems to worry about. When one is using a low impedance input, or the leads are long, it is relatively simple to re-arrange the first audio stage in the television receiver to operate as a standard cathode follower. By doing this one avoids all the inherent distortions encountered

using transformers, and reveals even more clearly the quality (sic) of the transmission.

Stroud,
Glos.

JOHN WEBSTER

"Amplitude-stabilized RC Oscillator"

I WAS very interested in Mr. E. Nelson-Jones's article on the phase-retard oscillator in the November 1965 issue.

I would like to point out that a high input-impedance amplifier is not essential with such oscillators provided that an additional resistor is included in the network, as shown below.

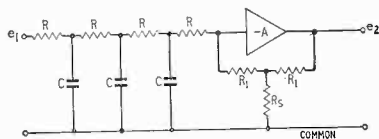
It may be shown that if the amplifier has zero input impedance (i.e. virtual earth) the fundamental frequency of oscillation is given by

$$f = \frac{\sqrt{10}}{2\pi CR} \left\{ \text{c.f. } \frac{\sqrt{6}}{2\pi CR} \right\}$$

and the amplifier output voltage by

$$e_o = \frac{1.274e_1}{56R} R_1 \left(2 + \frac{R_1}{R_2} \right)$$

where e_1 is the sine wave peak-to-peak amplitude and e_o is the square wave peak-to-peak amplitude. This output amplitude may readily be adjusted by varying R_1 .



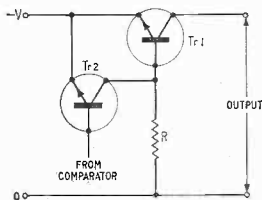
By using an operational amplifier a high open-loop gain may be achieved together with a low output impedance, thus ensuring a true virtual earth input and eliminating the need for a buffer between oscillator and load.

Basingstoke,
Hants.

J. W. HARMAN

Voltage-regulated Power Supplies

IN most published designs for voltage regulated power supplies, including that by Mr. J. S. Taylor in the April issue, the output transistor is driven by another power transistor. This method is expensive in that the driver transistor, often capable of dissipating about 30W, is only required to dissipate around 1W.



The circuit given above overcomes this objection. Tr1 is a silicon power transistor while Tr2 is a germanium device operating in the common emitter mode with a V_{CE} of about 0.7 V.

The advantages which may be claimed for the circuit are:—

- (i) Maximum power dissipation in Tr2 is normally low (usually ~ 100 mW) and is limited by R.
- (ii) A current limiter of the type described by Mr. D. Wilson in the Correspondence pages of the February issue may be readily incorporated.
- (iii) Tr2 is always operating in a region of substantial collector current. Leakage is therefore unimportant.
- (iv) The possibility of a high Tr1/Tr2 h_{FE} product is offered.

(v) The components required are inexpensive. At present a 2SO12A/2N1304 pair with a minimum h_{FE} product of 800 costs only about 15s and has very good power-handling capabilities.

The circuit's only disadvantage appears to be continuous dissipation in R but this is usually easily accommodated since no extra heat sink area is required.

Aberdeen. IAN H. HOWIE

Camera Image Stabilization

WHEN a television camera is used to take pictures of very distant objects with a telephoto lens, the optical effects of small camera movements caused by accidental shocks or vibration are greatly magnified and become noticeable as wobbling or jumping of the picture. This can be seen sometimes in outside broadcasts of sporting events. The B.B.C. are considering tackling the problem by fitting to their O.B. cameras a device which optically compensates for such movements—that is, movements more rapid than the normal panning ones. Known as the Dynalens, it comprises an adjustable prism placed in the optical path and gyrosopic displacement sensors mounted on the camera housing. A camera movement detected by the sensors produces a signal which alters the refraction angle of the prism so that the light rays are deflected in the same direction as the movement and the selected image is therefore not displaced from its original position on the camera tube target.

The prism is a lens-like structure comprising a volume of liquid enclosed between two transparent plates and a bellows. In normal operation the two plates are parallel and no deflection of the light rays occurs. When a displacement signal is received, however, the plates are moved at an angle to each other and the emerging rays are bent correspondingly. The plates are controlled about both horizontal and vertical axes so that the effects of camera movements in any direction can be compensated.

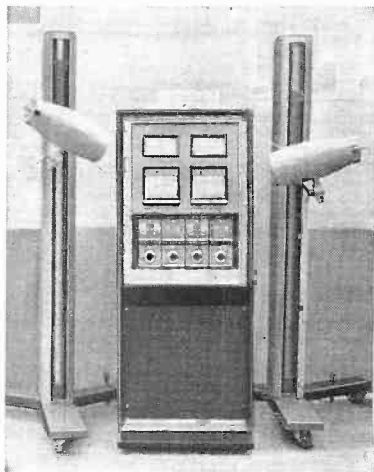
The device is marketed in the U.K. for the American

manufacturers, Dynasciences Corporation, by the Livingston Group. At present the B.B.C. are assessing its performance, which may depend to some extent on the speed of response which in turn depends on the inertia of the mechanical parts. It is thought that the stabilizer may prove particularly useful when cameras are mounted on high towers where they are likely to be buffeted by the wind.

An alternative solution to this type of problem adopted by another American firm, the Itek Corporation, is to interpose in the optical path an electronic image converter with electro-magnetic means for deflecting the electron beam (*Electronics*, September 20, 1965). This has the advantage that the electron beam deflection system has no mechanical inertia to limit the speed of response of the stabilizer.

New Aid for Renal Studies

KIDNEY function studies employing radioactive isotope tracers will be assisted by a specially developed renal function analyser with two separate channels. This equipment, made by Isotope Developments Ltd., employs two scintillation counters, which are so positioned on two vertical columns that the patient can be examined in a sitting position. The associated collimators, with special iso-count characteristics, are intended to minimize cross interference caused by the radiation emitted from each kidney. Both analogue counting channels are constructed of transistor modules, and each channel contains a high voltage unit with digital setting indication, a wideband amplifier with continuous gain control, a pulse height analyser, and a multi-range linear ratemeter. The mounting of the detectors permits other studies to be carried out, with the patient lying down or sitting up.



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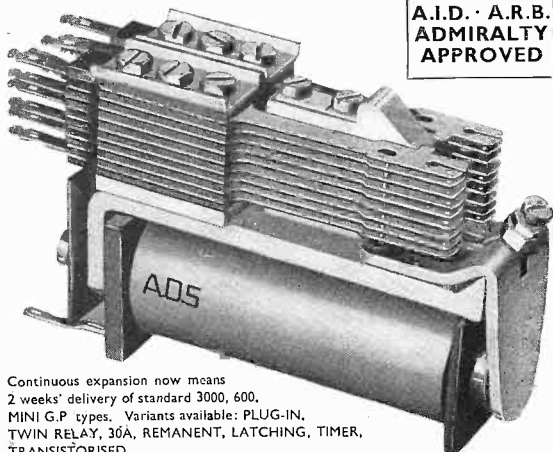
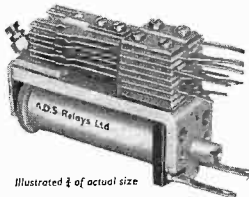
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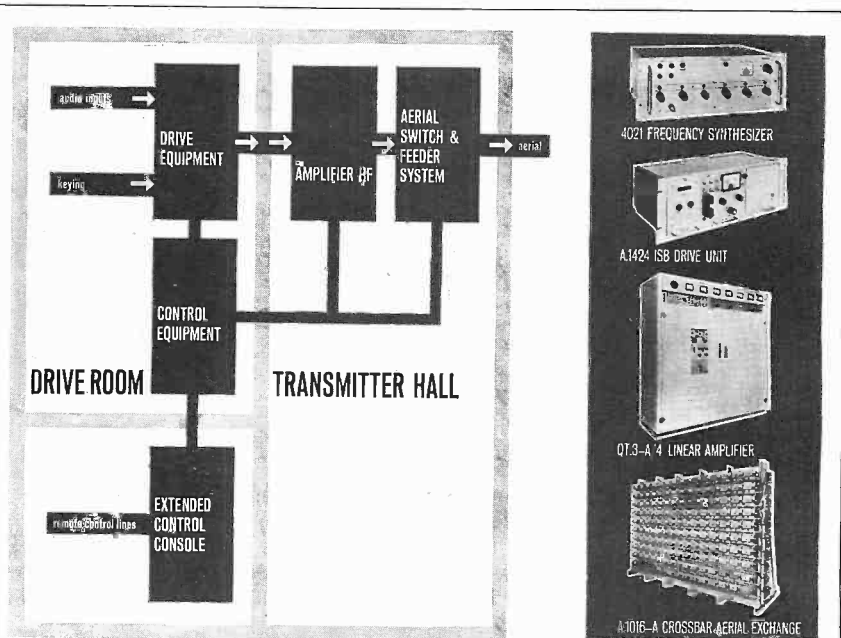
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5-Experiments and Decisions

(Concluding the series)

By D. A. BELL,* M.A., B.Sc., Ph.D., M.I.E.E., F.Inst.P.

DECISION-MAKING is an important activity in government, management and scientific activities; and from a theoretical viewpoint two of the major characteristics of decision-making are, first, that it involves loss of information and, secondly, that the task of scientific decision-making is to make logical decisions on the basis of doubtful evidence. The loss of information is inherent in the non-linear nature of decision-making, whether or not it is irreversible. Consider, for example, a criminal trial in which the evidence for the prosecution is largely circumstantial. Tens of thousands of words may be spoken about times and places, alleged recognition of persons, chemical analyses, etc., but the decision of the jury reduces all this to one binary digit's worth: "guilty" or "not guilty" (except in Scotland, where the verdict provides one ternary digit of information on the scale "guilty," "not proven" or "not guilty"). The discarding of information by decision-making is perhaps illustrated even more dramatically by the tradition in certain examinations of destroying the candidates' scripts as soon as the class list has been prepared.

It is equally true that there is a loss of information when a decision is made by a threshold circuit between the "mark" and "space" conditions of a signalling circuit. It is inherent in the philosophy of Shannon's ideal coding (as described in the geometrical model¹) that in order to preserve as much information as possible until a single decision shall be taken, identifying the received signal as a whole with some particular message, and the freedom from error which is associated with the ideal-coding theorem is lost if one decodes the signal one digit at a time. This viewpoint is also applicable to character-recognition schemes: those which divide the character area into a matrix of cells and then take a separate binary decision on each cell in turn must be inferior to those which retain all the information until a later stage.

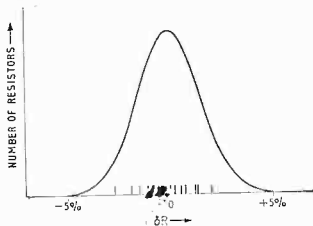
"Doubtful" evidence means statements which can be expressed as probabilities but not as facts; an example from electronic engineering is the estimate of the magnitude of a signal or other current on which is superimposed "noise" of some kind. The decisions are described as *logical* because the use of the appropriate statistical algorithm will lead to a unique result which does not involve any subjective judgment by the observer; and this is both the strength and the weakness of statistical decision methods. The absence of subjective factors is in general desirable, but there may be anxiety lest the statistical process might ignore "imponderable" factors which would be included in a subjective judgment but which are not amenable to quantitative formulation. The answer to this objection is that diligent study will often lead to a quantitative formulation of the so-called imponderable, in which case all that was hurt by the use of statistics would be the pride of the man who would otherwise have made a subjective judgment.

Experiments and decisions have been linked together because the purpose of an experiment is often to obtain information for use in making a decision: this can usually be described formally as deciding on the best out of n number of possible hypotheses. In measuring a length of x millimetres with a metre rule, you select one out of the thousand hypotheses from $x = 1$ to $x = 1000$. However, it would usually be unrealistic to suggest that all the different values are equally probable: for example the length of a man's shoe is unlikely to be more than 35 cm or less than 20 cm. One thus has some advance estimate of the probabilities of the various hypotheses—the *a priori* distribution of probabilities—and the experiment leads to a more restricted *a posteriori* distribution of probabilities. The difference between these two sets of probabilities, when suitably expressed on a logarithmic scale, is the *information* obtained from the experiment, in the particular sense of information/communication theory. Suppose that in given circumstances there are k probabilities to consider. (We might have $k = 57$ if we were waiting for the next symbol to appear on a teleprinter page, or perhaps $k = 9$ for the number of horses in a particular race.) Then the average logarithmic probability is obtained by multiplying the logarithm of each probability by the probability of its occurring, i.e. by itself. Since probabilities are by definition less than unity, their logarithms are negative quantities; and the positive quantity obtained by reversing the sign of the average logarithmic probability is the *entropy*

$$H = - \sum_{i=1}^k p_i \log p_i$$

Thus the whole basis of information measurement is statistical.

However, the application of statistics is more obvious when a particular statistical technique is used to reach a



The short vertical lines indicate by their positions along the scale of R the differences from nominal value of fifteen resistors. It is reasonable to assume that the bulk quantity of resistors, from which these fifteen were drawn, have the gaussian distribution centred on the correct value, which is represented by the curve?

*University of Hull.

decision on some specific type of question. For example, a sample of 15 resistors drawn from a large batch of 5% tolerance resistors has a mean value 0.22% below the nominal value: does this mean that the whole batch is probably a little low in value? To answer this question the variances of the whole population and of the sample must be known, because such a sample could be from normal production in the way sketched in Fig. 1. The curve, a normal distribution with a standard deviation of 1.5%, represents the distribution of frequency of occurrence of various magnitudes of discrepancy in the whole population, and the small arrows indicate the individual values of the 15 samples. The samples occur more thickly towards the centre of the distribution, but being random are not quite symmetrically distributed. Are these values which could reasonably have been drawn as a random sample from the population? The parameter used by statisticians to investigate this question is known as "student's t " and is defined by the formula*.

$$t = \frac{(\bar{x} - \bar{X})}{\sigma_s} \sqrt{n} \quad \dots \quad \dots \quad \dots \quad (1)$$

where \bar{x} is the mean of the sample, \bar{X} the mean of the whole population, n the number in the sample, and σ_s is related to the variance of the sample:

$$\sigma_s = \frac{\sum(x - \bar{x})^2}{n - 1}$$

The quantity $m = n - 1$ is said to be the number of "degrees of freedom" of the sample parameter σ_s ; one degree of freedom has been lost because instead of asking about the variance of the sample, the mean being fixed by the sum of the individual values in the sample, we are asking about the variability of the sample relative to the pre-determined mean of the whole population. (The idea of "degrees of freedom" will be further discussed below.)

Returning to the specific example, $\bar{x} - \bar{X} = 0.012$, $n = 15$, and $\sigma_s = (1/14) \sum(x - \bar{x})^2$ which for the sample values indicated in Fig. 1 is 1.701. It follows that $t = 0.508$. Now look in a table of the t distribution (e.g. Table XII in *Burington and May*), take the line of the table for $m = 14$, (14 degrees of freedom): the nearest entry, 0.537, occurs in the column headed 0.60 which means that there is a nearly 60% probability that the difference is random and no change in the characteristics of the production is indicated by the sample.

Degrees of freedom

The meaning of "degrees of freedom" is most easily seen in a "contingency table" which is a means of discerning relationships, similar to correlations, when the relevant characteristics are given only qualitatively instead of as continuous variables. For example, is it more difficult to get an Ordinary National Certificate in Engineering or in Building? The numbers enrolled and numbers obtaining O.N.C. at certain colleges could be set out in a table as follows†. The non-statistician would convert these figures to percentages passing, which is quite correct as far as it goes. But when the percentages work out to 49 and 54, are we sure this is significant?

The statistician may take a different approach. The total number of enrolled students is divided in two ways, first between those who take engineering and those who

	Obtain O.N.C.	Do not succeed	Total enrolled students
Engineering Building	539 (542) 54 (51)	561 (558) 48 (51)	1100 102
Totals	593	609	1202

take building and secondly between those who obtain O.N.C. and those who do not. If there is no difference in standard of courses or quality of students, the number who obtain O.N.C. in engineering should be obtained by multiplying the number of engineering students by the number of successes amongst all students i.e. $1100 \times 593/1202 = 542$ to the nearest integer. This number is shown in brackets in the table; and it is clear that given this number and the totals of rows and columns it is immediately possible to construct the figures which would be expected in the other cells of the table.

Since the figure for one cell (in addition to the totals) suffices to fix the whole distribution, this table is said to have one degree of freedom. However, if there are several rows and several columns in the table, it is necessary to specify the content of all but one cell in each row or column, and a table having r rows and s columns has $(r - 1)(s - 1)$ degrees of freedom (e.g. 6 degrees of freedom for a 4×3 table). The statistician can then write into each cell the difference δ between the value found experimentally and the value which would follow from multiplying independent probabilities: in the top left-hand cell of the present example $\delta = 539 - 542 = -3$, and since this table has only one degree of freedom all the other three cells must have $\delta = \pm 3$. Without entering into a proof it is plausible to say that the \pm ambiguity is eliminated by squaring δ , and that in general the square of δ is likely to be proportional to the independence value m ; so we derive a measure

$$\chi^2 = \sum(\delta^2/m)$$

where the summation is taken over all cells in the table. There are tables of χ^2 showing for various degrees of freedom the probabilities of various values of χ^2 arising by chance. If the value of χ^2 is greater than could occur by chance with an acceptable degree of probability, it is suggested that the two ways of splitting the population—into rows and into columns—are not independent. In the example shown above the value of δ is the same in all cells, but of course there are different values of m , so that $\chi^2 = 9 \left(\frac{1}{542} + \frac{1}{558} + \frac{2}{51} \right) = 0.386$

Now looking in the table of χ^2 under one degree of freedom we find $\chi^2 = 0.148$ for a probability of 0.7 and 0.455 for a probability of 0.5; so there is a probability of around 0.6 that such a result could arise by chance which is an unhappy value situation. There is no ground for affirming that it is easier to obtain the O.N.C. in building than in engineering, but the idea cannot be firmly excluded.

At this point it is well to be reminded that there is no magic in statistics. Tests such as "student's t " and χ^2 are concerned with evaluating the probable effect of random fluctuations due to sampling. The statistician's ideal model would be to suppose that we had a bag containing a very large number of ball bearings, mostly steel (engineering students) but a known proportion of bronze (building students) and a fixed proportion of the whole marked with a white spot (O.N.C.); if a handful is drawn at random, what proportions of the four different kinds may be expected? Our O.N.C. data do represent a sample because (a) they refer to certain colleges instead of to all colleges, (b) they refer to a single year and (c)

*This is the form given by Yule and Kendall in "An Introduction to the Theory of Statistics," *Burington and May*, in "Handbook of Probability and Statistics, give the algebraically equivalent and apparently simpler formula $t = \frac{(\bar{x} - \bar{X})}{\sigma_s} \sqrt{n-1}$ where σ_s is the ordinary variance of the sample. It is less easy to explain the factor $n-1$, rather than n , in this form. †Based on figures given on p. 356 of the *Growth Report*, "15 to 18," H.M.S.O., 1959.

when we ask whether one course is more difficult than the other, the people who actually enrolled in the courses are only a sample out of all the people who were qualified to enroll. Of course, the sample in social problems may be biased; but unless you are prepared to regard the data as a sample, either approximately unbiased or with a specified bias, you must not apply statistical tests which are based on sampling theory.

The possibility of averaging over several years brings in again the question of the distribution being stationary. In using autocorrelation in order to obtain the power spectrum, the function was integrated over a time which tended to infinity and it was therefore necessary that the distribution of amplitudes be the same at all times, i.e. that the continuing sequence of values form a stationary time series. When the durations of individual events are very short compared with the time of observation of the whole phenomenon, as in most forms of electrical noise, the distribution function is usually exactly stationary; but this may not be so in other problems. For example, when studying congestion in telephone systems one has individual calls lasting a few minutes but the total loading varying appreciably over a period of an hour at some times of day (the G.P.O. reckon the peak period to be the two hours from 10 a.m. to noon): it follows that one cannot take any averages over a period very long compared with the duration of one call.

Decision functions

The remaining major topic is *decision functions*. The binary symmetric channel (b.s.c.) is so often assumed that one tends to assume that the chance of 0 being converted to 1 by noise should necessarily be the same as the chance of 1 being converted to 0, but this is not generally correct. One can usually distinguish between a *false-alarm risk* α and a *lost-signal risk* β , a terminology which is applicable to many situations including radar. In the b.s.c. one makes the false-alarm risk (0→1) equal to the lost-signal risk (1→0) by appropriate choice of the decision-point or threshold. In a radar system, however, one might first specify a maximum acceptable risk of lost-signal (for a defined signal fixed e.g. by radar transmitter and receiver characteristics, distance to the horizon and assumed type of target) and then adjust the system to minimize the risk of false alarm subject to the specified maximum risk of lost signal. This is known as the Neyman-Pearson criterion for setting up the decision mechanism, and its difficulty is that it implicitly assumes at least two degrees of freedom in the system, one to be adjusted for each half of the criterion. If the only available adjustment is a threshold, then setting it for the specified β necessarily fixes also the value of α . Hence the Neyman-Pearson criterion is significant only if there are also other adjustments which can be made, e.g. bandwidth, and these adjustments do not affect signal and noise equally.

One will always be balancing the change in α against the change in β when adjusting the threshold; consequently one's decision should be made in terms of the likelihood ratio L of the received signal,

$$L(y) = \frac{p(y|H_1)}{p(y|H_0)}$$

where $p(y|H_1)$ and $p(y|H_0)$ are the probability density functions of the probabilities that the observed signal y would have resulted from transmitting a mark or space respectively. L is then to be compared with some decision threshold K in likelihood and y attributed to a mark or space according as L is greater or less than K .

The *a priori* probabilities of signal and no-signal are not necessarily equal. An obvious example of this is a ballistic-missile early warning radar system (BMEWS) where it is hoped that signals will always be absent. A more subtle example is teleprinter communication in English (or any West European) language. The letter E, which is far more common than any other letter, has only one mark element and four space elements, and other common letters, T, A, O, I, N have only two mark elements out of five, so there must be fewer mark than space elements in an English-language transmission and for minimum overall errors one should make α rather less than β , i.e. set the threshold a little higher than half way between space and mark. So let π_0 and $\pi_1 = 1 - \pi_0$ be the *a priori* probabilities of space and mark, and let C_0 and C_1 be the costs of false-alarm and lost-signal errors. Multiplying π_0 (the frequency of occurrence of spaces) by α (the proportion of mistakes made on spaces) and C_0 (the cost of each such mistake) one finds $\pi_0 \alpha C_0$ as the cost arising from false-alarm errors, and $\pi_1 \beta C_1$ for lost-signal errors. The total cost $R = \pi_0 \alpha C_0 + \pi_1 \beta C_1$ is minimized if for every signal y one computes the likelihood ratio $L(y)$ and sets the dividing line in L at the value

$$K = \frac{\pi_0 C_0}{\pi_1 C_1}$$

In general terms this says that one should set the dividing line in likelihood higher in proportion jointly as "space" signals are transmitted more often, and the cost of false-alarm errors is greater, or inversely for frequency of mark signals and cost of lost signals. This test, which minimizes cost when the *a priori* probabilities are known and average cost is a linear function of the absolute error probabilities, is known as a Bayes test.

Using the Bayes test

As an example, suppose that $\pi_0 = 0.6$ and $\pi_1 = 0.4$ (compare the teleprinter mark/space ratio) and that the error costs are the same, $C_0 = C_1$. Hence in this example $K = \pi_0/\pi_1 = 1.5$. The likelihood ratio is the ratio of the slopes of the error probability curves at the value of y in question. If a signal of amplitude y' is received over a channel with Gaussian noise power N , the probability of its being due solely to noise (the transmitted condition being "space") is

$$\alpha = \frac{1}{\sqrt{2\pi N}} \int_{y'}^{\infty} \exp\left(-\frac{y^2}{2N}\right) dy$$

The probability of its being due to a "mark" transmission with received amplitude m is

$$\beta = \frac{1}{\sqrt{2\pi N}} \int_{-\infty}^{y'} \exp\left(-\frac{(y-m)^2}{2N}\right) dy$$

The slopes $d\alpha/dy$ and $d\beta/dy$ are the integrands, so the likelihood ratio is

$$\begin{aligned} L(y) &= \frac{\exp\left(-\frac{(y'-m)^2}{2N}\right)}{\exp\left(-\frac{y'^2}{2N}\right)} \\ &= \exp\left[-\frac{m}{n} \left(\frac{m-y'}{2}\right)\right] \\ &= \exp\left[-k \frac{P}{N} + \frac{m y'}{N}\right] \end{aligned}$$

where P is the signal power and k is a function of the signal waveform. The received signal is recorded as a mark if its amplitude y' is such that $x(y')$ is greater than 1.5.

One further criterion should be mentioned, though its application is beyond the scope of this article, and that is the one known as a *minimax* test. The simple Bayes test requires a knowledge of the signal parameters π_0 and π_1 , but in fact these may be uncertain. Therefore one picks out the worst-case values of π and constructs a test which will give the minimum cost in errors if this worst case occurs. This is the *minimax* criterion, but for all other values of π it will be less efficient than a Bayes test constructed for the specific value of π .

It is difficult at this stage to construct realistic examples. As in all branches of mathematics, practice in use is essential to an understanding of theorems, and the reader who is seriously concerned with decision functions is advised to work out his own cases with the aid of the specialist literature.²

REFERENCES

- (1) "Communication in the Presence of Noise," by C. E. Shannon. *Proc. I.R.E.*, vol. 37, p. 10, 1949.
- (2) "Detection Theory," by I. Selin. Princeton University Press, 1965.

BOOKS RECEIVED

Electronic Components, Tubes and Transistors, by G. W. A. Dummer. Treating the component as a building block, the author explains the characteristics and measurement theory of components treated as discrete elements and then discusses the practical application of components in equipment. Separate chapters deal comprehensively with resistors, capacitors, magnetic materials, electromagnetic components, valves and semiconductor devices. Chapters contain references to sources for further reading; also questions and answers. Pp. 166; Figs. 76. Price 21s. Pergamon Press Ltd., 4 & 5 Fitzroy Square, London, W.1.

Guide to Radio Technique Vol. 1; Fundamentals, Valves, Semi-conductors, by E. Julander. This is a book intended to provide a basis for the enthusiast or student who seriously wishes to improve his theoretical knowledge of radio rather than make a continued practical approach. The opening chapters deal with the fundamentals of electricity, complex numbers and transmission units. A complete chapter deals with the construction, theory and operation of thermionic valves; this is then followed by a similar chapter on semi-conductors in which a comparison is made with valve circuits. The final chapter deals with generation, propagation and modulation of electro-magnetic waves. A 14-page appendix contains useful data presented in the form of tables and monograms. The book is issued as part of the Philips Technical Library and is a translation of the original Swedish edition. Pp. 238; Figs. 214. Price 37s 6d. Distributed by Macmillan & Co. Ltd., Little Essex St., London, W.C.2.

British Miniature Electronic Components Data 1965-66, edited by G. W. A. Dummer and J. Mackenzie Robertson. In this, the fourth annual issue of the publication nearly 170 different types of components from nearly 140 firms are listed. For most components, the textual information is supported by photographs, line drawings and tabular matter. Comprehensive details are given of each type of component and these range from accelerometers to wires and cables. Manufacturers have made available advance details of components to provide the most up-to-date information for designers, users and buyers. Pp. 984. Price £9. Pergamon Press Ltd., 4 and 5 Fitzroy Square, London, W.1.

A new British Standard, BS3934, "Dimensions of Semiconductor Devices," has recently been published. Although international discussion is currently in progress, the need within the electronics industry for a record of the dimensions of semiconductor devices in general use necessitated publication of the Standard as soon as possible. It has, however, been published in loose leaf form to facilitate amendments. The primary object of the Standard is to provide a basis of comparison for mechanical interchangeability; electrical and thermal data is not covered. Copies of the Standard may be obtained, price £4 10s from B.S.I. Sales Branch, 2 Park St., London, W.1.

Electrical & Electronic Trader Year Book 1966.—As usual this book, compiled in collaboration with *Electrical & Electronic Trader*, is a mine of information covering most of the general important aspects of the trade. The publication is divided into eight sections—Valve base diagrams—Specifications—Technical literature—Wholesalers—Proprietary names—Buyers' guide, radio and television—Buyers' guide, domestic electrical—Trade addresses—identified by thumb index tabs which facilitate rapid reference. Additional general information contains a legal guide which includes details of aerial erection, hire purchase and credit sales, noise abatement, tenancies of business premises and the Wireless Telegraphy Act. Pp. 480. Price 35s (plus postage). Iliffe Books Ltd., Dorset House, Stamford St., London, S.E.1.

Servicing Electronic Organs, by C. R. Pittman & E. J. Oliver. The book is written as a practical guide to the theory and operation of electronic organs for the serviceman requiring basic knowledge of this application of electronic techniques. Details are given of the basic circuits encountered and include illustrations and information reproduced by courtesy of ten American organ manufacturers. Originally published in America in 1962. Pp. 191; Figs. 112. Price 30s. W. Foulsham & Co. Ltd., Yeovil Road, Slough, Bucks.

Transistor Receivers and Amplifiers, by F. G. Rayer. Treatment used in the book is essentially of a practical nature and is intended to provide guidance for the home constructor interested in the use of transistors. Basic contents include general descriptions of operating characteristics of semi-conductors, aerials and r.f. amplifiers, superhet circuitry and power supplies. Additional chapters describe printed circuits, test equipment and fault finding. Pp. 164; Figs. 129. Price 30s. Focal Press Ltd., 31 Fitzroy Square, London, W.1.

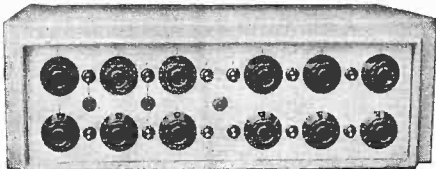
Many of the titles in the M.I.T. Radiation Laboratory Series of text books originally published by the McGraw-Hill Book Company and regarded as standard works of reference are now available in paper back form from Constable & Co. Ltd., 10-12 Orange St., London, W.C.2. They are:— **Computing Mechanisms and Linkages**, by Svoboda, 18s; **Microwave Antenna Theory and Design**, by Silver, 24s; **Microwave Transmission Circuits**, by Ragan, 24s; **Pulse Generators**, by Glasoe & Lebacoz, 24s; **Principles of Microwave Circuits**, by Montgomery, Dickie & Purcell, 18s; **Propagation of Short Radio Waves**, by Kerr, 24s; **Radar System Engineering**, by Ridenour, 24s; **Theory of Servomechanics**, by James, Nichols & Phillips, 18s; **Threshold Signals**, by Lawson & Uhlenbeck, 18s; **Vacuum Tube Amplifiers**, by Valley, Wallman & Henry, 24s; **Waveforms**, by Chance, Britton, Hughes, Vernon, MacNichol, Sayre, David & Williams, 26s; **Waveguide Handbook**, by Marcuvitz, 18s.

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The 8 valves and selenium rectifier draw a total of 25 watts.

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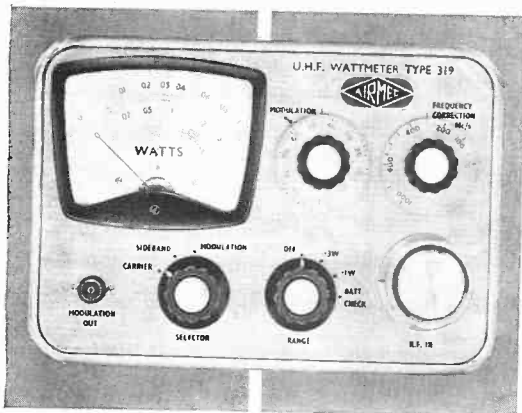
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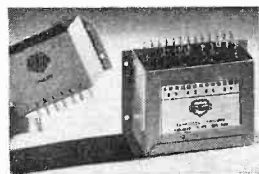
PANORAMIC MICROWAVE RECEIVER

THIS receiver, from Microwave Physics (U.S.A.), is for use in the design and testing of harmonic generators, solid state signal sources, transmitters, and other signal generators, where observation of spurious and harmonic signals is desirable. The circuit uses an electronically swept yttrium iron-garnet bandpass filter as a prescaler. The filter which has a decade tuning range, is followed by a sensitive crystal detector, and a low-noise video amplifier.

Two versions of this receiver are available; the MPR-U (0.5 to 5 Gc/s), and the MPR-X (2 to 12 Gc/s), both sets permitting continuous 'scope display of signals within the stated ranges.

The MPR-U and MPR-X are available either as plug-in modules for Tektronix scopes (£722) or with self-contained power supplies for use with other scopes (£876).

Marketed by Microwave Systems Ltd., 9-10 River Front, Enfield, Middx. WW 304 for further details



and motor field heating can be added to these basic circuits.

The MY5011 thyristor firing module will fire four 70 A thyristors connected in parallel or series, and provide continuously variable, and reliable control, over a power range of below 0.25% to above 99.9% of maximum power.

Size: 9.14 x 6.60 x 4.82 cm in an epoxy resin encapsulation.

Price: £14 18s.

Also available in the simple, single-phase module MY5000, capable of driving two 70A thyristors. Price: £5 5s.

Current limiting and surge suppression module (MY5051) is for use with existing thyristor stacks and firing modules. It will enable single-phase or three-

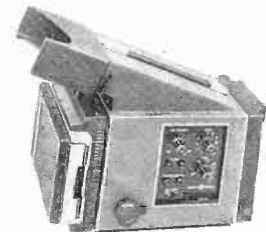
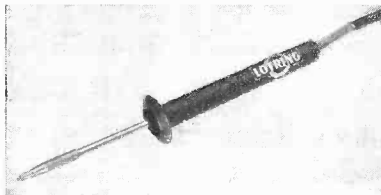
NICKEL BIT SOLDERING IRONS

SOLDERING irons with bits of solid nickel are now being produced by the West Germany company, Loring Werner Bittman, of Windscheid Strasse 18, Berlin 12. Nickel was chosen in

preference to copper because of its excellent resistance to corrosion by the chemicals and resins used as fluxes. It is readily "wetted" by the solder, while its thermal capacity and conductivity are satisfactory for this application.

Two models Pico Pen 15TS—(15W for small work) and Pico Post 30PL—(30W for larger work) have their heating elements and bits designed so that heating to 200 degC takes 0.7 and 1.5 minutes respectively.

WW 305 for further details



OSCILLOSCOPE CAMERA

THE Hewlett-Packard model 197A camera uses an electronically controlled shutter calibrated in nine even steps from 1/30 to 4 s. X-synchronization, particularly useful in photographing one-shot phenomena, is controlled by connections made to an external instrument panel.

Where an internal graticule is used on the oscilloscope an optional ultraviolet light source is available to illuminate the graticule. Pre-exposure for the graticule and exposure for the trace is performed automatically by the setting of a single control. The back of the camera can be rotated from a horizontal position to vertical so that two smaller photographs can be taken on a single film. Price of the camera is £184 or without the ultraviolet light facility £165.

WW 308 for further details

Capacitance Tester

FOUR-DIGIT in-line readout is utilized on the Model 1201 Digital Capacitance Tester manufactured by the American company of Micro Instrument Co. and marketed by Claude Lyons Ltd., Ware Road, Hoddesdon, Herts. The instrument has been designed for two- or three-terminal measurements and uses a 1 Mc/s, 35 mV r.m.s. test signal. A 0 to 100 V bias supply is provided for reverse biased diode and varactor capacitance measurements and for transistor C_{ob} (capacitance between base and collector with emitter o/c) and C_{cb} (capacitance between base and emitter with collector o/c) measurements. Overall accuracy is a combination of measurement accuracy, which is $\pm 3\%$, and a readout accuracy of ± 1 digit. The instrument will tolerate shunt resistances down to 10 k Ω without loss of accuracy and permits in-circuit capacitance measurements and measurements

on integrated circuits having low shunt resistance. Operation is based on the principle of frequency deviation. Two high-stability tunnel diode oscillators are used, one of which varies its frequency linearly with the applied capacitance. The difference frequency output from a mixer is applied to gating and zeroing circuitry and then to an electronic frequency counter for direct display as capacitance.

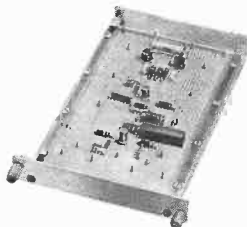
Three versions are available. Two single-range versions are models 1201 (0-99.99 pF) and 1201S (0-999.9 pF) both priced at £798 or £878 with print out, and a dual-range model 1201DS (0-99.99 and 0-999.9 pF) at £998 or £1,078 with print out. A transistor jig for TO-5 and TO-18 cans with C_{ob} and C_{cb} switch and 30 in coaxial leads is available for £40.

WW 306 for further details

Circuit Board for Module System

AN insulated board, the C.B.1, is perforated to the 0.1 in standard grid, and is suitable for mounting wire or tag-ended components. The board mounts on two tie rod supports behind individual module front panels. Each module can carry 1 or 2 boards. It is claimed that for experimental construction circuits, this system is neat, durable and versatile. Price: 4s 6d. A.P.T. Electronic Industries Ltd., Chertsey Road, Byfleet, Surrey.

WW 307 for further details



Matched Signal Divider

THE Bishop Type 080-001 matched signal divider permits a source to drive two paths, and maintains 50 ohms impedance at input and output. The ability to handle signals with fast rise time of 100 picoseconds makes it suitable for pulse as well as u.h.f. carrier wave applications. It can also be used to add signals from 50 ohm sources. Connectors are BNC (3).

Available in the U.K. from Claude Lyons Ltd., of Hoddesdon, Herts., the divider costs £10.

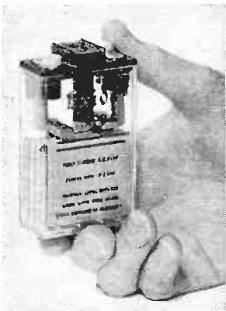
WW 309 for further details

Unspillable Lead Acid Cell

THE compact 2V Exide cell MRP7 which weighs 12oz filled and measures $1\frac{1}{2}$ in \times $2\frac{1}{4}$ in \times $3\frac{1}{4}$ in is unspillable, even when inverted. The plate alloy permits the cell to be maintained under floating and trickle charge conditions, and to be used where frequent discharge and recharge demands have to be met. The cell has a capacity of 4Ah at a 20 hour rate, and is suitable for intermittent, and standby use, since it is capable of standing idle for long periods.

Moulded in transparent styrene acrylonitrile, with four ribs to support the plates, and accommodate sediment, the container has red lines on it, to indicate maximum and minimum electrolyte levels. Specific gravity charging rate, and topping up instructions are also shown. Price 29s.

WW 310 for further details



Miniature Lever Switches

THE 5000 and 5100 BECUWE series of lever key switches made in France have stainless-steel frames, levers, spindles and securing plates. The keys are from 0.78in (20mm) to 1.73in (44mm) in length, depending on contact combinations, 0.425in (10.8mm) wide, and 1.76in (44.6mm) deep (behind panel). They are identical in all characteristics, except for the method of panel mounting; the 5000 series being supplied with a central stainless-steel milled nut for 0.325in (8.2mm) diameter single hole fixing, while the 5100 series have 2 fixing screws for 2 side holes 0.09in (2.3mm) diameter.

The cam is made of the hard plastic DELRIN, and the contact springs of nickel-silver. The self cleaning twin contacts in fine silver have the following ratings (non-inductive) for a.c. operation at 48V-2A, at 110V-1A, and



at 250V-0.6A. For d.c. operation 48V-0.6A, at 110V-0.3A, and at 250V-0.2A.

Contacts are also available in Palladium or other materials to order. The maximum number of contacts per side for all types are 3 change-over, or 5 normally-opened, or 5 normally-closed.

The conical stainless-steel lever can be colour coded by means of a slip-on neoprene mantle in Black, Green, Red, Blue or White. Full information from Britec Ltd., 17 Charing Cross Road, W.C.2.

WW 312 for further details

Time Interval Measurement

THE A. G. Brown MINITIME is designed to measure the time interval between opening, and/or closing of contacts. Moulded in black plastic, and powered by two batteries, this instrument has a 3in meter with a linear scale, calibrated 0 to 10 and 0 to 3. Persistence of reading is within 1% for at least 1 minute, provided the insulation resistance of the external circuit is high enough to prevent leakage leads or contacts. The ranges cover 0 to 300 μ s, 0 to 300ms, and 1 to 10s. Readings are $\pm 2\%$ f.s.d. Controls on the front panel are: range selector, function selector, zero set, calibrate.

Amongst suggested applications for the MINITIME are, measurement of delay between closing or opening of sets of contacts on relays or contactors, measurement of velocity of moving objects by operation of trip wires, etc.,



and measurement of exposure times in photography. Size: $5\frac{1}{2}$ in (14.5cm) \times $3\frac{1}{2}$ in (9.5cm) \times $1\frac{1}{4}$ in (4.5cm). Weight 1lb 4oz (0.6kgm) unpacked. Batteries B121, U10 (1 of each).

From A. G. Brown Electronics Ltd., Lower Mills, Busby, Glasgow.

WW 313 for further details

MICRO-CIRCUIT WELDER

A COMPACT multi-purpose micro-circuit bonding system, capable of welding, brazing, parallel-gap soldering and thin film diffusion bonding, is available from Hughes International (U.K.) Limited, Heathrow House, Bath Road, Cranford, Hounslow. Occupying only 20 inches of bench space, the system includes a power supply (MCW 550), two sizes of bonding heads, micro-positioning apparatus and accessories, and a stereozoom microscope. The smaller head, designated Model VTA-90, is designed for ultra-fine materials and can be adjusted to apply bonding pressures as light as 10 grams. The larger head, VTA66, is used for parallel-gap welding or soldering of flat packs, memory plane matrices and other surface bonding applications.

WW 311 for further details

INFORMATION SERVICE FOR PROFESSIONAL READERS

To expedite requests for further information on products appearing in the editorial and advertisement pages of *Wireless World* each month, a sheet of reader service cards is included in this issue. The cards will be found between advertisement pages 16 and 19.

We invite professional readers to make use of these cards for all inquiries dealing with specific products. Many editorial items and all advertisements are coded with a number, prefixed by WW, and it is then necessary only to enter the number on the card.

Postage is free in the U.K. but cards must be stamped if posted overseas. This service will enable professional readers to obtain the additional information they require quickly and easily.

COMMUNAL AERIAL AMPLIFIERS

NEW communal aerial amplifiers for television and sound signals and combined u.h.f. and v.h.f. crystal controlled converters have been introduced by Teleng Ltd. The design of the new

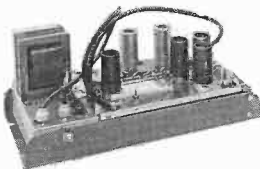


amplifiers is based on the dual channel U30 amplifier to which has been added a wide band amplifier and power supply. Two models are being introduced—"Popular Four" (Type E342) priced at £16 10s and "Popular Five" (Type E343) priced at £18 6s; the gain of the E342 is 25 dB for both TV and f.m., whereas the gain of E343 is 40 dB for TV and 35 dB for f.m. The amplifier can be preset for reception of any specified channel in Band I and Band III but for Band II the tuning range covers 88-100 Mc/s. Sensitivity is 250 μ V minimum and output is 100 mV maximum.

These amplifiers are most suitable for smaller networks where the original wide-band "U" series of amplifiers has

proved to be too expensive. Also in installations where provision for the later addition of other channels is not required, the units are claimed to be most competitive.

The u.h.f. and v.h.f. converters are based on a modified design of the U166 converter, which allows BBC-2 to be "added" to existing single-channel systems. This has resulted in two versions of the U166 which are designated Type U338 (£49 15s), having a gain of 50 dB, and Type U340 (£47 4s) having a gain of 35 dB. Maximum output levels for both amplifiers is 300 mV. The combination of the new converters and amplifiers produce a distribution system for three television channels and three

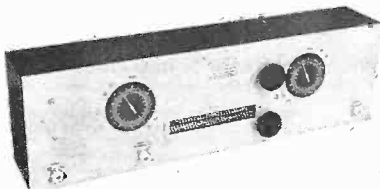


f.m. programmes which will provide high quality signals for up to 100 outlets. Available from Teleng Ltd., Teleng Works, Church Road, Harold Wood, Essex.

WW 313 for further details

R.F. Piston Attenuator

A PISTON attenuator giving an attenuation range of 100 dB at frequencies from 1 Mc/s to 2 Gc/s has been introduced by Flann Microwave Instruments (9, Old Bridge Street, Kingston-upon-Thames, Surrey) under the type number IF 100. Designed as a basic attenuation standard, it comprises a precision electroformed tube with an input-tunable launching loop and a sliding pickup loop coupled to a direct-reading



scale. The change in attenuation is indicated on dial gauges in steps of 2 dB per revolution. The upper linear scale is permanently fixed and indicates 0-100dB in 2dB steps. The lower linear scale is movable and provides an arbitrary datum when set to a cursor line anywhere within the range of the attenuator. Impedance is 50 Ω and the cut-off frequency is in the region 3-4 Gc/s. For the highest accuracy measurements - small setting corrections have to be made as the measurement frequency deviates from the design frequency (60 Mc/s), but a correction curve is supplied.

WW 315 for further details

260

Rectifier Bridges for Printed Circuits

THESE bridges by Pirelli (Milan) are single phase, resin cast units, with a body length of $\frac{1}{8}$ in. and $\frac{1}{16}$ in. diameter.

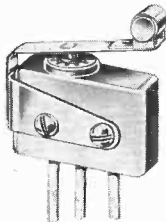
The WO2, WO4 and WO6 bridges which respectively have a p.i.v. of 200, 400 and 600 V, possess these principal ratings.—Max. average rectified output at 50°C, 1.5 A, at 100°C, 1 A; peak single-cycle surge current 50 A; max. forward voltage drop at 1 A and 25°C, 2 V; and max. leakage at rated p.i.v. and 25°C 10 μ A. The bridges have an operating temperature range of -55° to +125°C. Prices are: WO2—19s 6d; WO4—22s; and WO6—28s 6d.

Available from S.D.S. (Portsmouth) Ltd., 67-69, Commercial Road, Portsmouth, Hants.

WW 316 for further details

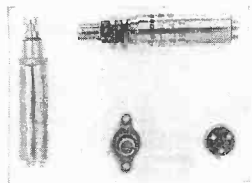
SUB-MINIATURE SWITCH

ALL the existing SE series of switches (with the exception of the 75E model) of Honeywell Controls Ltd., Brentford, Middlesex, is to be replaced by a recently developed miniature switch—the 91SE. The switch, designed primarily for use in the aircraft industry, incorporates a new fluoro-silicone seal, with a corrosion-resistant steel plunger which ensures protection from all oils and fuels. Basically, the switch comprises a standard 11SM1-T micro switch sealed in epoxy resin in a corrosion-resistant case 0.6 x 0.876 x 0.34 in. The cable connections are covered by an outer sheath of extruded p.t.f.e. which affords good protection against heat and abrasion. Designed to operate within a temperature range of -62 to 150°C, the 91SE can be fitted with a range of actuators from the Honeywell J.E. series. By this



means the switch can be operated from cams, slides or other devices which are not in line with the motion of the switch plunger. The illustration shows a roller leaf actuator, JE-5, for applications involving rapid cam or slide operation.

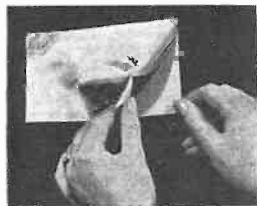
WW 317 for further details



HELICAL POTENTIOMETER

A TEN turn potentiometer with a resistance range of 50 ohms to 100 k Ω is offered by Reliance Controls Ltd. A linearity of $\pm 1\%$ is available, which can be improved by selection to $\pm 0.1\%$. With a temperature range of -55°C to $+100^{\circ}\text{C}$, the unit has been tested for vibration and shock to MIL.202. The style number is HEL. 05-10, and the unit has a body diameter of $\frac{1}{2}$ in. Available in flange and bush mounted versions, as seen in the illustration, from Reliance Controls Ltd., Sutherland Road, London, E.17.

WW 318 for further details



Copper Sheet for Printed Circuits

MANUFACTURED by Peak Sound (Harrow) Ltd., and marketed under the name "Cir-Kit" this adhesive copper sheet is intended for constructing prototype circuits.

The sheet (0.002 in thick) is coated on one side with a heat resistant adhesive, which, in turn, is protected by a paper backing. Offered in $\frac{1}{2}$ in wide strip (at 7s 6d for 24 feet) that can be cut to length, or in 6 in wide sheets (6s for 1 foot) permitting required dimensions to be drawn on the backing paper, and cut out with scissors or knife. The paper is stripped off, and the shape pressed firmly on a laminated board. Normal drilling and soldering may be carried out.

From Peak Sound (Harrow) Ltd., 10 Asher Drive, off Mill Ride, Ascot, Berks.

WW 319 for further details

195 CHANNEL MARINE TRANSMITTER

THE Redifon G.341 gives complete coverage of the marine bands 400 to 535 kc/s, 1.6 to 3.8 Mc/s, and 4 to 26 Mc/s, offering a choice of 195 channels between 400 kc/s and 26 Mc/s. As a ship's main transmitter, it also incorporates facilities for short, medium and long range radiotelephony on s.s.b. as well as on d.s.b. The output power of 1,200 W p.e.p. available for long distance s.s.b. h.f. radiotelephony can be reduced by preset adjustments so as to comply with particular national regulations.

Deck or bench mounted, the transmitter and power unit are in separate housings. Modular construction is used extensively, facilitating speedy interchange of units and sub-units, without realignment or soldering. Ease of access is ensured by glide-out units. Short circuiting or breaking the aerial connection will not damage the transmitter, and power supply circuits are fully protected against damage under fault conditions.

The G.341 is fully type approved by the G.P.O., complying with the 1965 Merchant Shipping (Radio) Rules, also conforming with C.C.I.R. Recommendation No. 258 (Los Angeles 1939) with respect to SSB equipment for maritime use.



Power supply: 440 V ($\pm 10\%$) 50/60 c/s 3 phase.

Power consumption: maximum 3 kVA.

Redifon Ltd., Broomhill Road, S.W.18.

WW 320 for further details

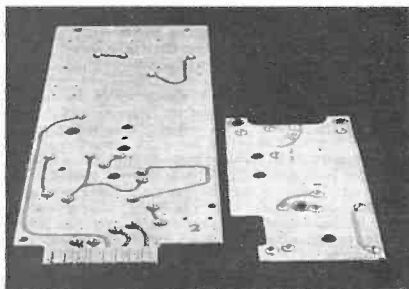
Cored Preform Washers

UNDER varying forms of vibration, printed circuit boards may show failure of contact at the eyelets. The small fluxed solder preforms from Enthoven are designed to go on to the printed circuit, and under the eyelet, which is then punched by the normal method. The preformed washer, now sandwiched between board and eyelet, can be hand or machine soldered, offering a sound mechanical joint, and minimizing this fault condition.

These preforms can be produced with diameters down to 0.047 in, and with thicknesses down to 0.01 in. The flux contained in the preforms is a non-corrosive organic type of material.

Large runs requiring solder preforms of this type, can be supplied to any specification with reference to the alloy (available with melting points from 190 deg C to 245 deg C), and dimensions. Enthoven Solders Ltd., Upper Ordnance Wharf, Rotherhithe, S.E.16.

WW 321 for further details



Mirror-Image

It was Alice who started it all when, in Looking Glass Land, she found that she had to run very fast indeed in order to stay in the same place. Today it has become an article of faith; a good old standby for the politician guest of honour at the electronics dinner, who never fails to remind his audience that electronics is a growth industry and therefore its economy must expand steadily if it is to survive.

The audience, mostly company chairmen, in turn never fail to applaud this sentiment because it is always safer to clap than to be obliged to argue later about something you don't understand. The politician, thus encouraged, follows up with an exhortation to one and all to double their export sales. This is received in deadly silence because experts are a chairman's recurring nightmare and he doesn't want to be reminded about them at social functions.

Eventually, the constant reiteration at such gatherings gets on the chairman's nerves to the point where the subject is put on the agenda for discussion at the next board meeting. The directors don't understand the economy-expansion bit either, but it would never do to admit it, so it is decided that vigorous action is called for. As this involves nothing more onerous than telling the general manager to pull his socks up, it is implemented forthwith.

The hapless G.M., having no wish to receive the golden handshake just yet, gets down to an analysis of the situation. He knows only too well that his export figures can only be doubled at the expense of the mighty American electronics industry, and the temptation to cast the situation in terms of David and Goliath is irresistible. But the stripling David, he reflects glumly, did not have to take the field with his hands tied behind his back by lashings of credit restrictions.

There is clearly only one possible course of action. He must build his stripling into something approaching the size of the opposition, starting with that genesis of exportable products, the research laboratories. Mass attack is henceforth the order of the day.

So, away to the universities and technical colleges and rope in anyone who has a degree and feels warm to the touch. And, in parallel with this operation, send all senior research men over to the States to get the know-how on this mass attack business.

This realistic approach brings results. The recruitment scheme donates so much new blood that there are not enough arteries at the laboratories for them to course through, so 50 Nissan huts have to be thrown up on the nearest available site, a piece of cow pasture some forty miles away. There the recruits sit like birds in the wilderness with not a multi-test set between them.

The delegation to the U.S.A. also brings in the sheaves. Seventy-two per cent of its members return to hand in their respective resignations. This is highly encouraging to the new blood because it makes for quick promotions. The overall situation is encouraging to the Board because without doubt the G.M. has succeeded in invoking the magic word "expansion," although the directors are a shade hazy as to the detail. It is also encouraging to Goliath as he goes about his lawful occasions of kicking the undercarriage from British avionics and covering every square inch of the country with microcircuits.

With the bit now firmly between his teeth the G.M. then holds an inquest on development and production, with his

senior management as chief witnesses. He enters the meeting thoroughly alarmed at the monumental ratio of clerical staff to those on the production floor and determined to enforce a ruthless streamlining. But the senior management have done their homework thoroughly and each produces statistics to prove beyond doubt that, far from being over-stuffed, his particular domain is well below skeleton level and only functions as well as it does by reason of genius at the helm.

Baffled in his attempt to beat Parkinson's Law, the G.M. in desperation invokes the most powerful ju-ju of all. He calls in a firm of Efficiency Consultants.

At zero hour the Assyrian comes down like a wolf on the fold. A small army of elegant young ladies and gentlemen disperse themselves around the territory like langorous Big Sisters and Brothers—watching. This is phase one. In phase two the meetings start. Interminable meetings, presided over by one or other of the voluble experts: meetings at which expressions like "procedures improvement," "procedures analysis," "timetable productivity," "function output" and "conversion integrals" sweep like machine gun fire over the stunned assembly. The entire organization dissolves into soviets as if all the secret societies in the world were holding their meetings in separate compartments under one roof. The telephones are all silent; no point in ringing anyone because he will be at a meeting.

At the end of a highly expensive twelvemonth the Chief Wizard pitches in his recommendations to the G.M. Shorn of its double talk it advises that several new departments should be inaugurated without delay. One of the most vital of these is the Image Promotion Dept., for it seems that the Company Image needs a face-lift which can only be achieved by a strenuous drive in prestige advertising in newspapers, journals, cinemas, television and exhibitions.

A 500% increase in time and motion study staff is considered essential, together with the institution of a Network Analysis Division. The immediate objective of the latter is to replace the old-fashioned term "work" with all its distasteful associations by the more socially acceptable word "activities," thus at last giving recognition to those who are perpetually in a high state of activity but do not work at all.

In depth, it exists to promote secret clubs, identified by such acronyms as BURP, SLUG, RANT or SQUIRT (the latter standing for System Quotient Unification in Rational Tendencies).^{*} The initiates go into hibernation with a computer for three months and then emerge with a wall-chart emblazoned with something which looks like a cross between a drunken octopus and a map of London's Underground.

In brief, a dazed G.M. adds all the efficiency-aid departments suggested. When this is done he finds himself at the head of an empire which is twice the size it was before the Efficiency Consultants got at it. The G.M. is accordingly restyled arch-comptroller; his erstwhile managers become comptrollers; section chiefs become managers and so on all down the line, with salaries increasing *pro rata* naturally.

The entire Company structure hums with new vigour. Everyone in the organization is hard at it from dawn to dusk answering internal memos and initiating their own quota. The whole outfit is now in a condition which has been aptly

^{*} The latest from the U.S.A. is PRIDE—Personal Responsibility In Daily Effort.—Ed.

described as self-oscillating, using paper-work coupling.

The Chairman can now listen with equanimity to any after-dinner politician, for his Company is now more than fulfilling the dictates of the economists and expanding like a soap bubble (if the simile is not too unfortunate).

Exports? . . . er . . . well. The full benefits of expansion will naturally not be felt for a year or so and—purely as an interim measure, you understand—the prices of end products will have to increase somewhat; a moderate 50% perhaps? After all, these foreigners should be only too glad to pay for solid British workmanship allied to 21st century efficiency.

MAY MEETINGS

Tickets are required for some meetings; readers should, therefore, communicate with the secretary of the society concerned

LONDON

2nd. I.E.E.—Colloquium on "Current topics in tropospheric propagation" at 2.30 at Savoy Pl., W.C.2.

9th. I.E.E.—Discussion on "An engineer's ideas on some of the electromagnetic fundamentals" at 5.30 at Savoy Pl., W.C.2.

10th. I.E.E.—"Automatic control in railway systems" by S. Jones at 5.30 at Savoy Pl., W.C.2.

11th. I.E.E.—"Holography" by Prof. D. Gabor at 6.0 at Savoy Pl., W.C.2.

13th. I.E.E.—Discussion on "The C.E.I., its implications on engineering courses" opened by Prof. M. G. Say at 5.30 at Savoy Pl., W.C.2.

23rd. I.E.E. & I.E.R.E.—Discussion on "Electrical aids to the handicapped" at 10 a.m. at Savoy Pl., W.C.2.

23rd. I.E.E.—Colloquium on "A new look at energy storage" at Savoy Pl., W.C.2.

BIRMINGHAM

18th. Television Soc.—"Take 625 lines" by L. Marsland Gander at 7.0 at Broadcasting House, Carpenter Rd., Edgbaston.

FARNBOROUGH

10th. I.E.E.—"Long-wavelength laser generation" by L. E. Mathias at 6.30 at the Technical College, Boundary Rd.

MANCHESTER

10th. I.E.E.—Hunter Memorial Lecture "Lasers and associated devices" by Dr. G. G. MacFarlane at 6.15 at Renold Building, the College of Science and Technology.

PAISLEY

10th. I.E.E.—Discussion on "Computers and computing in electrical engineering courses" at 6.0 at the College of Technology.

PLYMOUTH

9th. Television Soc.—"Video tape for studio inserts" by Reg Pirkin at 7.30 at the studios of Westward Television Ltd.

LATE APRIL MEETING

LONDON

26th. Soc. of Relay Engrs.—"The application of wideband transistor amplifiers to modern v.h.f. television networks" by R. J. Seacombe at 2.30 at the I.E.E., Savoy Pl., W.C.2.

WIRELESS WORLD, MAY 1966



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OVER 15,000 COMPONENTS

MADE THROUGHOUT IN OUR FACTORY AT BARKING
(A SMALL SELECTION)



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The world's largest range of Legend Indicators to choose from.



List No. J.35
2-pole insulated Short Panel Jacks conforming to B.S.666.



List Nos. P.538-9
Chrome or Gold plated Jack Plugs for co-ax. concentric cables.



List No. D.925
Miniature push-fit neon lamp with AMP 250 series tags.



List No. C.R.48
Insulated Crocodile Clip with tags suitable for AMP-type connectors.



List Nos. F.296, 297
All phenolic moulded panel mounting fuseholders. Front Knob carrier unswivel.



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List No. B.1
Panel-mounting Cell Holder accepting one Cell U.11, etc. size.



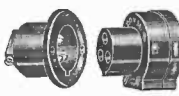
List No. B.6
Base-fixing Cell Holder accepting one Cell U.2, etc. size.



List No. K.492
One of a range of bar knobs with aluminium 'decor' inserts.



List No. K.303
A pair of concentric grub-screw fixing Instrument Control knobs.



List No. P.73/S.E.
Fully insulated side-entry connectors with one-piece floating electrodes.



List No. S.M.370/P.D.
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WW-124 FOR FURTHER DETAILS.

FLIGHT DATA RECORDING

DETAILS FROM THE AEROSPACE INSTRUMENTATION SYMPOSIUM AT CRANFIELD

SPONSORED by the College of Aeronautics and the Instrument Society of America, the fourth international Aerospace Instrumentation Symposium held at the College at Cranfield, Beds, at the end of March, was the best attended yet. Well over 300 delegates were present from 10 countries and 31 papers were read.

Held in conjunction with the symposium was an exhibition of aerospace instrumentation equipment. Flight data recording equipment was well to the fore and exhibited on the stand of S. Davall and Sons Ltd. was the recorder retrieved from the crash of a Vanguard at London Airport last October. Apparently the recorder suffered an impact shock equivalent to a deceleration approaching 1000g for 5ms and during the subsequent fire it was subjected to a temperature of approximately 1000°C. The recording wire however was unbroken and the data has been available for the investigation of the crash.

Details of some of the electronic equipment for the Concorde were also available. Elliott Bros. (London) Ltd. were showing the Elliott AIR-3, a high-density digital recorder accommodating 33 channels of 800 bit per inch information on one inch tape and designed by Elliott-Automation especially for the Concorde. Where necessary signals are converted from analogue to digital form and multiplexed. Information from about 300 points on the aircraft will be monitored, and in some cases the sampling rate will be five times per second. With flight recorders so much in the news it was interesting to note that a paper presented by J. J. Smith (Consolidated Electrodynamics Corporation) discussed the practical aspects of aerospace tape recorders. A point emphasized by the author was the effect of space environment on the performance of a recorder as compared to operation in laboratory conditions. An example given by the author was that where acceleration and vibration cause an increase in power consumption by the recorder due to high radial and thrust loads placed on rotating shafts. The increase in power is that drawn by the motor speed control system which ensures that adequate torque is provided to maintain the speed requirements. The solution was to orientate the drive motor so that the axis of its rotation was parallel with the axis of maximum mechanical energy input. Shock and vibration also introduce mechanical effects which cause flutter (instantaneous departure of the tape from a nominal speed) and dynamic skew (non uniform velocity of the tape across its width) and time displacement error (the integral of flutter). These errors have been minimized by the design of appropriate shock and vibration mounts which has been achieved by close co-operation of the shockmount designer, the recorder manufacturer and the ultimate user.

Looking towards the future, the author outlined new developments which would improve performance. Integrated circuitry with its advantages of reduced weight and space and the ability to incorporate the integrated circuit amplifier in the tape head would make possible better signal-to-noise ratios and a high signal level. Higher longitudinal information packing density is expected and lateral density will depend on the number of tracks per inch. Present magnetic tape limited longitudinal packing where surface magnetization at short wavelengths does not allow optimum use of the energy storage characteristic of the oxide, but new materials are expected to provide a higher figure with

thinner coatings. Problems of lateral density are the need for shielding between adjacent tracks and the guiding of the tape between record and reproduce heads where small deviations are significant. At the moment 42 tracks per inch are available. The author also discussed the possibilities of incremental recorders, fast stop-start recorders which record only the data desired. These offer good possibilities where requirements of an aerospace data recording system are for discontinuous recording or flight editing. Their advantages over continuous recorders are obvious, but at the moment these recorders are limited to the recording of digital information where the requirements of flutter and speed accuracy needed for analogue information are not important factors.

THE MONTH'S CONFERENCES AND EXHIBITIONS

Further details are obtainable from the addresses in parentheses

LONDON

- May 2-9 Olympia
Numerically Controlled Machine Tool Exhibition
(Machine Tool Trades Assoc., 25-28 Buckingham Gate, S.W.1)
May 10-20 Earls Court
Mechanical Handling Exhibition
(Mechanical Handling, Dorset House, Stamford St., S.E.1)
May 23-28 Olympia
I.E.A.-Instruments, Electronics & Automation Exhibition
(Industrial Exhibitions, 9 Argyll St., W.1)

BRIGHTON

- May 15-19 Hotel Metropole
R.T.R.A. Conference
(Radio & Television Retailers' Assoc., 19 Conway St., W.1)

EASTBOURNE

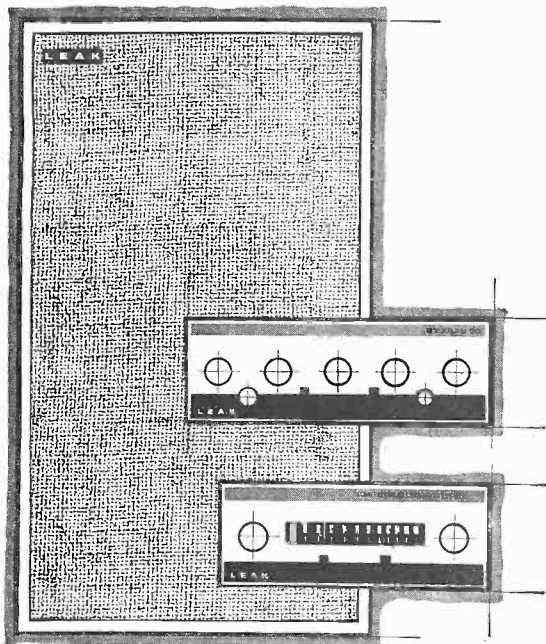
- May 3-5 Congress Theatre
British Joint Computer Conference
(Joint Conference Secretariat, I.E.E., Savoy Pl., London, W.C.2)

SALFORD

- May 5-6 Royal College of Advanced Technology
Ultrasound in Medicine
(Dr. Basil Brown, Dept. of Physics, Royal College of Advanced Technology, Salford, Lancs.)

OVERSEAS

- Apr. 29-May 15 Oslo
Britain 66
(British Overseas Fairs, 21 Tothill St., London, S.W.1)
Apr. 30-May 8 Hanover
Hanover Fair
(Schenkers Ltd., 13 Finsbury Sq., London, E.C.2)
May 2-4 Washington
Communications Satellite Systems Conference
(I.E.E.E., 345 East 47th St., New York, N.Y. 10017)
May 4-6 Washington
Electronic Components Conference
(I.E.E.E.E., 345 East 47th St., New York, N.Y. 10017)
May 10-12 Boston
National Telemetering Conference
(I.E.E.E., 345 East 47th St., New York, N.Y. 10017)
May 22-26 Sydney
Radio & Electronics Engineering Convention
(Instn. of Radio & Electronics Engrs., Box 3120 G.P.O., Sydney, N.S.W.)



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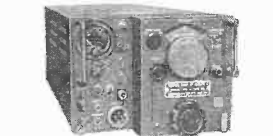
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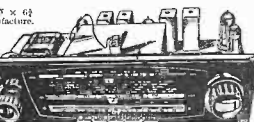
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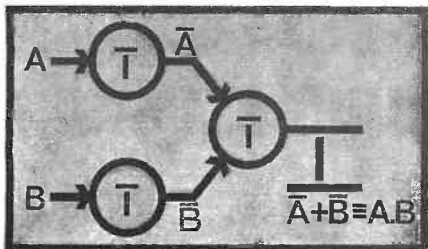
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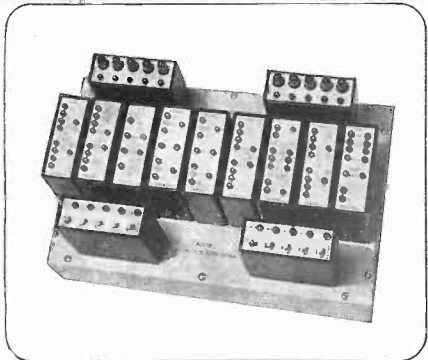
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11M17	7/6	10A7	2/6	10K92	6/6	10P28	8/6	10P22	6/6	11M51	6/6
12X01T	7/6	12A50	6/6	10K93	6/6	10P29	8/6	10P23	6/6	11M52	6/6
1B5	4/6	14A17	4/6	10K94	6/6	10P30	8/6	10P24	6/6	11M53	6/6
1B5	4/6	14A27	4/6	10K95	6/6	10P31	8/6	10P25	6/6	11M54	6/6
1B5	4/6	14A37	4/6	10K96	6/6	10P32	8/6	10P26	6/6	11M55	6/6
1B5	4/6	14A47	4/6	10K97	6/6	10P33	8/6	10P27	6/6	11M56	6/6
1B5	4/6	14A57	4/6	10K98	6/6	10P34	8/6	10P28	6/6	11M57	6/6
1B5	4/6	14A67	4/6	10K99	6/6	10P35	8/6	10P29	6/6	11M58	6/6
1B5	4/6	14A77	4/6	10K100	6/6	10P36	8/6	10P30	6/6	11M59	6/6
1B5	4/6	14A87	4/6	10K101	6/6	10P37	8/6	10P31	6/6	11M60	6/6
1B5	4/6	14A97	4/6	10K102	6/6	10P38	8/6	10P32	6/6	11M61	6/6
1B5	4/6	14A07	4/6	10K103	6/6	10P39	8/6	10P33	6/6	11M62	6/6
1B5	4/6	14A17	4/6	10K104	6/6	10P40	8/6	10P34	6/6	11M63	6/6
1B5	4/6	14A27	4/6	10K105	6/6	10P41	8/6	10P35	6/6	11M64	6/6
1B5	4/6	14A37	4/6	10K106	6/6	10P42	8/6	10P36	6/6	11M65	6/6
1B5	4/6	14A47	4/6	10K107	6/6	10P43	8/6	10P37	6/6	11M66	6/6
1B5	4/6	14A57	4/6	10K108	6/6	10P44	8/6	10P38	6/6	11M67	6/6
1B5	4/6	14A67	4/6	10K109	6/6	10P45	8/6	10P39	6/6	11M68	6/6
1B5	4/6	14A77	4/6	10K110	6/6	10P46	8/6	10P40	6/6	11M69	6/6
1B5	4/6	14A87	4/6	10K111	6/6	10P47	8/6	10P41	6/6	11M70	6/6
1B5	4/6	14A97	4/6	10K112	6/6	10P48	8/6	10P42	6/6	11M71	6/6
1B5	4/6	14A07	4/6	10K113	6/6	10P49	8/6	10P43	6/6	11M72	6/6
1B5	4/6	14A17	4/6	10K114	6/6	10P50	8/6	10P44	6/6	11M73	6/6
1B5	4/6	14A27	4/6	10K115	6/6	10P51	8/6	10P45	6/6	11M74	6/6
1B5	4/6	14A37	4/6	10K116	6/6	10P52	8/6	10P46	6/6	11M75	6/6
1B5	4/6	14A47	4/6	10K117	6/6	10P53	8/6	10P47	6/6	11M76	6/6
1B5	4/6	14A57	4/6	10K118	6/6	10P54	8/6	10P48	6/6	11M77	6/6
1B5	4/6	14A67	4/6	10K119	6/6	10P55	8/6	10P49	6/6	11M78	6/6
1B5	4/6	14A77	4/6	10K120	6/6	10P56	8/6	10P50	6/6	11M79	6/6
1B5	4/6	14A87	4/6	10K121	6/6	10P57	8/6	10P51	6/6	11M80	6/6
1B5	4/6	14A97	4/6	10K122	6/6	10P58	8/6	10P52	6/6	11M81	6/6
1B5	4/6	14A07	4/6	10K123	6/6	10P59	8/6	10P53	6/6	11M82	6/6
1B5	4/6	14A17	4/6	10K124	6/6	10P60	8/6	10P54	6/6	11M83	6/6
1B5	4/6	14A27	4/6	10K125	6/6	10P61	8/6	10P55	6/6	11M84	6/6
1B5	4/6	14A37	4/6	10K126	6/6	10P62	8/6	10P56	6/6	11M85	6/6
1B5	4/6	14A47	4/6	10K127	6/6	10P63	8/6	10P57	6/6	11M86	6/6
1B5	4/6	14A57	4/6	10K128	6/6	10P64	8/6	10P58	6/6	11M87	6/6
1B5	4/6	14A67	4/6	10K129	6/6	10P65	8/6	10P59	6/6	11M88	6/6
1B5	4/6	14A77	4/6	10K130	6/6	10P66	8/6	10P60	6/6	11M89	6/6
1B5	4/6	14A87	4/6	10K131	6/6	10P67	8/6	10P61	6/6	11M90	6/6
1B5	4/6	14A97	4/6	10K132	6/6	10P68	8/6	10P62	6/6	11M91	6/6
1B5	4/6	14A07	4/6	10K133	6/6	10P69	8/6	10P63	6/6	11M92	6/6
1B5	4/6	14A17	4/6	10K134	6/6	10P70	8/6	10P64	6/6	11M93	6/6
1B5	4/6	14A27	4/6	10K135	6/6	10P71	8/6	10P65	6/6	11M94	6/6
1B5	4/6	14A37	4/6	10K136	6/6	10P72	8/6	10P66	6/6	11M95	6/6
1B5	4/6	14A47	4/6	10K137	6/6	10P73	8/6	10P67	6/6	11M96	6/6
1B5	4/6	14A57	4/6	10K138	6/6	10P74	8/6	10P68	6/6	11M97	6/6
1B5	4/6	14A67	4/6	10K139	6/6	10P75	8/6	10P69	6/6	11M98	6/6
1B5	4/6	14A77	4/6	10K140	6/6	10P76	8/6	10P70	6/6	11M99	6/6
1B5	4/6	14A87	4/6	10K141	6/6	10P77	8/6	10P71	6/6	11M100	6/6
1B5	4/6	14A97	4/6	10K142	6/6	10P78	8/6	10P72	6/6	11M101	6/6
1B5	4/6	14A07	4/6	10K143	6/6	10P79	8/6	10P73	6/6	11M102	6/6
1B5	4/6	14A17	4/6	10K144	6/6	10P80	8/6	10P74	6/6	11M103	6/6
1B5	4/6	14A27	4/6	10K145	6/6	10P81	8/6	10P75	6/6	11M104	6/6
1B5	4/6	14A37	4/6	10K146	6/6	10P82	8/6	10P76	6/6	11M105	6/6
1B5	4/6	14A47	4/6	10K147	6/6	10P83	8/6	10P77	6/6	11M106	6/6
1B5	4/6	14A57	4/6	10K148	6/6	10P84	8/6				



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59/6

See pages 86/87
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THE SINCLAIR MICRO FM is more than an F.M. Tuner; more than an F.M. Receiver for it combines the advantages of both with many other unique features to make it the most advanced set of its kind in the world. Anyone can construct it, for unlike other FM constructional kits, the Micro FM needs no aligning and is ready to work as soon as it is finished. Pulse counting detection gives better audio quality than any other discriminator system. Excellent sensitivity assures good reception using no more than the small telescopic aerial

included in the kit in all but the worst reception areas. When built, the Sinclair Micro FM has all the appearance of a professionally engineered set both inside and out. Its distinctive, elegant exterior makes it particularly pleasing to own and to operate whether as a tuner for amplifier or tape recorder or independently as a self-contained pocket F.M. portable. Yet with all these features the Micro FM costs pounds less and enables you to enjoy F.M. reception to the full at once.

ONLY THE SINCLAIR MICRO FM HAS ALL THESE EXCLUSIVE FEATURES

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- Requires no aligning
- Supply voltage—9 V. from self-contained standard battery
- Consumption—5 mA.
- Sensitivity—Typically 3 microvolts
- Signal to Noise Ratio—30 dB at 30 microvolts
- Audio frequency response—10-20,000 c/s.
- Two Audio Outputs—One for feeding to amplifier or tape recorder. One for feeding to earpiece to enable set to be used as a portable receiver.
- A.F.C.
- Black plastic case with brushed and polished aluminium front panel.
- Inserting plug of earpiece or tuner lead switches set ON

TECHNICAL DESCRIPTION

THE SINCLAIR MICRO FM is a completely self-contained double-purpose FM superhet housed within a case less than 3in. high x 1 1/2in. wide with a depth of 2in. It uses 7 transistors and 2 diodes in new circuitry. The R.F. amplifier is followed by a self-oscillating mixer and three stages of I.F. amplification which dispenses with I.F. transformers and all problems of alignment. The final I.F. amplifier produces a square wave of constant amplitude which is converted into uniform pulses so arranged that the original modulation is reproduced exactly.

A pulse-counting detector ensures absolute linearity and therefore better audio quality at the output stages. After equalisation the signal is channelled to one output for feeding to amplifier or recorder and to another in which the receiver's own audio amplifying stage enables the Micro FM to be used as an independent self-contained pocket portable. A.F.C. is used to lock the programme tuned in; the telescopic aerial included with the kit will be found sufficient for all but the worst signal areas.

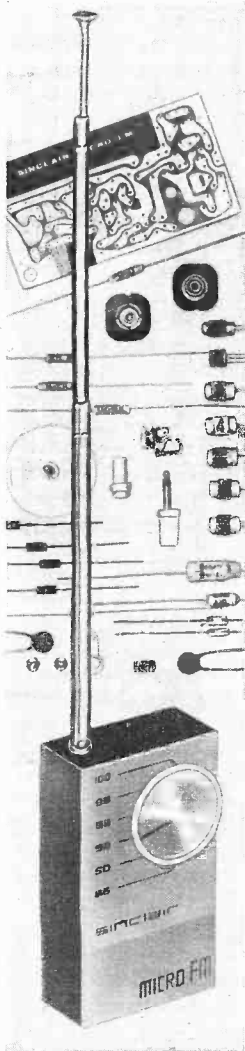
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Complete kit of parts, including transistors, aerial, case, front panel, dial, earpiece and clearly presented stage-by-stage instructions manual.

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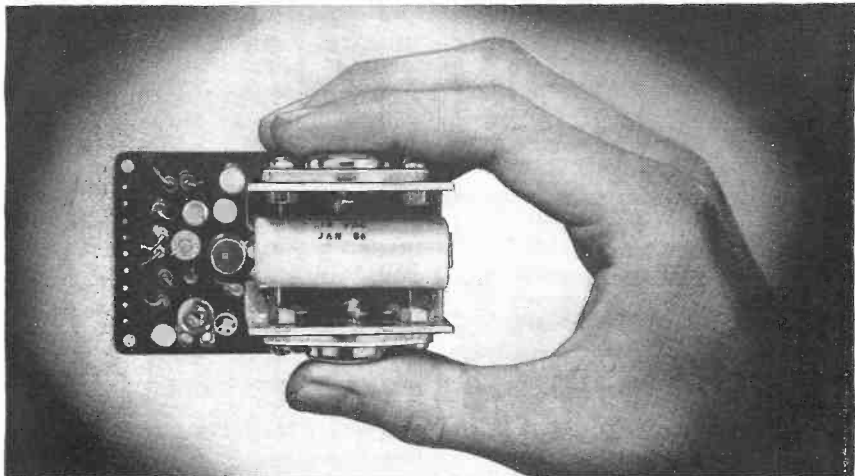
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More power per square inch than any other amplifier in the world!

The Z.12 has quality. The Z.12 has power, and it is certainly marvellously compact. In fact, this wonderful unit which is complete with its own high gain pre-amplifier and ready to connect to any input will give you an output of up to **SIX WATTS PER SQUARE INCH** of its total size—a standard of performance unequalled by anything in its class. And because of its size and excellent circuitry, you can now use quality amplification in applications never before possible.

The Sinclair Z.12 uses 8 special H.F. transistors with generous negative feedback and ultralinear class B push-pull output to achieve the highest possible standards of quality. The unit can be powered from 6 to 20 v. d.c. and for those not using a battery, the new PZ.3 will be found ideal. Signal to noise ratio is better than 60dB, and the output may be fed directly into any load from 3 to 15 ohms, or two 3 ohms speakers may be used in parallel. The Manual included with the Z.12 gives full details of matching tone and volume control circuits for mono and stereo, together with multi-input switching facilities.

- ⊗ 12 WATTS R.M.S. OUTPUT
Continuous Sine Wave (24 w. Peak)
15 Watts R.M.S. Music Power (30 w. Peak)
- ⊗ SIZE 3" x 1 1/4" x 1 1/4"
- ⊗ IDEAL FOR 12 V. OPERATION
- ⊗ INPUT 2mV. INTO 2K OHMS
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- ⊗ SUITABLE FOR HI-FI, TUNER, GUITAR, P.A., INTERCOM, ETC.

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sinclair Z.12

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HI-FI AMPLIFIER AND
PRE-AMPLIFIER



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SEE PAGE 85

WW-129 FOR FURTHER DETAILS.

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- 5 amp. c/o Sub-miniature Micro-switch.

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★ 1/2 milips
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Single Throw 1/6 each

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- ★ 10 amp. 2 circuit
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- ★ COMPARE OUR SPEC. & OUR PRICES WITH OTHER SIMILAR TYPES.

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V-10-1A Solder Tags 1/11 each per 1,000



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- ★ 4 gms. Wire actuator.
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- ★ 15 amp. c/o micro-switch fitted.

- ★ LARGE RANGE OF A.C. & D.C. OPERATIONS.

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- ★ 5 AMPS. DOUBLE POLE, DOUBLE THROW SWITCHING INCORPORATED.
- Accuracy 1%.
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- ★ FOR BATCHING, CONVEYORS, MACHINE TOOL CONTROL, PACKAGING, SORTING, etc.
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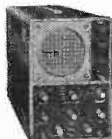
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Relay protected against overload. Less than 150 microvolts ripple. Regulation less than ± 15 millivolts. Output impedance less than 0.1 ohm. Compact 9 1/2" x 6 1/2" x 1 1/2". Weight 1 1/2 lb. net. Full details on request.

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Casters or legs available as optional extras.

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A new concept in Heathkit loudspeaker design. The cabinet shell is assembled and finished in superb Queensland walnut veneer. Two specially designed speakers, a 12in. bass unit and 4in. mid/high frequency unit and an L.C. cross-over network provide the smooth 30-17,000 c/s. frequency response. Its professional cabinet styling will blend with both traditional and contemporary decor. 15 ohm nominal impedance. Size 28" x 17" x 7 1/2" deep.

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- * Printed circuit board ensures consistent performance, easiest possible construction.
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- * Attractive anodised aluminium front panel.
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KIT Model MA-12**

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(Price includes carriage and duty)



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A range of equipment cabinets is now available. Designed for maximum operating convenience or for where room space is an over-riding consideration, this range includes easy-to-build kits or ready-assembled cabinets in the white for finish to own requirements.

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**HI-FI STEREO AMPLIFIER
KIT Model S-99**

18 w. output (9 per channel with 0.2 per cent. distortion at 9 w. per channel). It has ganged controls. Stereo/Mono gram, radio and tape recorder inputs and push-button selection. Ultra-linear push-pull output. P.C. board. Attractive Perspex front panel with golden surround and grey metal cabinets.
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Ducted-port bass reflex cabinet "in the white". Frequency response is 40-16,000 c/s. Power rating 10 watts. Matched speaker units 8 in. high flux (2,000 lines) with hyperbolic cone and 4 in. wide angle dispersion type for higher frequencies.
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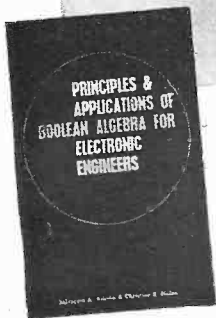
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6.0001	2m2+HD	M1105	15
2.50001	2C2K	M1417	18.6
5.00001	3CO	M1424	17.6
2.50001	2mHD	M1458	15
4.023	2CO2K	M1462	15
2.50002	1m1HD	M1482	19.5
18093	3C2K	M1487	17.6
47011	2C2K	M1490	19.5
47011	2C2K	M1493	19.5
07003	4K	M1507	19.5
S.T.C. 45G	1mHD	4180R13	19.5
1204	1mHD	4180E13	19.5
2003	2CO2M	4190HD	19.5
2.50013	2CO	4184GE	19.5
7.50013	2CO	17.2250A21	17.6
7.50013	2CO	2.2250A21	17.6
		N2350A21	17.6

ERICSSON
 7.50013 2CO
 7.50013 2CO

RECTIFIER UNITS. Input 230/440 v. A.C. output 50 v. 60 v. 100 v. D.C. £35.

FREQUENCY METERS. 45-50 cycles per second, 230 v., 6 1/8 in. Flush Round. Brand new, £10/10/-.

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Amps 0.04. All D.C. 27.6 ea.

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Volts 0.2, 45/-; 0.250-0.250, 60/-

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Input 230 v. A.C. Output variable 0-260 v. A.C. at 2.5 amp. Fitted in beautifully finished steel case. Complete with voltmeter, pilot lamp, fuse, switch, carrying handle, £9/17/6 P. & C. 5/-.

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Designed for Panel Mounting

Input 230 v. A.C. 50/60 Hz. Output variable.

0-260 v. 1 amp. £3 3 0

1 amp. £4 17 6

2 1/2 amp. £5 12 6

1 AMP. 1 AMP.



50 AMPS.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

1 AMP.

INPUT 230 v. A.C. 50/60 Hz. BRAND NEW. Carriage Paid. Buy direct from the importer, keener prices in the country. All Types (and Spares) from 1 to 50 amp. available from stock.

0-260 v. at 1 amp.	4	10 0
0-260 v. at 2.5 amps.	5	17 6
0-260 v. at 4 amps.	8	7 6
0-260 v. at 5 amps.	9	6 0
0-260 v. at 8 amps.	13	10 0
0-260 v. at 10 amps.	17	0 0
0-260 v. at 12 amps.	19	0 0
0-260 v. at 15 amps.	22	0 0
0-260 v. at 20 amps.	32	0 0
0-260 v. at 37.5 amps.	65	0 0
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10 ohm, 1.5 a.; 25 ohm, 1 a.; 50 ohm, .75 a.; 100 ohm, .5 a.; 250 ohm, .3 a.; 500 ohm, .2 a.; 1,000 ohm, .15 a.; 1,500 ohm, .12 a.; 2,500 ohm, .1 a.; all at 14/6. P. & P. 1/6.



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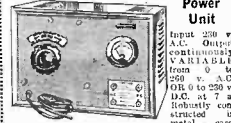
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Input 230 v. A.C. Output continuously variable from 0 to 250 v. D.C. at 7 A. Industrially constructed in metal case, complete with safety fuse, on indicator and voltmeter. Size: 17 x 12 1/2 in. Weight 80 lb. Price £24/10/-, Carriage 20/-

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Transformer to suit the above. Input 200-240 v. A.C., 12 volt 36 watts, 16/6. P. & P. 1/6
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100 ohm, 3 c/o, 2 leads, 2 1/2 watt, 8 ohm, 6/6. 500 ohm, 1 heavy duty c/o, 600 ohm, 2 heavy duty c/o, 2000 ohm, 1 make, 1 break, 25,000 ohm, 2 make, 2 break, All at 12/6 each, 20,000 ohm, 2 heavy duty make. Plus 1/6 P. & P.

SOLENOID. Overall length 3 1/2 in., stroke 1/2 in. Maximum push 8 oz. 12-24 v. D.C. operation. D.C. resistance 25 ohms, 2 1/2 watt, P. & P. 1/6.

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Four 5 v. units connected in series connected in high capacity polyethylene enclosure. Fine lead connections.

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200-250 v. A.C. Powerful 2-speed motor, 11,000 and 13,000 R.P.M., 17/6, plus P. & P. 2/6.
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A complete complete apparatus, comprising a robustly built transformer and electro-magnetic with removable coils and pole pieces. Kit topped for 230 v., 220 v., 110 v., 115 v.; 6, 12, 24, 36, 48, 60, 72, 84, 96, 110, 120, 132, 144, 156, 168, 180, 192, 216, 228, 240, 252, 264, 276, 288, 300, 312, 324, 336, 348, 360, 372, 384, 396, 408, 420, 432, 444, 456, 468, 480, 492, 504, 516, 528, 540, 552, 564, 576, 588, 600, 612, 624, 636, 648, 660, 672, 684, 696, 708, 720, 732, 744, 756, 768, 780, 792, 804, 816, 828, 840, 852, 864, 876, 888, 900, 912, 924, 936, 948, 960, 972, 984, 996, 1,000. All at 12/6 each, plus 1/6 P. & P.



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B 6 hole, 25 watt all non-bridging, full wires, plain coil, 950 ohm. Unused 43/2/6. P. & P. 4/6.

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All 1/2 watt 5% to 10% each, 4.3 v., 4.7 v., 5.1 v., 5.6 v., 6.2 v., 6.6 v., 7.5 v., 8.2 v., 9.1 v., 10 v., 12 v., 15 v., 16 v., 22 v.

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All primaries 220-240 volts.

Type No.	Sec. Taps	Price	Car.
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0.5 amp. D.C.M.I. 2 1/2 in. Rnd. 11/6
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VAN DE GRAAF ELECTROSTATIC GENERATOR. Fitted with motor drive for 230 v. A.C. giving a potential of 50,000 volts. Supplied absolutely complete including accessories for carrying out a number of interesting experiments, and full instructions. This instrument is completely safe, and ideally suited for school demonstrations. Price 65/6, plus 4/- P. & P. 1/6 on request.

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2 1/2 Edgewise Meter
250 Micro-amp, sealed in dB's. New 29/6, P. & P. 1/6

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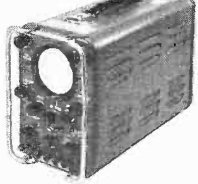
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A good selection is available for callers at from £30. All are in good condition and are functioning perfectly. Realign.

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Input 220 volts to 250 volts at 50 cycles tapped every 5 volts. Secondary, 50 volts at 15 amp. Very conservatively rated. Size 7 1/2 x 5 1/2 x 7in. Wt. 28 lb. Brand new. £5/17/6. Carr. 7/6.

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A high frequency version of the well known BC-221. Accuracy 0.05%. Frequency range 80 to 1,000 Mc/s. As new condition with original calibration book. Tested and guaranteed. £75.

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A 12in. demonstration oscilloscope with all controls at rear. BRAND NEW £80.

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S61. Dual range 0 and 0-100 v. D.C. FSD 1 mA. 3in. scale. Recent manufacture. Ideal for schools. Complete in super quality carrying case, with test leads and leads.

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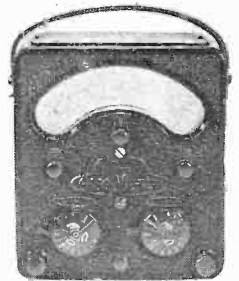
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MARCONI TF 937. Covers 85 Kc/s to 30 Mc/s in 8 switched ranges. Effective length of film scale is 50ft. Output level variable in 4 dB steps from 1uV to 100 mV (75Ω). Also 1V Outputs down to 0.1uV from an outlet at 7.5Ω. Int. mod. at 400 c/s, 1 Kc/s, 1.6 Kc/s and 3 Kc/s. FM frequencies above 394 Kc/s. Variable mod. depth and deviation. Crystal calibrator 200 Kc/s and 2 Mc/s. Monitor speaker for beat detection. Fully metered, blower cooled, Panclastic. A.C. mains 100 to 150 and up to 250 volts, 45 to 100 c/s. 17 1/2 x 20 1/2 x 17 1/2in. Weight 117 lbs. Fully tested and guaranteed. Under 10% of original price. £65. Carr. 50/-.
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A modern instrument, all solid state, which accurately measures the capacity of electrolytic condensers from 0.1uF to 1,000.0uF under operating conditions. Leakage current and polarizing voltage are separately metered. Inductances from 100 mH to 100 H can also be measured with currents up to 100 mA. A.C. mains operation. Untested with handbook. £100.
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A matching instrument to the above. All solid state. Mains operation. Measures from 0.002pF to 100.0F. Unused with handbook. £100.

V.H.F. SIGNAL GENERATOR DIRECTLY CALIBRATED. Covers 10 to 300 Mc/s (4 bands). DIRECTLY calibrated. Int. Mod. at 400, 1,000 and 5,000 c/s. Attenuated or forced out. Guaranteed over-hauled, accurate and in perfect working order. £32,10/-, Carr. £1.
HAMMARLUND SP-600-JX
20 valve dual conversion superhet receiver covering 540 kc/s to 54 Mc/s in 6 bands. This is a professional quality receiver of recent design with stability of 0.01% or better. Second channel rejection 74 dB down and spurious responses are at least 100 dB down. Bandwidth from 200 c/s to 13 kc/s. Crystal filter with crystal phasing control. Operates directly from A.C. mains 90-260 v. 50-50 c/s. Original cost £510. Supplied overhauled and in first class working order, £100. Fuller details on request.

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These receivers are in as new condition. They have 3 wavebands with a frequency coverage of 190-570 metres, 2.3-7.3 Mc/s, 7.0-23 Mc/s, R.F. stages, 6 valves. Output for 3 ohm speaker. Require external Power Supply or can be fitted with internal Mains Power Supply for £2 extra. Circuit simplified. Fully tested prior to dispatch. £8/19/6. Carriage 10/-. Vibrator Supply Unit for operation from 12 v. car battery for caravans or boats, 15/6. Carriage 7/6.
DISTORTION FACTOR METER
MARCONI TF-142E. This instrument measures the percentage of total harmonic distortion in the fundamental frequency range 100 to 8,000 c/s. The lowest scale engraving is 0.05%. Will handle 2 watts (continuous) and will give satisfactory readings with only 1mV input. Mains operated. Output impedance 600 ohms. Very good condition. £30. Carr. £1.
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B. & R. RELAYS TYPE CO3. 230 volt A.C. coils. Very heavy duty contact assembly. 4 makes. Size 4in. x 3in. x 2 1/2 in. weight 2 lb. BRAND NEW. 27/6. P. & P. 2/6.
HICKOCK OSCILLOSCOPE OS-88/U
A high grade general purpose instrument made to exacting U.S.A. Navy specifications. Detachable cover with carrying handle. Compact (13 1/4 x 6 x 8 1/2 in.) weight 17 lbs. Green trace 3 in. tube. Bandwidth amplifier D.C. to 2 Mc/s (D.C. coupled). Sensitivity 40 mV/cm. "X" amp. can be used separately, similar spec to "Y" amp. Leads are housed in case. For A.C. mains 105 to 125 v. 50 to 1,000 c/s. BRAND NEW, tested and guaranteed. £25. Carr. 10/-. Auto transformer 15/6 extra.

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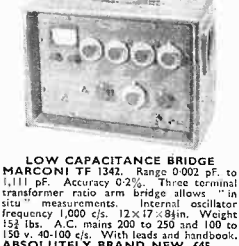
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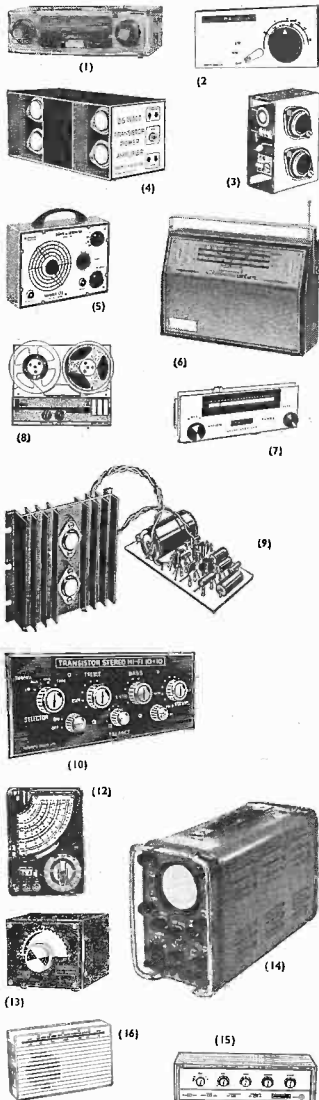
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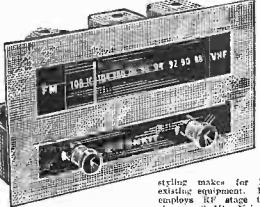
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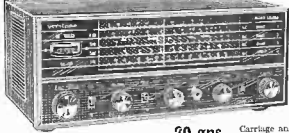
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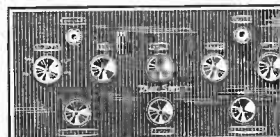
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TWO VALVE PRE-AMPLIFIER

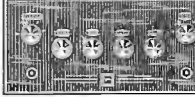


Specially designed for use with the Mullard sets of 3, 19 and 20 watt amplifiers but entirely suitable for all Hi-Fi power units not requiring input of more than 200 mV for full output. Features include input for crystal and variable reluctance pickups with 21KΩ equalisation, plus radio, sensitive microphone, and tape replay direct or indirect. Controls include 6-position selector, volume and wide range bass and treble boost. Volume 2 x 120Ω, size 9 1/2 x 4 1/2 x 2 1/2 in. high. Power 10 x 2 1/2 in. Requires 300 v. at 3 mA and 0.2 v. at 0.05 amp.

Assembled and tested £9 - 10 - 0

Add 4/- carriage. Kit £6 - 6 - 0

THREE VALVE PRE-AMPLIFIER



Specially recommended for use with the 40-100 amplifier, but also suitable for use with all Mullard series mono amplifiers and any high quality unit requiring an input of up to 200 mV for full output. Features include wide range bass and Treble controls, high and low pass filters, auxiliary input and record output sockets, switched inputs for crystal or reluctance pickups, radio, tape replay direct or indirect. Power required 250 v. at 0.6 mA and 0.3 v. at 1 amp. Volume 100Ω and 100KΩ, size 11 x 4 1/2 x 2 1/2 in. high. Power 12 x 4 1/2 in.

Assembled and tested £13 - 13 - 0

Add 5/- carriage. Kit £10 - 10 - 0

Actually twin Two-Valve Pre-amplifiers which together form an extremely versatile and sensitive input arrangement for the 7m plus Ten amplifier in place of the Passive Control Unit. Features include dual inputs for mono and stereo Alpacado, Variable Reintance, Crystal and Ceramic pickups, dual inputs direct and amplified, plus 100Ω and 100KΩ, plus 100Ω and 100KΩ, plus 100Ω and 100KΩ, plus 100Ω and 100KΩ.

Assembled and tested and supplied complete with Ten plus Two Amplifier. £34 - 0 - 0

Kit Add 10/8 Carriage. £27 - 0 - 0

Dual channel pre-amplifier only—Assembled and tested. £15 - 0 - 0

Kit Add 5/- Carriage. £12 - 10 - 0

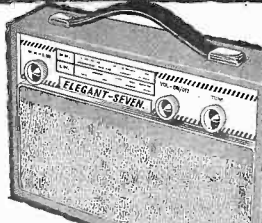
'ELEGANT SEVEN' MK. II

POWER SUPPLY KIT to purchasers of Elegant Seven parts, incorporating mains-transformer, etc. A.C. mains 200-250 v. Output 9 v. 50 mA 7/6d. extra.

COMBINED PORTABLE & CAR RADIO. The Radio with the STAR features, 4in. SPEAKER.

- * 7-transistor superhet. Output 350mW.
- * Wooden cabinet fitted handle with silver coloured fittings. Size 12in. x 8in. x 3in.
- * Horizontal tuning scale, size 1 1/2in. x 2 1/2in. in silver and black lettering.
- * All stations clearly marked.
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- * I.F. Neutralisation on each stage. 460 kc/s.
- * D.C. coupled output stage with separate A.C. negative feedback.
- * All components, ferrite rod and tuning assembly mounted on printed board.
- * Operated from PP9 battery.
- * Full comprehensive instructions and point-to-point wiring diagrams.
- * Printed circuit board, built with all component values.
- * Fully tunable over medium and long waveband.
- * Car aerial socket.
- * Full after-sale service.

Parts list and circuit diagram 2/6. Free with parts.



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£4.4.0 (plus 6/6 P. & P.)

'MAYFAIR' 5-Transistor TAPE RECORDER

Capstan-driven, battery operated 7 1/2 and 3 1/2 i.p.s. Precision made. Push-button controls. High quality 2 1/2in. speaker. Push-pull circuit. Output: 400 mW. Frequency response: 200-7,000 kc/s. Fast rewind. Up to 1 hour twin track playing time. Automatic erasing for re-recording. Dimensions 8in. x 11in. x 3in. Weighs only 7lb. Takes 5in. spools.



plus 7/6 P. & P.

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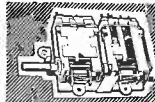
40W. FLUORESCENT LIGHT KIT

Incorporating GEC Choke size 8 1/2in. x 1 1/2in. x 1 1/2in., 2 bi-pin holders, starter and starter holder. **11/6** P. & P. 4/6.
Similar to above: 80W. Fluorescent Light Kit incorporating GEC choke size 11 1/2in. x 1 1/2in. x 1 1/2in. **17/6** P. & P. 5/6.



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The above items: P. & P. 6/-.



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Complete with PC.88 and PC.86 Valves. Full variable tuning. New and unused. Size 4 1/2in. x 5 1/2in. x 1 1/2in. Complete with circuit diagram. Plus 2/6 P. & P. **35/-**



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5 1/2in. L.P. 1,200ft. 11/6	5in. T.P. 2,400ft. 32/6
7in. L.P. 1,800ft. 18/6	7in. T.P. 3,600ft. 42/6

P. & P. on each 1/6, 4 or more post free.

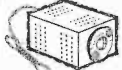
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In metal case, size 3 1/2in. x 2 1/2in. x 2in. incorporating mains transformer, rectifier and condensers. 230/250 A.C. Mains. Output: 9 v. 100 mA. Price 10/6 plus 3/-, P. & P.



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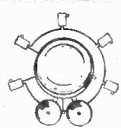
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Push-pull X amplifier; Fly-back suppression; Internal Time-base Scan Wave form available for external use; pulse output available for checking TV line C/P Transformers, etc. Provision for external I.F. and C.R.T. Brightness Modulation. A.C. mains 200/250 v. £18/18/-, P. & P. 10/-, FULL 12 MONTHS' GUARANTEE INCLUDING VALVES and TUBE.



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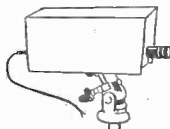


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LOW COST BEULAH D.80

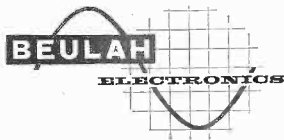
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Britain's lowest priced Domestic Closed Circuit TV Camera is fully transistorised and is designed to operate in conjunction with standard VHF domestic television receivers on either 405, 525 or 625 line systems.

The performance is outstanding, and it transmits signals over 1 mile of co-axial without amplification and consumes only 15 watts. Supplied complete with a high quality 25 mm. f/1.9 Television Lens on type C mount with fully adjustable focus and iris. 2 1/2 Mc/s resolution. Operates on 200-250 volts a.c. Mains supply (or 110 volts to order). Price 79 gns.



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Suppliers to: Ministry of War, G.P.O., Government Communication Headquarters, North Thames Gas Board, B.A.C., Education Committees, Ford Motor Company and many other large industrial concerns.

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Buy it and build it in easy stages with the famous BEUKITS—

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Plus 7/6 P. & P. for each kit. (Kits cannot be split. Instruction Manual cannot be supplied separately.)

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A PREAMPLIFIER FOR PERFECTIONISTS

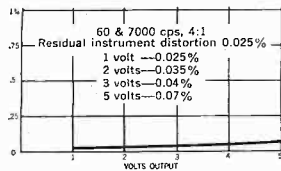


PAS-3X

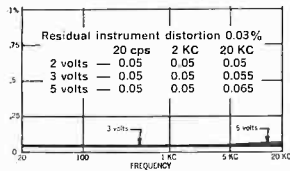
For years since its introduction, the Dynaco preamplifier design has been generally accepted as one in which the noise and distortion are so low, and the sound quality so high that attempts to improve it would be laboratory exercises rather than commercial enterprises.

However, our avowed philosophy of perfectionism has kept us looking for some way to improve the circuit—and this has now led to the first major change in our preamplifier design since it was initiated. This development (on which patents are pending) is applicable to all continuous tone control systems and immediately makes them superior to the far more costly switch type controls. We have kept the infinite resolution capability of the continuous control, but all frequency and phase discriminating networks are removed from the circuit when the control is rotated to its mechanical center. This new design, which combines the advantages of both step-type and continuous tone controls is now available at no increase in price. And, a conversion kit TC-3X is available at nominal cost to update any Dynaco PAS-2 or PAS-3.

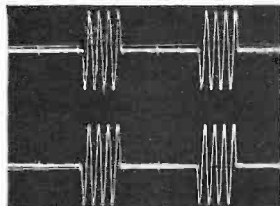
Can you hear the difference? We doubt it. The preamp was amazingly good in the past. We have improved it for the sake of improvement, not because we think it needed it. It has always surpassed every other preamplifier without regard to cost. And, it is superior on more than measurements—listening tests prove that the Dynaco preamp adds no coloration to the sound and that its inclusion in the hi fi chain is undetectable. Partially diagrammed below is the performance you can expect from the PAS-3X—why you can never get better overall quality regardless of how much money you spend. Complete specifications on request.



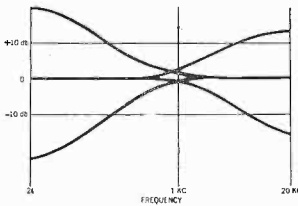
Intermodulation Distortion



Harmonic Distortion



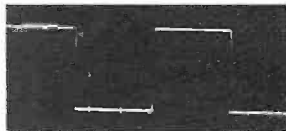
Four cycle 20KC tone burst from generator (above) matches PAS-3X (below)



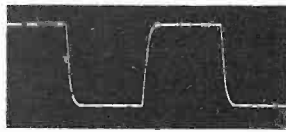
Tone Control Range



10 cps Square Wave



1 KC Square Wave



10 KC Square Wave

There are Dynakit amplifiers in all power brackets which will do justice to the perfectionist's preamplifier.

All are rated for RMS continuous power.



2 Mark III
60 watts/
channel



Stereo 70
35 watts/
channel



Stereo 35
17.5 watts/
channel

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1A7U	7/6	9P7	7/6	DK59	6/6	EM37	7/6	R19	6/9	TYPE
1B2	5/6	10A5	3/6	DK61	5/6	EM40	8/6	R21	6/9	TYPE
1B6	6/6	10C3	12/6	DM70	6/6	EV1	5/6	R51	2/6	Output
1H6T	7/6	10E7	11/6	DM75	6/6	EV2	5/6	R55	2/6	Output
1L1	2/6	10F1	1/6	DY6	6/6	EV3	5/6	T41	1/6	500 v.c.k. 2
1L5	4/6	10F5	6/6	BY87	5/6	EV4	5/6	T104	7/6	100 v.c.k. 2
1L5C	4/6	10F5B	6/6	Z80F	2/6	EV5	5/6	T111	15/6	100 v.c.k. 2
1N8GT	8/6	10P13	12/6	EV8F	8/6	EV7	6/6	T123	6/6	100 v.c.k. 2
1S5	6/6	10P14	11/6	EV8G	7/6	EV8	6/6	T125	5/6	100 v.c.k. 2
1B4	5/6	12A0	5/6	EV8P	10/6	EV9	5/6	T127	5/6	100 v.c.k. 2
1S5B	6/6	12A0B	5/6	EV9	10/6	EV10	5/6	T128	5/6	100 v.c.k. 2
1F4	2/6	12A2B	6/6	EV6	6/6	EV11	5/6	T129	5/6	100 v.c.k. 2
7E4	5/6	12A2C	6/6	EV12	5/6	EV12	5/6	T130	5/6	100 v.c.k. 2
7E4B	5/6	12A2D	6/6	EV13	5/6	EV13	5/6	T131	5/6	100 v.c.k. 2
7E4C	5/6	12A2E	6/6	EV14	5/6	EV14	5/6	T132	5/6	100 v.c.k. 2
7E4D	5/6	12A2F	6/6	EV15	5/6	EV15	5/6	T133	5/6	100 v.c.k. 2
7E4E	5/6	12A2G	6/6	EV16	5/6	EV16	5/6	T134	5/6	100 v.c.k. 2
7E4F	5/6	12A2H	6/6	EV17	5/6	EV17	5/6	T135	5/6	100 v.c.k. 2
7E4G	5/6	12A2I	6/6	EV18	5/6	EV18	5/6	T136	5/6	100 v.c.k. 2
7E4H	5/6	12A2J	6/6	EV19	5/6	EV19	5/6	T137	5/6	100 v.c.k. 2
7E4I	5/6	12A2K	6/6	EV20	5/6	EV20	5/6	T138	5/6	100 v.c.k. 2
7E4J	5/6	12A2L	6/6	EV21	5/6	EV21	5/6	T139	5/6	100 v.c.k. 2
7E4K	5/6	12A2M	6/6	EV22	5/6	EV22	5/6	T140	5/6	100 v.c.k. 2
7E4L	5/6	12A2N	6/6	EV23	5/6	EV23	5/6	T141	5/6	100 v.c.k. 2
7E4M	5/6	12A2O	6/6	EV24	5/6	EV24	5/6	T142	5/6	100 v.c.k. 2
7E4N	5/6	12A2P	6/6	EV25	5/6	EV25	5/6	T143	5/6	100 v.c.k. 2
7E4O	5/6	12A2Q	6/6	EV26	5/6	EV26	5/6	T144	5/6	100 v.c.k. 2
7E4P	5/6	12A2R	6/6	EV27	5/6	EV27	5/6	T145	5/6	100 v.c.k. 2
7E4Q	5/6	12A2S	6/6	EV28	5/6	EV28	5/6	T146	5/6	100 v.c.k. 2
7E4R	5/6	12A2T	6/6	EV29	5/6	EV29	5/6	T147	5/6	100 v.c.k. 2
7E4S	5/6	12A2U	6/6	EV30	5/6	EV30	5/6	T148	5/6	100 v.c.k. 2
7E4T	5/6	12A2V	6/6	EV31	5/6	EV31	5/6	T149	5/6	100 v.c.k. 2
7E4U	5/6	12A2W	6/6	EV32	5/6	EV32	5/6	T150	5/6	100 v.c.k. 2
7E4V	5/6	12A2X	6/6	EV33	5/6	EV33	5/6	T151	5/6	100 v.c.k. 2
7E4W	5/6	12A2Y	6/6	EV34	5/6	EV34	5/6	T152	5/6	100 v.c.k. 2
7E4X	5/6	12A2Z	6/6	EV35	5/6	EV35	5/6	T153	5/6	100 v.c.k. 2
7E4Y	5/6	12A30	6/6	EV36	5/6	EV36	5/6	T154	5/6	100 v.c.k. 2
7E4Z	5/6	12A31	6/6	EV37	5/6	EV37	5/6	T155	5/6	100 v.c.k. 2
7E5	5/6	12A32	6/6	EV38	5/6	EV38	5/6	T156	5/6	100 v.c.k. 2
7E5B	5/6	12A33	6/6	EV39	5/6	EV39	5/6	T157	5/6	100 v.c.k. 2
7E5C	5/6	12A34	6/6	EV40	5/6	EV40	5/6	T158	5/6	100 v.c.k. 2
7E5D	5/6	12A35	6/6	EV41	5/6	EV41	5/6	T159	5/6	100 v.c.k. 2
7E5E	5/6	12A36	6/6	EV42	5/6	EV42	5/6	T160	5/6	100 v.c.k. 2
7E5F	5/6	12A37	6/6	EV43	5/6	EV43	5/6	T161	5/6	100 v.c.k. 2
7E5G	5/6	12A38	6/6	EV44	5/6	EV44	5/6	T162	5/6	100 v.c.k. 2
7E5H	5/6	12A39	6/6	EV45	5/6	EV45	5/6	T163	5/6	100 v.c.k. 2
7E5I	5/6	12A40	6/6	EV46	5/6	EV46	5/6	T164	5/6	100 v.c.k. 2
7E5J	5/6	12A41	6/6	EV47	5/6	EV47	5/6	T165	5/6	100 v.c.k. 2
7E5K	5/6	12A42	6/6	EV48	5/6	EV48	5/6	T166	5/6	100 v.c.k. 2
7E5L	5/6	12A43	6/6	EV49	5/6	EV49	5/6	T167	5/6	100 v.c.k. 2
7E5M	5/6	12A44	6/6	EV50	5/6	EV50	5/6	T168	5/6	100 v.c.k. 2
7E5N	5/6	12A45	6/6	EV51	5/6	EV51	5/6	T169	5/6	100 v.c.k. 2
7E5O	5/6	12A46	6/6	EV52	5/6	EV52	5/6	T170	5/6	100 v.c.k. 2
7E5P	5/6	12A47	6/6	EV53	5/6	EV53	5/6	T171	5/6	100 v.c.k. 2
7E5Q	5/6	12A48	6/6	EV54	5/6	EV54	5/6	T172	5/6	100 v.c.k. 2
7E5R	5/6	12A49	6/6	EV55	5/6	EV55	5/6	T173	5/6	100 v.c.k. 2
7E5S	5/6	12A50	6/6	EV56	5/6	EV56	5/6	T174	5/6	100 v.c.k. 2
7E5T	5/6	12A51	6/6	EV57	5/6	EV57	5/6	T175	5/6	100 v.c.k. 2
7E5U	5/6	12A52	6/6	EV58	5/6	EV58	5/6	T176	5/6	100 v.c.k. 2
7E5V	5/6	12A53	6/6	EV59	5/6	EV59	5/6	T177	5/6	100 v.c.k. 2
7E5W	5/6	12A54	6/6	EV60	5/6	EV60	5/6	T178	5/6	100 v.c.k. 2
7E5X	5/6	12A55	6/6	EV61	5/6	EV61	5/6	T179	5/6	100 v.c.k. 2
7E5Y	5/6	12A56	6/6	EV62	5/6	EV62	5/6	T180	5/6	100 v.c.k. 2
7E5Z	5/6	12A57	6/6	EV63	5/6	EV63	5/6	T181	5/6	100 v.c.k. 2
7E6	5/6	12A58	6/6	EV64	5/6	EV64	5/6	T182	5/6	100 v.c.k. 2
7E6A	5/6	12A59	6/6	EV65	5/6	EV65	5/6	T183	5/6	100 v.c.k. 2
7E6B	5/6	12A60	6/6	EV66	5/6	EV66	5/6	T184	5/6	100 v.c.k. 2
7E6C	5/6	12A61	6/6	EV67	5/6	EV67	5/6	T185	5/6	100 v.c.k. 2
7E6D	5/6	12A62	6/6	EV68	5/6	EV68	5/6	T186	5/6	100 v.c.k. 2
7E6E	5/6	12A63	6/6	EV69	5/6	EV69	5/6	T187	5/6	100 v.c.k. 2
7E6F	5/6	12A64	6/6	EV70	5/6	EV70	5/6	T188	5/6	100 v.c.k. 2
7E6G	5/6	12A65	6/6	EV71	5/6	EV71	5/6	T189	5/6	100 v.c.k. 2
7E6H	5/6	12A66	6/6	EV72	5/6	EV72	5/6	T190	5/6	100 v.c.k. 2
7E6I	5/6	12A67	6/6	EV73	5/6	EV73	5/6	T191	5/6	100 v.c.k. 2
7E6J	5/6	12A68	6/6	EV74	5/6	EV74	5/6	T192	5/6	100 v.c.k. 2
7E6K	5/6	12A69	6/6	EV75	5/6	EV75	5/6	T193	5/6	100 v.c.k. 2
7E6L	5/6	12A70	6/6	EV76	5/6	EV76	5/6	T194	5/6	100 v.c.k. 2
7E6M	5/6	12A71	6/6	EV77	5/6	EV77	5/6	T195	5/6	100 v.c.k. 2
7E6N	5/6	12A72	6/6	EV78	5/6	EV78	5/6	T196	5/6	100 v.c.k. 2
7E6O	5/6	12A73	6/6	EV79	5/6	EV79	5/6	T197	5/6	100 v.c.k. 2
7E6P	5/6	12A74	6/6	EV80	5/6	EV80	5/6	T198	5/6	100 v.c.k. 2
7E6Q	5/6	12A75	6/6	EV81	5/6	EV81	5/6	T199	5/6	100 v.c.k. 2
7E6R	5/6	12A76	6/6	EV82	5/6	EV82	5/6	T200	5/6	100 v.c.k. 2
7E6S	5/6	12A77	6/6	EV83	5/6	EV83	5/6	T201	5/6	100 v.c.k. 2
7E6T	5/6	12A78	6/6	EV84	5/6	EV84	5/6	T202	5/6	100 v.c.k. 2
7E6U	5/6	12A79	6/6	EV85	5/6	EV85	5/6	T203	5/6	100 v.c.k. 2
7E6V	5/6	12A80	6/6	EV86	5/6	EV86	5/6	T204	5/6	100 v.c.k. 2
7E6W	5/6	12A81	6/6	EV87	5/6	EV87	5/6	T205	5/6	100 v.c.k. 2
7E6X	5/6	12A82	6/6	EV88	5/6	EV88	5/6	T206	5/6	100 v.c.k. 2
7E6Y	5/6	12A83	6/6	EV89	5/6	EV89	5/6	T207	5/6	100 v.c.k. 2
7E6Z	5/6	12A84	6/6	EV90	5/6	EV90	5/6	T208	5/6	100 v.c.k. 2
7E7	5/6	12A85	6/6	EV91	5/6	EV91	5/6	T209	5/6	100 v.c.k. 2
7E7A	5/6	12A86	6/6	EV92	5/6	EV92	5/6	T210	5/6	100 v.c.k. 2
7E7B	5/6	12A87	6/6	EV93	5/6	EV93	5/6	T211	5/6	100 v.c.k. 2
7E7C	5/6	12A88	6/6	EV94	5/6	EV94	5/6	T212	5/6	100 v.c.k. 2
7E7D	5/6	12A89	6/6	EV95	5/6	EV95	5/6	T213	5/6	100 v.c.k. 2
7E7E	5/6	12A90	6/6	EV96	5/6	EV96	5/6	T214	5/6	100 v.c.k. 2
7E7F	5/6	12A91	6/6	EV97	5/6	EV97	5/6	T215	5/6	100 v.c.k. 2
7E7G	5/6	12A92	6/6	EV98	5/6	EV98	5/6	T216	5/6	100 v.c.k. 2
7E7H	5/6	12A93	6/6	EV99	5/6	EV99	5/6	T217	5/6	100 v.c.k. 2
7E7I	5/6	12A94	6/6	EV100	5/6	EV100	5/6	T218	5/6	100 v.c.k. 2
7E7J	5/6	12A95	6/6	EV101	5/6	EV101	5/6	T219	5/6	100 v.c.k. 2
7E7K	5/6	12A96	6/6	EV102	5/6	EV102	5/6	T220	5/6	100 v.c.k. 2
7E7L	5/6	12A97	6/6	EV103	5/6	EV103	5/6	T221	5/6	100 v.c.k. 2
7E7M	5/6	12A98	6/6	EV104	5/6	EV104	5/6	T222	5/6	100 v.c.k. 2
7E7N	5/6	12A99	6/6	EV105	5/6	EV105	5/6	T223	5/6	100 v.c.k. 2
7E7O	5/6	12A100	6/6	EV106	5/6	EV106	5/6	T224	5/6	100 v.c.k. 2
7E7P	5/6	12A101	6/6	EV107	5/6	EV107	5/6	T225	5/6	100 v.c.k. 2
7E7Q	5/6	12A102	6/6	EV108	5/6	EV108	5/6	T226	5/6	100 v.c.k. 2
7E7R	5/6	12A103	6/6	EV109	5/6	EV109	5/6	T227	5/6	100 v.c.k. 2
7E7S	5/6	12A104	6/6	EV110	5/6	EV110	5/6	T228	5/6	100 v.c.k. 2
7E7T	5/6	12A105	6/6	EV111	5/6	EV111	5/6	T229	5/6	100 v.c.k. 2
7E7U	5/6	12A106	6/6	EV112	5/6	EV112	5/6	T230	5/6	100 v.c.k. 2
7E7V	5/6	12A107	6/6	EV113	5/6	EV113	5/6	T231	5/6	100 v.c.k. 2
7E7W	5/6	12A108	6/6	EV114	5/6	EV114	5/6	T232	5/6	100 v.c.k. 2
7E7X	5/6	12A109	6/6	EV115	5/6	EV115	5/6	T233	5/6	100 v.c.k. 2
7E7										

C. T. SUPPLY Co.

THE CAR RADIO SPECIALISTS

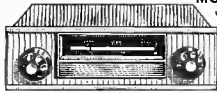
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FIT ANY CAR - ALL TRANSISTOR
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LESS THAN 1/2 PRICE

MODEL 1. 2-Wave full coverage.



£20 **£8-0-0**

or £3. 0. 0. deposit, followed
 by 9 monthly instalments
 of 15/- (total £3. 15. 0)

More facts about Car Radio.

1. size 7" x 5 1/2" x 2" STANDARD SIZE FIT ANY CAR.
 2. 12 volt positive earth.
 3. Medium Wave (87 mm) - 550 mm)
 4. Aerial trimmed ESPECIALLY FOR PIRATE STATIONS.
 5. 3 Watt output. FRONT PLATES MAY VARY
- OPTIONAL EXTRAS 7" x 4" Matching Speaker 12/6d 3 stage telescopic aerial 28/6 18/6 POSTAGE 6/6.



MODEL 2.

£6-6-0

More facts about Car Radio.

1. Full Medium Wave coverage to get Home/Light and all Pirate Stations.
 2. size 7" x 5 1/2" x 2" STANDARD SIZE FIT ANY CAR.
 3. 12 volt positive earth.
 4. Aerial trimmed ESPECIALLY FOR PIRATE STATIONS.
 5. 3 Watt output. FRONT PLATES WILL VARY
- OPTIONAL EXTRAS 7" x 4" speaker 12/6d 3 stage telescopic aerial 18/6. POSTAGE 6/6d.



MODEL 3.

14/29NS

or Deposit
 £4. 0. 0. 8 monthly payments
 of £1. 9. 6d. (Total £17. 5. 0).

- 5—Push button. Medium/Long Wave "EASI-TUNE". British Made 4 watt output—instant station selection. Fully Transistorised and guaranteed.
- 12 volt negative or positive earth. Easy fitting to any car.
- FREE Fitting Kit 'FREE Speaker and baffle.
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Use this coupon as your ready made Order form.
 Please send me Car Radio. I enclose cash/cheque
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 18/6. Aerial.
 12/6 Speaker.

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C. T. Supply Co., 10-12 Hainault St., Ilford, Essex. TEL: ILF 4564



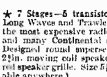
All components used in our transistors may be purchased separately if desired. Parts price lists and easy build plans available separately at prices stated.

NEW TRANSONA FIVE



7 Stages—6 transistors and 2 diodes.
 Covers M. and L. Waves and Traveller Bands. Incorporating ferrite rod aerial, tuning condenser, volume control, and new type fine tone super dynamic 2 1/2" speaker, attractive case. Size 9 1/2" x 4 1/2" x 1 1/2" approx.
Total cost of all parts NOW ONLY £2/6
 Parts price list and easy build plans 2/- (Free with kit.) P. & P. 2/6.

POCKET FIVE



7 Stages—6 transistors and 2 diodes. Covers Medium and Long Wave and Traveller Bands. Features specially fitted to suit the most expensive radios. On test. Home, Light, Intermediate and many Continental stations were received loud and clear. Improved aerial incorporating ferrite rod aerial and fine tone 2 1/2" moving coil speaker, built into attractive black case with red speaker grille. Size 7 1/2" x 4 1/2" x 1 1/2" (with 220 battery, available any where).
Total cost of all parts NOW ONLY 42/6
 Parts price list and easy build plans 1/6 (Free with kit.) P. & P. 2/6.



Med. and Long Wave version with miniature speaker. **29/6** P. & P. 2/6.

ROAMER SIX NEW!



8 Stages—6 transistors and 2 diodes.
 8 wavebands.
 * Now with Philips micro-alloy R.F. transistors. Listen to stations half a world away with this 6 waveband portable. Tunable on Med and Long Wave, Traveller Bands and Three Short Waves. Push-pull output. Benelux ferrite rod aerial and telescopic aerial for short wave. Ten grade transistor. 2 inch speaker. Transistorised with gift fitting. Size 7 1/2" x 5 1/2" x 1 1/2". * Extra hand for easier tuning of L.W. etc.
Total cost of all parts NOW ONLY £3. 19. 6
 Parts price list and easy build plans 2/- (Free with kit.) Carrying strap 1/6 extra. P. & P. 2/6.

NEW! ROAMER SEVEN Mk. IV

7 WAVEBAND PORTABLE OR CAR RADIO
 AMAZING PERFORMANCE AND SPECIFICATION
 * NOW WITH PHILIPS MICRO-ALLOY R.F. TRANSISTORS.
 * Only tunable on all wavebands.
 8 Stages—7 transistors and 2 diodes.
 Covers M. and L. Waves, Traveller Bands and three Short Waves to approx. 18 stations. Push-pull output for room-filling volume from rock found 2 1/2" 4 ohm speaker. Air cooled general tuning condenser. Ferrite rod aerial for M. & L. Waves and telescopic aerial for 3 Short Waves. Metal leather case with gift fitting and shoulder straps and straps. Size 9 1/2" x 5 1/2" approx. The perfect portable and the ideal car radio. (Uses P.C. batteries available any where).
 * Extra hand for easier tuning of pirate stations, etc.
Total cost of all parts NOW ONLY £5. 19. 6
 P. & P. 2/6.



SUPER SEVEN

8 Stages—7 transistors and 2 diodes.
 Covers M. and L. Waves and Traveller Bands. The ideal radio for home, car or can be fitted with carrying strap for outdoor use. Completely portable—built-in ferrite rod aerial for wonderful reception. Special direct incorporating 2 I.F. stages, push-pull output, 2" speaker (will drive larger speaker). Size 7 1/2" x 5 1/2" x 1 1/2". (Uses P.C. batteries, available any where).
Total cost of all parts NOW ONLY £3. 19. 6
 Parts price list and easy build plans 2/- (Free with kit.) P. & P. 2/6.

NEW!! TRANSONA SIX

8 Stages—6 transistors and 2 diodes.
 A top performance receiver covering full M. & L. Waves and Traveller Bands. Push-pull output. High grade speaker makes listening a pleasure. Ferrite rod aerial. Many stations heard in one covering including Luxembourg, loud and clear. Attractive case in grey with red grille. Size 6 1/2" x 4 1/2" x 1 1/2". (Uses P.C. batteries, available any where).
Total cost of all parts 59/6
 Parts price list and easy build plans 2/- (Free with kit.) P. & P. 2/6.



MELODY SIX NEW!!

8 Stages—6 transistors and 2 diodes. Our latest completely portable transistor radio covering M. and L. Waves. Incorporates pre-selector, heavy duty speaker, two grade transistor, volume control, tuning condenser, wave change slide switch, sensitive fine ferrite rod aerial. Push-pull output. Waveband switching of B.B.C. Home and Light. 280 and many Continental stations. Handsome leather look pocket size case, only 6 1/2" x 3 1/2" x 1 1/2" approx. with gift speaker grille and hand and shoulder straps. **Total cost of all parts £3. 9. 6**
 P. & P. 2/6.

SEND S.A.E. for list of COMPONENT BARGAINS

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61 HIGH ST., BEDFORD. 'Phone: 52320

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(ELECTRONICS LTD)

LONDON'S LARGEST SUPPLIERS OF
ELECTRONIC EQUIPMENT
9 & 10 CHAPEL STREET
LONDON, N.W.1
Tel. PAD 7851 AMB 5125

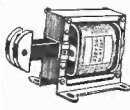


HOURS 9.30-6
OPEN ALL DAY
SAT.
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HEAVY DUTY L.T. TRANSFORMERS ALWAYS IN STOCK.

All Primaries 200-240 VOLTS. TER-
MINAL BLOCK CONNECTIONS.
*DENOTES SHROUDED TYPES.

No.	Sec. Taps.	Amps	Price	Carr.
1.	25-30-35 v.	40	£12 15 0	15/-
2.	25-30-35 v.	20	£7 19 6	9/6
3.	25-30-35 v.	10	£5 10 0	6/6
4.	25-30-35 v.*	2	£2 7 6	4/6
5.	10-17-18 v.*	10	£2 10 0	4/6
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12.	6-12 v.	20	£4 19 6	6/6
12a.	6-12 v.*	20	£4 17 6	6/6
13.	24 v.*	10	£2 17 6	6/6
14.	30-32-34-36 v.*	5	£3 5 0	6/6
15.	6-12 v.	10	£2 10 0	6/6
15a.	12 v.*	5	£1 15 0	4/6
16.	12 v.*	5	£1 12 6	4/6
17.	24 v.*	10	£2 17 6	6/6
18.	24 v.*	5	£2 9 6	4/6
19.	48-56-60 v.*	1	£1 15 0	4/6
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23.	12-24 v.*	2	19 6	3/6
24.	9-15 v.	2	19 6	3/6
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26.	12-18 v.	10	£2 2 6	5/6
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28.	48-56-60 v.*	2	£2 7 6	4/6
29.	12 v.	50	£7 5 0	8/6
30.	12-24-36-48 v.	15	£7 19 6	9/6



VARLEY A.C. SOLENOIDS

BRAND NEW
200-240 v. very powerful,
Base size 2 1/2 x 2 1/2 in. x 2 1/2 in.
Length pulling spindle, 2 1/2 in.
22/6. P.P. 2/6.
D.C. types 180 v. Size 1 1/2 x 1 1/2 in. Length of spindle 1 1/2 in.
6/6. P.P. 2/2. 50 v. D.C., 7/6. P.P. 2/2.

BLOCK CAPACITORS

New and Guaranteed. Send for List, over 2,000 in stock at a fraction of maker's price.

LATEST ARRIVALS

Maker	Mid.	D.C.V.	Temp	Price	Carr.
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T.C.C.	10	350	60°C	6/6	2/6
T.C.C.	8	1,500	60°C	17/6	3/6
T.C.C.	8	750	60°C	8/6	2/6
T.C.C.	8	600	60°C	7/6	2/6
T.C.C.	8	400	71°C	6/6	2/6
T.C.C.	2	2,000	60°C	12/6	2/6
Dubilier	8	600	60°C	7/6	2/6
Dubilier	4	800	71°C	5/6	2/6
Dubilier	1	3,000	100°C	17/6	2/6
Dubilier	0.5	5,000	60°C	17/6	2/6
Dubilier	0.1+0.1	6,000	71°C	8/6	2/6

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Mid.	D.C.V.	Price	Postage
15	1,000	22/6	3/6
12	1,000	19/6	3/6
10	1,500	19/6	3/6
8	600	8/6	2/6
7	600	6/6	2/6
10 tubular	300	6/6	2/6

SOLARTRON POWER SUPPLY SUB-UNIT

TYPE AS955.
A.C. input 100-125 v. and 200-240 v. Stabilised D.C. output continuously variable between 200 and 200 v. at 200 mA. Also two 6.3 v. A.C. at 6 a. and one at 2 a. Supplied BRAND NEW with instruction manual. Manual sent on a returnable deposit of 10/- £12/10/0. Carr. 10/0.

400 WATT AUTO TRANSFORMERS

By famous maker, fully tropicalised. Sapped 200-210-220-230-240-250, 50-75-85-100-105-110-150-160 v.
Table top connections, 49/6. Carr. 5/-.
*See P.P. 2/6.

DAVID BROWN GEARS, LTD. RADICON WORM DRIVE REDUCERS.

Reduction 25:1. Size 1 1/2 in. RHU. Drive spindle 3/8 in. dia. Reduction spindle 3/8 in. dia. Brand new in maker's carton, 65/-. Carr. 5/-.
*See P.P. 2/6.

SMITH 4 MINUTE TIMERS

Switch Contacts 15amps, 250 v. A.C. Complete with chrome bezel and knob. Brand new, 17/6. P.P. 2/6.

DRAYTON THERMOSTATS TYPE CT

Flow range 100-200 deg.F. Switch contacts 15 amp, 250 v. A.C. Make, break or change over-circuit. Complete with metal straps, flexible conduit and data sheet. 35/-. P.P. 5/-. Brand new.

TEDDINGTON AIR PRESSURE SWITCHES TYPE 13A/1A/B

Single pole 15 amp, 250 v. switch contacts. Approx. 1/6. Pressure, 12/6. P.P. 2/6. Brand new.

EAC LTD. A.C. 0-300 v. METERS

50 cycles, 6 in. flush mounting, iron case, protected. Panel hole 6 1/2 in. Supplied new, in sealed cartons at a fraction of maker's price, 75/-. Carr. 3/-.
*See P.P. 2/6.

SIEMENS MINIATURE RELAYS

Plug-in type, 9,000/62 makes, 16,000/121 makes, 5,500/24 CQ. Complete with base, all at 8/6 each. 1/6 P.P. G.P.O. 3,000 type 500/1 change over contacts 15/- P.P. 2/2.

AUTO TRANSFORMERS

240-110 v. 2,000 watts. Completely enclosed in beautifully finished case. Fitted with 2 two-pin American sockets or terminal blocks, neon indicator, on/off switch, and carrying handle. £10/9/6. Carr. 7/6. 1,500 watts, £7/10. Carr. 7/6. Also available, completely shrouded, fitted with 2-pin American sockets or terminal blocks. Please state which type required. Wattage Price Carr. 1,000 £4 19 6 7/6 500 £3 15 0 6/6 300 £2 9 6 5/6 150 £1 11 6 4/6 80 £1 12 6 4/6

A.C. SYNCHRONOUS GEARED MOTORS

200-250 v., very powerful, 40 R.P.M. Size 2 1/2 x 1 1/2 in. Easily adapted to oscillate up to half a revolution. 12/6. P.P. 2/6.

REDCLIFFE HT TRANSFORMERS

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400	1.7-0.5	TTGD	65/6	5/6
758	0.7	TT	49/6	6/6
1,622	1	ST	57/6	6/6
1,624	1	TT	65/6	6/6
1,540	0.5	TT	75/6	6/6
12,000	0.03	TTGD	45/6	6/6
0.1	2	ST	65/6	6/6
1	16	STF	17/6	4/6
130	1	STF	7/6	2/6

ST - Single Tube TT - Twin Tube GD - Gear Drive. F = Fixed.

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All Primaries 200, 230, 250 v.

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D	3-4-5-6-8-9-10-12-15-18-20-24-30 v.	8 65/6 7/6
E	6-7-8-10-14-15-17-19-25-31-33-40-50 v.	3 49/6 5/6
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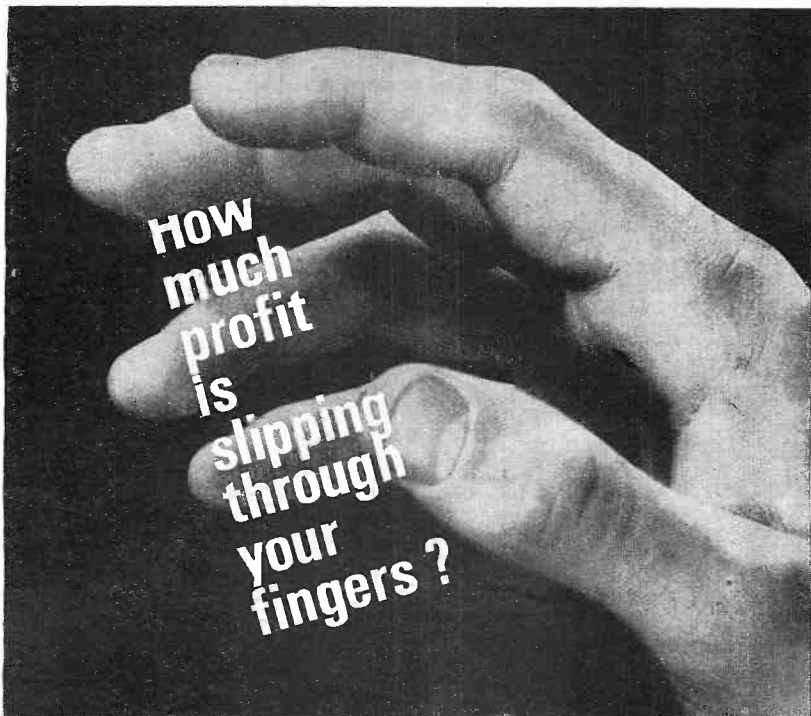
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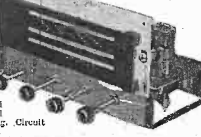


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Please write with full details, quoting reference WW2990.B to:

Technical Staff Officer, English Electric House, Strand, London, W.C.2.



MONTREAL QUEBEC

PYE CAMBRIDGE WORKS, Ltd., Hale Rd., Cambridge.
 * SINGLE sideband equipment.
 * VHF radiotelephone equipment.
 * H.F. reproduction equipment.
 WE require trained personnel for production testing and fault finding of micro equipment.
 WE have limited vacancies for more senior and experienced men with driver, who can lead small teams entrusted on this work.
 WE have also limited vacancies for persons of less experience who can be trained for such work.
 APPLY TO: The personnel Manager. (131)

SENIOR Electronics Engineers—Vacancies exist for 2 senior engineers, B.Sc. or B.Eng. or equivalent. Candidates must be first-class technicians with some practical experience and must be capable of administration and co-ordination of complete projects. Considerable periods will be spent abroad and only single men will be considered.—Box W.W. 1467, Wireless World.

GRAMPIAN REPRODUCERS, Ltd., Hanworth Trading Estate, Feltham, Middlesex, require a senior engineer for development of audio frequency transmitter equipment in public address applications. 11405

COLOUR television engineer, well versed latest techniques, required by new development company, apply in confidence, giving age, details of experience.—Box W. W. 54, c/o Wireless World.

A FULL-TIME technical experienced Salesman required for retail sales; write giving details of relevant experience, salary required to—The Manager, Henry's Radio, Ltd., 303, Edgware Rd., London, W.4
SALES engineers with good knowledge of Hi-Fi equipment and the electronic industry are offered responsible and remunerative position in various parts of the United Kingdom. Applicants are invited to write giving full details of their work to: Bridgeway, Ltd., 15-22, Great Russell St., London, W.C.1. 11644

VACANCIES FOR PRODUCTION INSPECTORS AND TECHNICAL AUTHORS MINISTRY OF DEFENCE (NAVY DEPARTMENT)

Production Inspectors and Assistant Production Inspectors are required in the Navy Department for a wide range of duties concerned with production, inspection and testing of equipment in the fields of Radio Communication, Sonar, Radar, Missile and Weapon Control, and Computers. The work includes Test and Inspection of Ladders, Testing and Tuning of Weapon systems in Ships, Calibration, Qualification Approval, preparation of Production Specifications, Technical Issues with Guidance for Research and Development Teams. There are also vacancies in the field of Calibration and Test Technicians to prepare Technical Handbooks used in maintenance of Weapon and Ship Production equipment in the Fleet.

LOCATION

Portsmouth, Bath, Cuppage, Portland, Slough and other manufacturing and shipbuilding centres. At Portland mainly concerned with production of Submarine Weapons and Sonar, at Portsmouth with Radar and Communication equipment. Portsmouth is also the normal base for mobile teams responsible for setting to work, tuning and testing of Weapon systems in ships, but work also in other main shipbuilding areas. Posts at Bath are concerned with production of surface and Underwater Weapons, at Slough with Naval Ordnance, and at Cuppage with Test, Inspection, Calibration and Condition.

Technical Author posts at Bath, Portsmouth, Portland and Slough.

QUALIFICATIONS

and apprenticeship; practical experience in the above fields; possession of an ONC (or equivalent) in Electronic or Electrical Engineering; Possession of Higher Technical qualifications an advantage.

SALARY SCALES

Salary scales are—
 Production Inspectors (Technical Class Grade 1) £1,120 per annum by four annual increments to £1,228 per annum.
 Assistant Production Inspectors (Technical Class Grade 11) £704 per annum to £1,120 per annum by four annual increments.
 (Above rates under review. Posts in London area attract an allowance)

Prospects of preferable employment and promotion to higher grade posts.

APPLICATIONS

Write to Superintendent of Production Pool (Naval), Ministry of Defence (Navy) Offices, Kingsley Road, Harlow, Essex for Application Form. 1468

ELECTRONIC Service Engineers required to service and install Airborne Navigation Equipment at London Airport, Home Counties and Overseas.—Apply The Decca Navigator Co. Ltd., Spar Road, Feltham, Middlesex. Tel. Feltham 3188.

AIRCRAFT Radio Radar Engineers and Mechanics with specific workshop experience in one of the following: X Band Radar, VHF, HF, ILS/AOR, or ADP; 40-hour week, 3 weeks holiday, pension scheme; apply—Miss J. D. Trotter, Personnel Officer (c/o) Ltd., Willow Rd., Colnbrook, Slough, Bucks. 11496

OPPORTUNITY for man to join electronics team enhanced on wide variety of applications to scientific research. Applicants should be able to read, understand and have experience in design or development of electronic equipment as well as servicing experience. Salary up to £2,250 p.a. inclusive. Application forms from Establishment Officer, University College London, Gower St., W.C.1, quoting Chem.6. 11483

PENNSYLVANIA

A well-known American Company based in the Pennsylvania area of America, at present engaged in the design, development and application of Semi-Conductor devices.

require

Qualified engineers experienced in Semi-Conductor work to join the company. Top salaries will be paid to the successful applicants, plus an excellent relocation allowance including fares for their families.

APPLY IN THE FIRST INSTANCE TO—

TEMTECH AGENCY LIMITED

7, Southampton Place, London, W.C.1.

Telephone HO1born 7033

BOROUGH POLYTECHNIC

Borough Road · London · S.E.1

Applications are invited for the following posts in the Department of Electrical and Electronic Engineering:

Chief Technician (Electronic)
 Chief Technician (Instrumental Workshop—Instrument-type)
 Senior Technician (Electrical Machines)
 Technician (Electrical Machines)
 Technician (Electronics)

A formal experience in Electrical/Instrumental workshop practice is essential in all posts, except the senior posts, although this is desirable for the senior post. Services in one of the above fields would be most appropriate.

Salary, according to age, experience and qualifications:

Chief Technician: £1,120-£1,200 per annum
 Senior Technician: £811-£1,115 per annum
 Technicians: £650-£820 per annum

Apply by writing, stating the post for which application is made, giving the address of your present or the Borough.



The Civil Service

Professional and Technical Appointments

POST OFFICE EXECUTIVE ENGINEERS

Forty-five posts in London and Provinces for mechanical, electrical and electronic engineers to develop and design communications systems and postal service equipment.

QUALIFICATIONS: Degree or Dip. Tech. in Mechanical or Electrical Engineering, Physics, or Applied Physics or, exceptionally, very high professional attainment. Final year students may apply.

SALARY (Inner London): £877—£1,806. Salary under review. Promotion prospects.

AGE: At least 21 and normally under 35 on 31.12.66. Some extensions for service in H.M. Forces or Overseas Civil Service. (Reference: S/322).

ELECTRICAL AND MECHANICAL ENGINEERS

For design and installation of Heating, Ventilating and other building services urgently required by Ministry of Public Building and Works to fill 30 posts in London and the provinces. The work is varied and will provide experience in latest techniques. A further 150 vacancies for Electrical and Mechanical Engineers exist in other Government Departments.

QUALIFICATIONS: Degree or Dip. Tech. with 1st or 2nd class honours, or have passed all examinations for A.M.I.E.E., A.M.I.Mech.E., A.M.I.E.R.E., A.M.I. Prod.E., or A.F.R.Ac.S.

SALARY (Inner London): £1,143 (at 25)—£1,718. Salary under review. Promotion prospects.

AGE: Normally at least 25 and under 35 on 31.12.66. Some extensions for service in H.M. Forces or Overseas Civil Service. (Reference: S/85).

ENGINEERING DRAUGHTSMEN

Vacancies in Ministry of Public Building and Works, Ministry of Defence, Post Office, and in other Departments for Engineering Draughtsmen in the fields of **MECHANICAL, ELECTRICAL, and HEATING AND VENTILATING ENGINEERING.**

QUALIFICATIONS: O.N.C. (or equivalent) in appropriate subject, three years' training and, in addition, at least one year's drawing office experience.

SALARY (Inner London): £718 (at 20)—£1,108 (at 28 or over)—£1,209. Salary under review. Annual leave allowance 3 weeks and 3 days rising to 6 weeks.

AGE: At least 20. Promotion prospects. Where appropriate, time off for further technical study may be given. (Reference: S/68).

The above posts are pensionable and **APPLICATION FORMS** are obtainable from the Secretary, Civil Service Commission, Savile Row, London, W.1. Please quote appropriate reference.

English Electric Leo Marconi

Computer servicing

English Electric Leo Marconi is Britain's foremost computer company. Ever increasing sales mean that we need more engineers to maintain computers on our customers' premises.

Experience in the development, testing or servicing of transistorised electronic equipment and the ability to understand the logic of advanced computers are the qualifications we are looking for.

Training, salaries and fringe benefits are all that you would expect from a leader in the computer field and prospects are limited only by ability.

Jobs exist in most parts of the country, but the greatest number of vacancies is in the London Area, the Midlands and South West Lancashire.

To obtain more information write to:—
The Personnel Officer,
Dept. W.W.M.16,
English Electric-Leo-Marconi Computers Ltd.,
24, Minerva Road,
London, N.W.10.



AN ENGLISH ELECTRIC COMPANY

Instrumentation Technicians and Mechanics

If you are a Technician or Mechanic we invite **YOU** to apply for one of the several vacancies in our Commissioning or Laboratory Section.

What do we want?

You must have experience in the operation and/or commissioning of some of the following systems

- (a) RELAY CIRCUITS, MAGNETIC AMPLIFIERS, NUCLEONIC INSTRUMENTS, PRESSURE TRANSDUCERS WITH ELECTRICAL OUTPUT, HEALTH PHYSICS INSTRUMENTS
- or (b) FLOW MEASUREMENT, TEMPERATURE MEASUREMENTS, CONTROL VALVES, PNEUMATIC AND HYDRAULIC CONTROL SYSTEMS

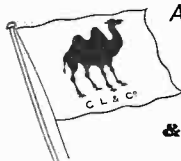
What qualifications must you have?

- (a) H.N.C. ELECTRICAL OR MECHANICAL
- (b) O.N.C. ELECTRICAL OR MECHANICAL
- (c) TRADE APPRENTICESHIP

What can we offer?

- (1) FOR APPLICANTS WITH H.N.C. AND CONSIDERABLE EXPERIENCE, TECHNICIAN GRADE I
- (2) FOR APPLICANTS WITH H.N.C. AND LIMITED EXPERIENCE OR O.N.C. AND CONSIDERABLE EXPERIENCE, TECHNICIANS GRADE II
- (3) IF YOU HAVE SERVED AN APPRENTICESHIP AND HAVE SOME EXPERIENCE IN INSTRUMENTATION WE WILL OFFER TO SUCCESSFUL APPLICANTS THE POST OF MECHANIC, AN HOURLY PAID JOB WITH GOOD PROMOTION PROSPECTS TO TECHNICIAN GRADE

The successful staff applicants will be required to participate in our super-annuation scheme and will be given assistance with removal expenses in appropriate cases.



*All applications to be addressed to the Personnel Officer,
quoting reference IT/1*

CAMMELL LAIRD

& Co (Shipbuilders & Engineers) **Ltd · Birkenhead**

RACAL

Need a technical author

Racal Instruments Limited have a vacancy for a Technical author to prepare handbooks for their new range of Digital Measuring Instruments. Candidates should have formal electronic training; some experience in the application of logical techniques would be an advantage. Consideration will be given to engineers who have a desire to enter this field but have not had previous experience as Technical Authors. The work calls for personal initiative. The successful applicant will be based in the Reading area.

And a technical publications writer

To work on a range of publications dealing with HF Communications Systems and Digital Instrumentation. Ideally, candidates will have had a basic training in Electronics, or at least a background knowledge and be willing in the first instance to generally assist in the production of Commercial Technical Publications. A good educational level including 'O' or 'A' at English is essential. Candidates need not have had previous experience in this type of work and will be assisted and encouraged by experienced writers. This is an opportunity to train for a worthwhile and interesting profession.

Attractive conditions of employment are offered including entry to the Company's Superannuation Scheme, etc. Applications in writing, enclosing full personal details to:-

RACAL
THE RACAL GROUP

Personnel Manager
Racal Electronics Ltd.
Western Road, Bracknell,
Berkshire

TELECOMMUNICATIONS SUPERINTENDENT

Required by NIGERIAN RAILWAY CORPORATION on contract for two tours, each of 18 months in first instance. Salary £2,475 gross a year. Gratuity at rate of 20% of total salary drawn. Quarters provided at moderate rental. Free passages. Liberal leave.

Candidates must be experienced telecommunication engineers possessing a recognised University Degree/Diploma, or be A.M.I.E.E. or A.M.I.E.R.E. P.M.G. Certificate desirable but not essential. They should have sound experience of railway telecommunication systems, and be capable of advising the Corporation on problems arising during the installation of a complex communications system. Possibility of secondment will be considered for successful applicant.

Apply to CROWN AGENTS, M. Dept., 4 Millbank, London, S.W.1 for application form and further particulars, stating name, age, brief details of qualifications and experience, and quoting reference M2R/62781/W.F.

RADIO TECHNICIANS

A number of suitably qualified candidates will be required for training, leading to permanent and pensionable employment, normally at Cheltenham but with opportunities for service abroad or assignment to other U.K. establishments including London.

Applicants must be 19 or over and be familiar with the use of Test Gear and have had Radio/Electronic Workshop experience. They must offer at least 'O' level (GCE passes in English language, Mathematics and/or Physics or hold Intermediate Certificate or equivalent technical qualifications.

Pay according to age, e.g. at 19 £747; at 25 £902 (which pay on entry) rising by four annual increments to £1,104.

Prospects of promotion to grades in salary range £1,032-£1,691.

Annual leave allowance of 3 weeks 3 days, rising to 4 weeks 2 days.

Normal Civil Service sick leave regulations apply.

Apply:

Personnel Officer (R/T/3),

Government Communications Headquarters,

Osley, Priors Road, Cheltenham, Glos.

AIRCRAFT Radio Engineers and Mechanics with specific workshop experience in one of the following: VHF, HF, or ADF, 40-hour week, 3 weeks' holiday; Periodic Engine Director, Air Transport (Charters) (G.T.), Ltd., Willow Road, Colnbrook, Bucks. 11485

INTERNATIONAL ALLOYS LIMITED, Aylesbury, require Electronics Technician for maintenance and fault finding on direct reading spectrographic equipment; experience on this type of equipment desirable but not essential; experience on high voltage equipment advantageous; salary based on qualifications and experience; hours of work 37 per week; additional hours paid at overtime rates; company operates pensions and life insurance schemes; three weeks' holiday per year after first year; canteen facilities available—Apply to Personnel Manager. 11488

INSTITUTE OF CANCER RESEARCH: ROYAL CANCER HOSPITAL, ELECTRONICS TECHNICIAN required in the Chester Beatty Institute, Fulham Rd., S.W.3 to take charge of the servicing of equipment (e.g. spectrometers and other instruments) and the construction and development of prototype apparatus. Applicants will be required to hold Technician's Grade, M.R.C. scale, E955-£1,265 with London Allowance, depending on age, qualifications and experience. Apply with names of two referees to the Secretary, 34, Sunnier Place, S.W.7, quoting ref. 501/B/624. 11474

HYDROLOGICAL research unit: electronic engineer: An electronic engineer is required by the hydrological Research Unit, Natural Environment Research Council, Howbery Park, Wallingford, Berkshire, to assist in the development and construction of mechanical instruments for hydrological measurements. The post is at assistant experimental officer grade and applicants should have an H.N.C. in electrical engineering and some years' experience in electronic development work, examples of salary: £741 at 21, £970 at 25; a day release scheme operates for further studies.—Apply to the above address, telephone number Wallingford 2581. 11459

BOOKS, INSTRUCTIONS, ETC.

CIRCUITS for audio and tape recording 7.5 raw available at all good bookshops or 8.6 post free from Haymarket Press, Ltd., 9, Harrow Rd., London, W.2.

MANUALS, circuits of all British ex-W.D. 1919-45 wireless equipment and instruments from original R.E.M.E. instructions; sale for 15s. over 70 types—W.H. Batley, 167a, Moffat Rd., Thornton Heath, Surrey. 11479

BOOKS WANTED

MANUALS or instruction books, data, etc. on American or British Army, Navy or Air Force types of electrical equipment—H. Harris, 93, Wardour St., London, W.1. 11416

ELECTRONIC ENGINEERS

Service Engineers required for London Area and Provincial Offices of well-known Crown manufacturing Electronic Desk Calculating Machines. Applicants should possess a sound knowledge of basic electronics with experience in electronics, Radio, Radar and T.V., or similar field. Position is permanent and pensionable. Comprehensive training on full pay will be given to successful applicants. Please send full details of experience to:

Service Manager,
SUMLOCK COMPOTMETER LTD.,
102/108 Clerkenwell Road,
London, E.C.1.

Commissioning Organisation

SENIOR WEAPONS ENGINEER

An experienced electronic engineer with administrative ability is required to control the weapons section of our machinery test organisation.

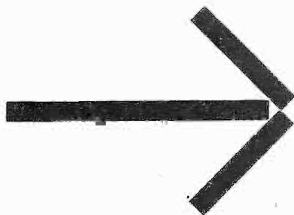
Likely background would be that of a commissioning engineer on radar, radio and/or sonar equipment with some control over other engineers.

Salary will be commensurate with qualifications and experience. The successful applicant will be required to participate in our superannuation scheme. Assistance in appropriate cases will be given with removal expenses.



Applications initially to the Personnel Officer

**CAMMELL LAIRD
& CO (Shipbuilders & Engineers) LTD
BIRKENHEAD**



Electronics Technicians

The Job

Servicing a wide range of high quality electronic instruments—oscilloscopes, digital and valve voltmeters, power supplies, pulse generators, etc. Location will be at Farnborough, in one of the most modern and attractively situated companies in South East England.

The Men

Radio and T.V. service engineers; technicians leaving the armed forces; men with industrial experience. The real requirement is a good knowledge of electronics and experience in fault diagnosis and rectification, preferably backed up by a sound technical training (City & Guilds Telecommunications for example).

The Benefits

Good pay; first-class fringe benefits; up-to-date workshops, superbly equipped; on-the-job training; day-release for suitable candidates; prospects of career advancement and broader industrial experience within the Company; security; companionable working atmosphere; assistance for people willing to move into the area.

The Right Thing

to do now is to get yourself an interview; talk to the people in our Service Department, and find out for yourself the advantages of linking your future to SOLARTRON. Write a descriptive letter to:



C. S. J. Mardell, Personnel Officer,
THE SOLARTRON ELECTRONIC GROUP LTD.,
Victoria Road, Farnborough, Hants.

TECHNICAL CLASS VACANCY AT R.A.F. CHESSINGTON

The Ministry of Defence (Air Force Department) has a vacancy for a Civilian Technical Class Grade III officer at No. 248 Maintenance Unit, Royal Air Force Chessington.

Duties

The review of serviceable stocks of medical electronic equipment held at RAF Chessington and testing before issue, the small scale production of locally-developed equipment, and the servicing and repair of equipment at hospitals and other medical establishments.

Qualifications

Applicants must be British subjects. They should have served an apprenticeship, and hold the ONC or an equivalent qualification, but applicants with substantial experience in the electronics field with technical knowledge up to ONC standard will be considered.

Pay

£796 at age 21 increasing by annual increments to a maximum of £1,129. Age 28 and over start at £1,009. These rates are under review.

Prospects for Pension

Established appointment (for pension and gratuity) can be obtained by passing Civil Service Commission Open Competitions once the ONC or an equivalent certificate is held.

Studies for ONC, HNC, etc.

Encouraged by release from work one day a week and by payment of fees, etc., by Department.

Holidays, Sickness, etc.

Five day week, Annual leave three weeks, three days at start increasing to six weeks. Sick leave benefit.

Selection is by interview at Chessington.

Application forms from:— **Officer Commanding,**
No. 248 Maintenance Unit,
Royal Air Force Chessington,
Chessington,
Surrey.



TELECOMMUNICATIONS

We have vacancies for Fault Finders, Testers, and Inspectors to work on interesting and advanced equipment including H.F. SINGLE SIDEBAND, V.H.F. RADIO TELEPHONES, U.H.F. MINIATURE EQUIPMENT.

Transistor experience is desirable but not essential. Vacancies exist at all levels and training will be given where necessary.

Apply: **Personnel Manager,**
CAMBRIDGE WORKS LTD.,
Haig Road,
Cambridge.

BERRY'S RADIO

Require

**TECHNICAL STAFF,
ENGINEERS AND
SALES ASSISTANTS**
GOOD PROSPECTS, PERMANENCY

Write giving full details of experience, past situations, etc. in confidence to

25 HIGH HOLBORN, LONDON, W.C.1

LANCASHIRE CONSTABULARY

Wanted Radio Technicians

The Lancashire Constabulary has vacancies for Radio Technicians in the Wireless Workshops at Hutton, Preston.

Candidates should have a sound fundamental knowledge and practical experience of Frequency Modulated V.H.F. Radio Telecommunications Engineering. Salary, £820 per annum at 21, rising by eight annual increments to a present top rate of £1,104 per annum; subject to an efficiency bar at age 25.

Lancashire County Council Contributory Superannuation Scheme applies.

Application forms and Conditions of Service from **The Chief Constable, P.O. Box 77, Lancashire Constabulary, Preston.**

Science Research Council RADIO AND SPACE RESEARCH STATION Ditton Park, Slough

Physicists, Electrical Engineers and Mathematicians required for the following posts:—

(1) **Senior Scientific Officers/Scientific Officers** for work on:—

The propagation of radio waves through the ionosphere and the troposphere. Study of the upper atmosphere and ionosphere with apparatus in rockets and satellites. The use of an 80ft. radio telescope in the study of radio waves and their travel through the ionosphere and troposphere. The engineering development of specialised electronic and other apparatus for use in these investigations.

(2) **Experimental Officers/Assistant Experimental Officers** to assist the scientific staff in design and development of apparatus, in experiments and analysing results. Background in physics, electrical engineering or applied mathematics necessary. Opportunities for overseas service.

Qualifications required:—

S.O./S.S.O. 1st or 2nd Class Honours Degree (or equivalent) in appropriate subject plus (for S.S.O.) at least three years post-graduate experience.
A.E.O./E.O. Suitable Degree, Dip. Tech., H.N.C. (or equivalent), or if under age 22, two G.C.E. 'A' levels in scientific subjects.

Salaries:—

S.O./S.S.O. Between £926 and £2,155.
A.E.O./E.O. Between £508 and £1,734.

Send for details to

The Secretary, S.R.C. Radio and Space Research Station, Ditton Park, Slough, Bucks.

ELECTRONICS TEST ENGINEER

An opportunity has arisen for a capable man to fill a vacancy concerned with the testing of electronic instruments and sub-assemblies in our modern scientific instrument factory in Cambridge. Experience of electronic fault-finding, inspection and testing OR radio and television servicing is desirable. Technical qualifications at an intermediate level would be an advantage.

This appointment will carry staff status and an attractive rate of pay.

Please apply in confidence to:

The Personnel Manager,
Unicam Instruments Ltd.,
York Street, Cambridge.
Telephone: 61631.



THE RANK ORGANISATION

projects engineer

Due to the expansion of the Communication Division of Rank Relay Services a vacancy has arisen for a Projects Engineer with detailed knowledge of closed circuit television and audio techniques.

He will be required to produce work costings, prepare tenders and liaise with customers. In addition there will be work of a more practical nature in the way of surveying, commissioning and supervising installations.

The post is based at Whyteleafe, near Purley, Surrey, but considerable travelling throughout the United Kingdom will be necessary. It is desirable that applicants have some formal qualifications such as H.N.C. or City and Guilds Telecommunications although men with good practical experience should certainly apply. An attractive starting salary will be offered along with Company Car and comprehensive fringe benefits.

Please write showing how you meet the above requirements to:—

**The Personnel Officer,
Central Personnel Department,
The Rank Organisation
Limited, 439/445 Godstone
Road, WHYTELEAFE,
Surrey.**



**HOME & LEISURE
SERVICE
DIVISION**

Vacancies in Cambridge

Considerable expansion of our Radio Communications Division is planned and immediate vacancies exist for the following technical staff—

- (1) Senior Development Engineer for S.S.B. Radio Equipment. Experience with solid state circuitry essential.
- (2) Engineer familiar with micro-miniaturisation techniques.
- (3) Three Assistant Circuit Development Engineers with experience of H.F. S.S.B. equipment.
- (4) Six advanced Test Engineers and fault finders for work on sophisticated apparatus.

Really attractive salaries with ample opportunity for advancement will be offered to selected applicants. Send details of your qualifications, background and experience, which will be treated in confidence, to:—

**Personnel Manager, Labgear Limited,
Cromwell Road, Cambridge.**

ARTICLES FOR SALE

GOOD secondhand Ferragros often available—Reg. 2745 (Lon.).
2 Philips pro. 20 tape machines. 2 track stereo. £650 each, new. mint condition: offers.—Box 1469, Wireless World.

PUBLIC Address amplifiers, mains operated. 25w A output. £15; 40w £25, 70w £35, 120w £50; external re. current transformer. Es 10 each.—Harrington Suppliers, Ltd., 435, Green Lanes, London, N.4. Mou. 5241.

DISC cutter with power amplifier, equalizer amplifier, hot stents unit, pick-up and pre-amplifier, spare head, several blankets, etc. New. £150 o.n.o. Further details Malconi Disc Recordings, 592, Hesle Rd., Hull.

AMOEX series 600 professional recording tape, 1575 601, 1800; Altyar on 7in spool, complete with motor, brand new in makers box, retail price £3. Limited quantity available at 67/- for two spools, p. & Co., (for 277, Smith, p. & I. S.)—Pamsey & Co., 95, Tottenham Court Rd., London, W.1. Euston 2300.

FOR sale—1961 Decca River Radar equipment, Type 215, comprising aerial, search, receiver, display unit, motor alternator, voltage controller, hood, vision magnifier unit, pivot mast and seat. Squares, Type 2153, cover for radar mast, wind. Full brochures and address for viewing on application to County Supplies Office, County Supplies Department, Sandring Road, Nuneaton, Kent, to whom offers must be returned on or before 2nd May, 1966.

ELECTROID Voltmeter and Ohmmeter (Electronic Desizes Inc.) Gossor P.M. Receiver Alignment Generator Model 1324 Serial 450; Advance Signal Generator Type 62 Serial 321; Avo Test Bridge, 50 c/s 240V; Micron Crystal Calibrator; pair S.G. Brown Super-K headphones; Stereoephone; Permuton Gain Leverage Meter; peak level reflectograph; quantity air electric trimmer; test leads; capacitors; semi-conductors. Property of deceased electronics engineer. Reasonable offer accepted.—Apply J. W. Barber, Slough Arms, Great West Road, Slough, 141, 2385-526.

HIGH quality professional recording and reproducing equipment complete. Quad 22 stereo pre-amp and two Quad main amplifiers. Quad FM radio unit, Vortexion four-way mixer, Shure pickup and Garrard transcription turntable. Ampex 6 track 1200 series tape recorder. Equipment housed in small canvas cabinet 30 in. section long. Price 250 guineas o.n.o. Two Lockwood burst monitor enclosures complete with 100 W. Tannoy dual-cone speakers. Price 125 guineas o.n.o.—Tel. Weston 5155 or write to Prov. 30, Monkton Court, Melbury Rd., London, W.11.

SYSTEMS TEST ENGINEERS



TELECOMMUNICATIONS LTD. NEWMARKET ROAD, CAMBRIDGE

require Systems Test Engineers for testing complete electronic switching systems.

Applicants should have some experience in the field of transistor switching techniques.

Successful applicants can expect to receive a salary in accordance with their experience and qualifications.

Apply to:

**THE PERSONNEL MANAGER,
TELEPHONE:
CAMBRIDGE 61222**

TEST ENGINEERS

Due to Company expansion, The Decca Navigator Company Limited wishes to recruit the services of Test Engineers for the testing of Radio Navigation and Survey Equipment.

We are world leaders in this field and have an extensive and interesting programme planned for the future.

Applicants with a first-class background of T.V. and Radio Servicing or Telecommunication Electronic and Control Circuiting are required.

The salary paid will be commensurate with ability and experience.

Please apply for interview giving brief details of experience, qualifications and present salary to:

The Personnel Officer, The Decca Navigator Company Limited, 88 Bushey Road, Raynes Park, London, S.W.20. Telephone: WIAbledon 8011.



**ELECTRIC
CONSTRUCTION**

ELECTRONICS ENGINEERS

for work on Thyristor applications. Some experience with transistor circuits desirable.

MINIMUM QUALIFICATIONS H.N.C.

Pensions and Life Assurance and Sick Pay Schemes in operation.
Modern Restaurant and Recreational Facilities available.

Apply

**Personnel Manager,
Electric Construction (W'ton) Limited,
Bushbury Engineering Works,
WOLVERHAMPTON.**

STC INSPECTORS AND TESTERS

Owing to continued production expansion within our Radio Division we urgently require Male Inspectors and Testers. There are vacancies on day work and a night-shift is being started to cope with the additional load.

Applicants must have had at least two years broad experience of inspection or fault finding work on assembled and wired electronic units used in the Radio Communication and Aircraft Navigational Aids field or alternatively some years experience of Radio and Television servicing.

Good rates will be paid to suitable applicants who will also enjoy an excellent variety of welfare amenities, including a sickness, accident benefit and pension scheme.

Write or telephone the Male Employment Officer, ENTenterprise 1234, Extension 437, or call at the Personnel Department.

Standard Telephones and Cables Limited

Oakleigh Road, New Southgate, London, N.11, between the hours of 9-11 a.m. or 2-3.30 p.m. Monday to Friday.

BOROUGH POLYTECHNIC Borough Road, London, S.E.1.

Applications are invited for the post of:

HEAD OF THE DEPARTMENT OF ELECTRICAL AND ELECTRONIC ENGINEERING

which will become vacant on the 1st September, 1966.

The work of the Department includes a well-established Honours Degree course (C.N.A.A.) full-time and part-time courses in preparation for the Part III Examinations of the Institution of Electrical Engineers, and a wide range of special courses of post-graduate standard.

The whole of the Department is to be rehoused in a new building, on which work has recently commenced.

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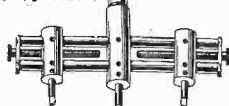
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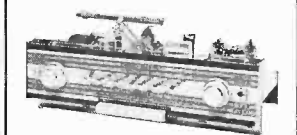
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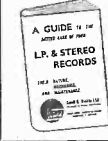
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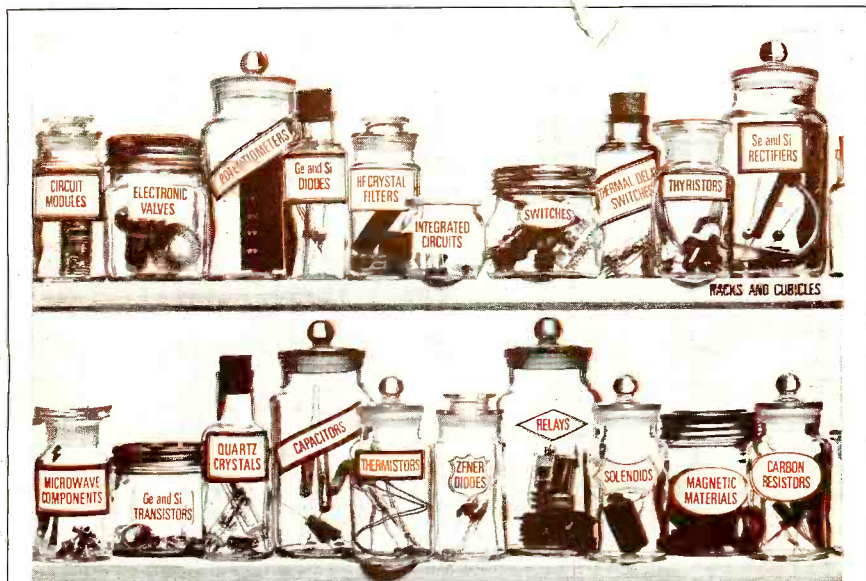
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18	.048	1.219	182
22	.028	.711	536
24	.022	.558	865
26	.018	.46	1,292
28	.014	.375	1,911
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